











TPS51206 SLUSAH1C - MAY 2011 - REVISED AUGUST 2016

TPS51206 2-A Peak Sink / Source DDR Termination Regulator With VTTREF Buffered Reference for DDR2, DDR3, DDR3L, and DDR4

1 Features

- Supply Input Voltage: Supports 3.3-V Rail and 5-V Rail
- VLDOIN Input Voltage Range: VTT+0.4 V to 3.5 V
- VTT Termination Regulator
 - Output Voltage Range: 0.5 V to 0.9 V
 - 2-A Peak Sink and Source Current
 - Requires Only 10-μF MLCC Output Capacitor
 - ±20 mV Accuracy
- VTTREF Buffered Reference
 - VDDQ/2 ± 1% Accuracy
 - 10-mA Sink and Source Current
- Supports High-Z in S3 and Soft-Stop in S4 and S5 with S3 and S5 Inputs
- Overtemperature Protection
- 10-Pin, 2 mm x 2 mm SON (DSQ) Package

2 Applications

- DDR2, DDR3, DDR3L, and DDR4 Memory Power Supplies
- SSTL_18, SSTL_15, SSTL_135 and HSTL Termination

3 Description

The TPS51206 is a sink and source double date rate (DDR) termination regulator with VTTREF buffered reference output. It is specifically designed for low-input voltage, low-cost, low-external component count systems where space is a key consideration. The TPS51206 maintains fast transient response and only requires 1 × 10- μ F of ceramic output capacitance. The TPS51206 supports a remote sensing function and all power requirements for DDR2, DDR3 and Low-Power DDR3 (DDR3L), and DDR4 VTT bus. The VTT current capability is ±2-A peak. The device supports all of the DDR power states, putting VTT to High-Z in S3 state (suspend to RAM) and discharging VTT and VTTREF in S4 or S5 state (suspend to disk).

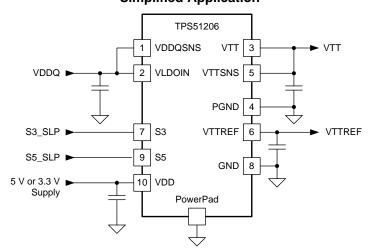
The TPS51206 is available in 10-pin, 2 \times 2, SON (DSQ) PowerPADTM package and specified from -40° C to 85° C.

Device Information⁽¹⁾

PART NUMBER	PACKAGE	BODY SIZE (NOM)
TPS51206	WSON (10)	2.00 mm × 2.00 mm

(1) For all available packages, see the orderable addendum at the end of the datasheet.

Simplified Application



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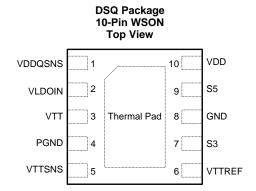
4 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

CI	nanges from Revision B (December 2014) to Revision C	Page
•	Added references to DDR4 compatibility	1
•	Added Receiving Notification of Documentation Updates section	18
•	Added Community Resources section	18
CI	nanges from Revision A (October 2013) to Revision B	Page
•	Added Pin Configuration and Functions section, ESD Ratings table, Feature Description section, Device Functional Modes, Application and Implementation section, Power Supply Recommendations section, Layout section, Device and Documentation Support section, and Mechanical, Packaging, and Orderable Information section	1
CI	nanges from Original (MAY 2011) to Revision A	Page
•	Added minimum and maximum values to the wake up condition of the VDD UVLO threshold voltage specification	5



5 Pin Configuration and Functions



Pin Functions

PIN		1/0	DESCRIPTION	
NAME	NO.	1/0	DESCRIPTION	
GND	8	_	Signal ground	
PGND	4	_	Power GND for VTT LDO	
S3	7	1	S3 signal input	
S5	9	1	S5 signal input	
VDD	10	1	Device power supply input (3.3 V or 5 V)	
VDDQSNS	1	I	DDQ sense input, reference input for VTTREF	
VLDOIN	2	I	Power supply input for VTT/ VTTREF	
VTT	3	0	Power output for VTT LDO, need to connect 10-μF or greater MLCC for stability	
VTTREF	6	0	VTTREF buffered reference output. Need to connect 0.22-µF or greater MLCC for stability	
VTTSNS	5	I	VTT LDO voltage sense input	
Thermal Pad		_	Solder to the ground plane for increased thermal performance.	



6 Specifications

6.1 Absolute Maximum Ratings⁽¹⁾

		MIN	MAX	UNIT
	VDD, S3, S5	-0.3	7	V
Input voltage (2)	VLDOIN, VTTSNS, VDDQSNS	-0.3	3.6	
	PGND	-0.3	0.3	V
Output voltage (2)	VTT, VTTREF	-0.3	3.6	
Junction temperature, T _J			125	°C
Operating free-air temperature, T _A		- 55	150	°C
Storage temperature, T _{stg}		- 55	150	°C

⁽¹⁾ Stresses beyond those listed under Absolute Maximum Ratings⁽¹⁾ may cause permanent damage to the device. These are stress ratings only and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

6.2 ESD Ratings

			VALUE	UNIT
		Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 (1)	±2000	
V _(ESD)	Electrostatic discharge	Charged-device model (CDM), per JEDEC specification JESD22-C101 (2)	±500	V

⁽¹⁾ JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process.

6.3 Recommended Operating Conditions

		MIN	NOM MAX	UNIT
Supply voltage	VDD	3.1	6.5	V
	S3, S5	-0.1	6.5	
Input voltage range ⁽¹⁾	VLDOIN, VTTSNS, VDDQSNS	-0.1	3.5	V
	PGND	-0.1	0.1	
Output voltage range ⁽¹⁾	VTT, VTTREF	-0.1	3.5	V
Operating free-air temp	Operating free-air temperature, T _A		85	°C

⁽¹⁾ All voltage values are with respect to the network ground terminal unless otherwise noted.

6.4 Thermal Information

		TPS51206	
	THERMAL METRIC ⁽¹⁾	DSQ (WSON)	UNIT
		10 PINS	
$R_{\theta JA}$	Junction-to-ambient thermal resistance	70.3	
$R_{\theta JC(top)}$	Junction-to-case (top) thermal resistance	46.3	
$R_{\theta JB}$	Junction-to-board thermal resistance	33.8	°C/W
ΨЈТ	Junction-to-top characterization parameter	2.9	C/VV
ΨЈВ	Junction-to-board characterization parameter	33.5	
$R_{\theta JC(bot)}$	Junction-to-case (bottom) thermal resistance	16.3	

For more information about traditional and new thermal metrics, see the Semiconductor and IC Package Thermal Metrics application report (SPRA953).

⁽²⁾ All voltage values are with respect to the network ground terminal unless otherwise noted.

²⁾ JEDEC document JEP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process.



6.5 Electrical Characteristics

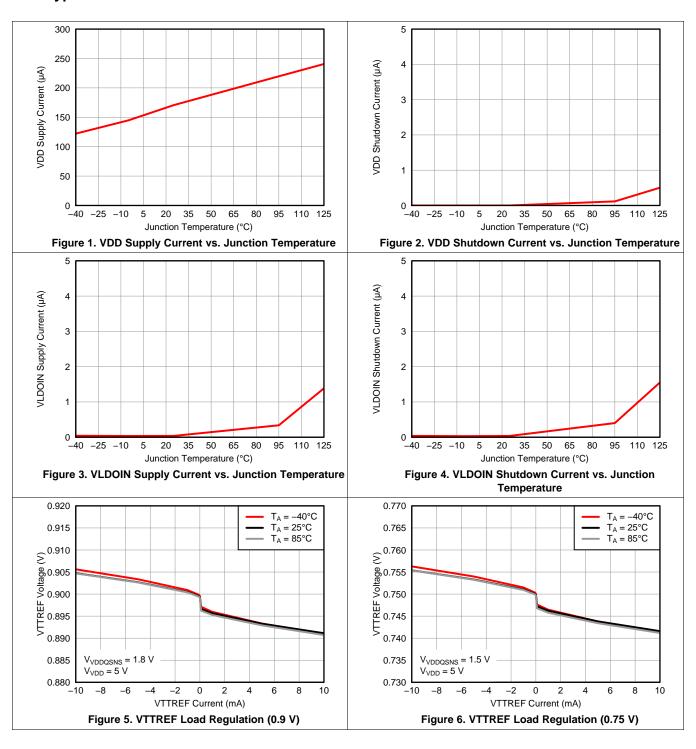
over operating free-air temperature range, $V_{VDD} = 5 \text{ V}$, VLDOIN is connected to VDDQSNS, $V_{S3} = V_{S5} = 5 \text{ V}$ (unless otherwise noted)

	PARAMETER	TEST CONDITION	MIN	TYP	MAX	UNIT
SUPPLY CURR	ENT					
I _{VDD(S0)}	VDD supply current, in S0	T_A = 25°C, No load, V_{S3} = V_{S5} = 5 V, $V_{VDDQSNS}$ = 1.8 V		170		μА
I _{VDD(S3)}	VDD supply current, in S3	$T_A = 25$ °C, No load, $V_{S3} = 0$ V, $V_{S5} = 5$ V, $V_{VDDQSNS} = 1.8$ V		80		μΑ
I _{VDDSDN}	VDD shutdown current, in S4 and S5	$T_A = 25$ °C, No load, $V_{S3} = V_{S5} = 0$ V, $V_{VDDQSNS} = 1.8$ V			1	μА
I _{VLDOIN(S0)}	VLDOIN supply current, in S0	T_A = 25°C, No load, V_{S3} = V_{S5} = 5 V, V_{LDION} = 1.8 V			5	μА
I _{VLDOIN(s3)}	VLDOIN supply current, in S3	T_A = 25°C, No load, V_{S3} = 0 V, V_{S5} = 5 V, V_{LDION} = 1.8 V			5	μΑ
I _{VLDOINSDN}	VLDOIN shutdown current, in S4 and S5	T_A = 25°C, No load, V_{S3} = V_{S5} = 0 V, V_{LDION} = 1.8 V			5	μΑ
VTTREF OUTPL	UT					•
V _{VTTREF}	Output voltage		V	/DDQSNS/2		V
V	Output valtage tel t- V	$ I_{VTTREF} $ < 10 mA, 1.5 V \leq V _{VDDQSNS} \leq 1.8 V	49%	50%	51%	
V _{VTTREFTOL}	Output voltage tolerance to V _{VDDQSNS}	I _{VTTREF} < 10 mA, 1.2 V ≤ V _{VDDQSNS} < 1.5 V	48.75%		51.25%	
I _{VTTREFSRC}	Source current	V _{VDDQSNS} = 1.8 V, V _{VTTREF} = 0 V	10			mA
I _{VTTREFSNK}	Sink current	$V_{VDDQSNS} = 0 \text{ V}, V_{VTTREF} = 1.8 \text{ V}$	10			mA
I _{VTTREFDIS}	VTTREF Discharge current	$T_A = 25$ °C, $V_{S3} = V_{S5} = 0$ V, $V_{VTTREF} = 0.5$ V		1.3		mA
VTT OUTPUT	5					
V _{VTT}	Output voltage		V	/DDQSNS/2		V
		I _{VTT} ≤ 10 mA, 1.4 V ≤ V _{VDDQSNS} ≤ 1.8 V	-20	2240110	20	
		$ I_{VTT} < 1 \text{ A, } 1.4 \text{ V} \le V_{VDDQSNS} \le 1.8 \text{ V}^{(1)}$	-30		30	
	Output voltage tolerance to $V_{\text{VDDQSNS}}/2$	$ I_{VTT} < 2 \text{ A, } 1.4 \text{ V} \le V_{VDDQSNS} \le 1.8 \text{ V}^{(1)}$	-40		40	mV
V_{VTTTOL}		I _{VTT} ≤ 10 mA, 1.2 V ≤ V _{VDDQSNS} ≤ 1.4 V	-20		20	
		$ I_{VTT} < 1 \text{ A, } 1.2 \text{ V} \le V_{VDDOSNS} \le 1.4 \text{ V}^{(1)}$	-30		30	
		$ I_{VTT} < 1.5 \text{ A}, 1.2 \text{ V} \le V_{VDDQSNS} < 1.4 \text{ V}^{(1)}$	-40		40	
I _{VTTOCLSRC}	Source current limit	$V_{VDDQSNS} = 1.8 \text{ V}, V_{VTT} = V_{VTTSNS} = 0.7 \text{ V}$	2			Α
VTTOCLSNK	Sink current limit	$V_{VDDQSNS} = 1.8 \text{ V}, V_{VTT} = V_{VTTSNS} = 1.1 \text{ V}$	2			A
I _{VTTLK}	Leakage current	$T_A = 25^{\circ}\text{C}$, $V_{S3} = 0$ V, $V_{S5} = 5$ V, $V_{VTT} = V_{VTTREF}$			5	μА
I _{VTTSNSBIAS}	VTTSNS input bias current	$V_{S3} = 5 \text{ V}, V_{S5} = 5 \text{ V}, V_{VTTSNS} = V_{VTTREF}$	-0.1		0.1	μА
I _{VTTSNSLK}	VTTSNS leakage current	$V_{S3} = 0 \text{ V}, V_{S5} = 5 \text{ V}, V_{VTTSNS} = V_{VTTREF}$	-0.1		0.1	μА
I _{VTTDIS}	VTT Discharge current	T _A = 25°C, V _{S3} = V _{S5} = V _{VDDQSNS} = 0 V, V _{VTT} = 0.5 V	-	7		mA
VDDQ INPUT		-				
I _{VDDQSNS}	VDDQSNS input current	V _{VDDQSNS} = 1.8 V		30		μА
UVLO/LOGIC T	<u> </u>					
		Wake up	2.67	2.90	3.00	3.00
V _{VDDUV}	VDD UVLO threshold voltage	Hysteresis		0.2		V
V_{LL}	S3 and S5 low-level voltage				0.5	V
V_{LH}	S3 and S5 high-level voltage		1.8			V
V _{LHYST}	S3 and S5 hysteresis voltage			0.3		V
I _{LHLK}	S3 and S5 input leak current		-1		1	μΑ
OVER-TEMPER	ATURE PROTECTION					
т	Over temperature protection	Shutdown temperature ⁽¹⁾		150		۰۵
T _{OTP}	Over temperature protection	Hysteresis ⁽¹⁾		10		°C

⁽¹⁾ Ensured by design. Not production tested.



6.6 Typical Characteristics

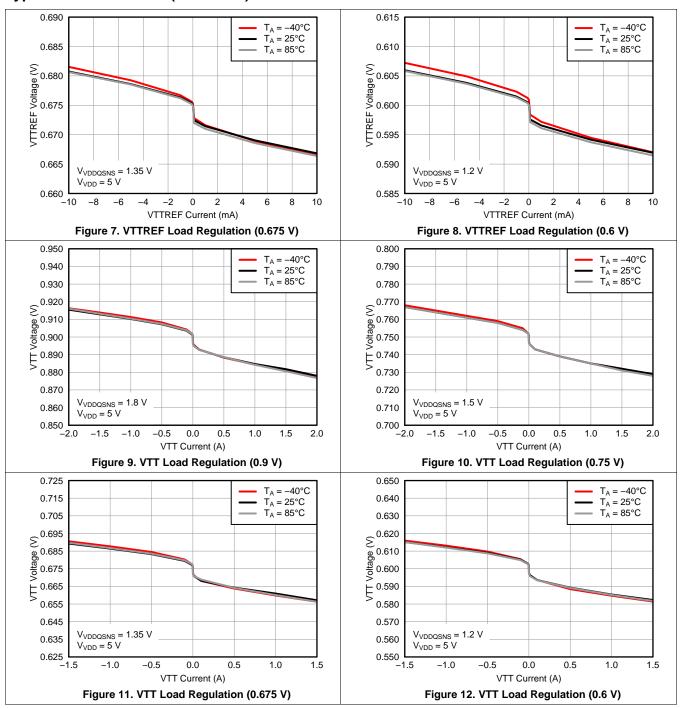


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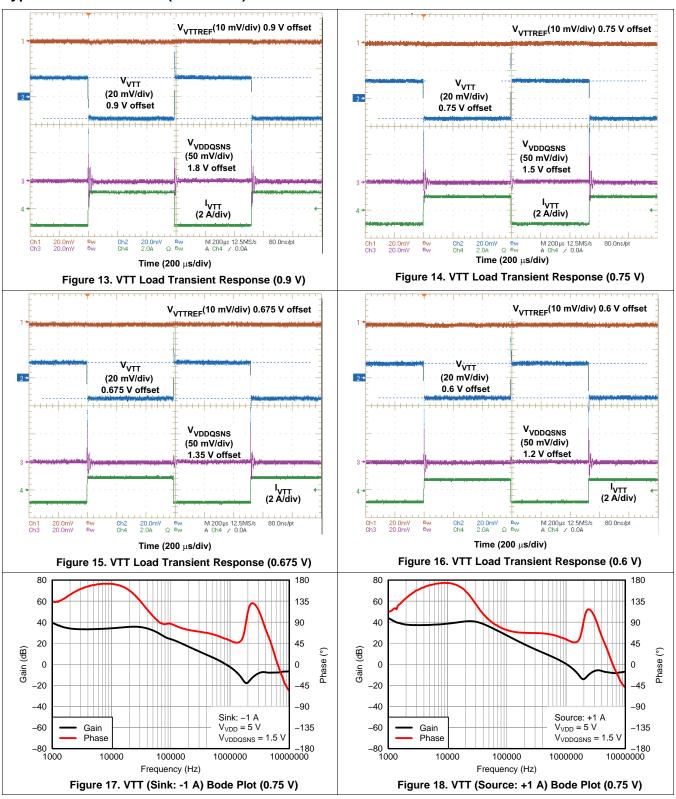


Typical Characteristics (continued)



TEXAS INSTRUMENTS

Typical Characteristics (continued)

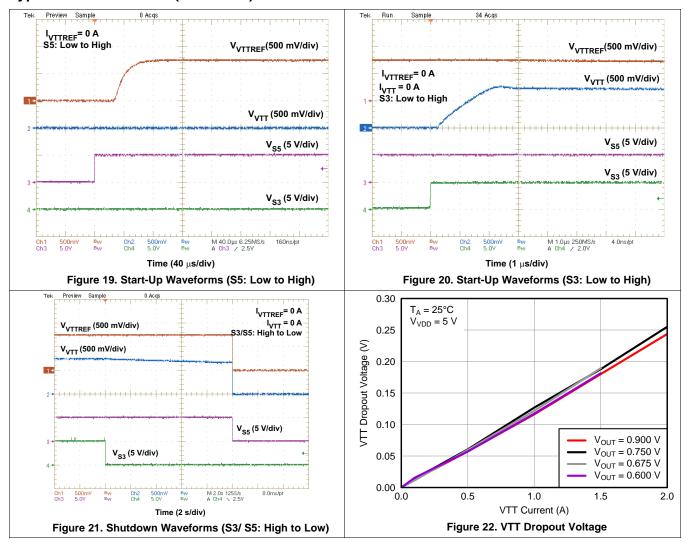


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Typical Characteristics (continued)



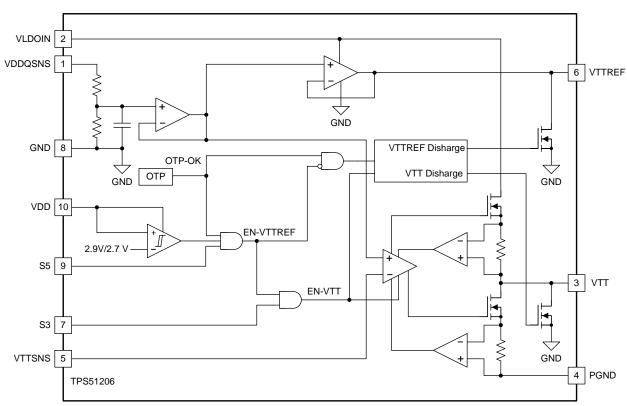


7 Detailed Description

7.1 Overview

The TPS51206 is a sink or source double date rate (DDR) termination regulator with VTTREF buffered reference output.

7.2 Functional Block Diagram



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7.3 Feature Description

7.3.1 VTT Sink and Source Regulator

The TPS51206 is a sink or source tracking termination regulator specifically designed for low input voltage, low cost, and low external component count systems where space is a key application parameter. The TPS51206 integrates a high-performance, low-dropout (LDO) linear regulator (VTT) that has ultimate fast response to track $\frac{1}{2}$ VDDQSNS within 40 mV at all conditions, and its current capability is 2 A for both sink and source directions. A 10- μ F (or greater) ceramic capacitor(s) need to be attached close to the VTT terminal for stable operation. A grade of X5R or better is recommended. To achieve tight regulation with minimum effect of trace resistance, the remote sensing terminal, VTTSNS, should be connected to the positive terminal of the output capacitor(s) as a separate trace from the high current path from the VTT pin.

The device has a dedicated pin, VLDOIN, for VTT power supply to minimize the LDO power dissipation on user application. The minimum VLDOIN voltage is 0.4 V above the ½ VDDQSNS voltage.

7.3.2 VTTREF

The VTTREF pin includes 10 mA of sink or source current capability, and tracks ½ of VDDQSNS with ±1% accuracy. A 0.22-µF ceramic capacitor needs to be attached close to the VTTREF terminal for stable operation; X5R or better grade is recommended.



Feature Description (continued)

7.3.3 VDD Undervoltage Lockout Protection

The TPS51206 input voltage (VDD) includes undervoltage lockout protection (UVLO). When the VDD pin voltage is lower than UVLO threshold voltage, VTT and VTTREF are shut off. This is non-latch protection.

7.3.4 Overtemperature Protection

This device features internal temperature monitoring. If the temperature exceeds the threshold value, VTT and VTTREF are shut off. This is a non-latch protection.

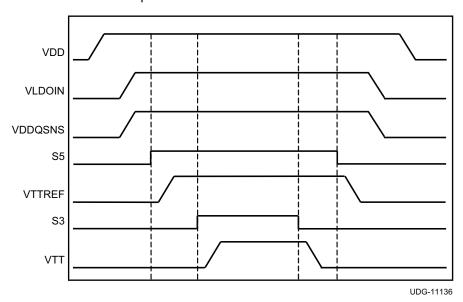


Figure 23. Typical Timing Diagram

7.4 Device Functional Modes

7.4.1 Power State Control

The TPS51206 has two input pins, S3 and S5, to provide simple control of the power state. Table 1 describes S3 and S5 terminal logic state and corresponding state of VTTREF and VTT outputs. VTT is turn-off and placed to high impedance (High-Z) state in S3. The VTT output is floated and does not sink or source current in this state. When both S5 and S3 pins are LOW, the power state is set to S4 and S5. In S4 and S5 state, all the outputs are turn-off and discharged to GND.

Table 1. S3 and S5 Control Table

STATE	S3	S5	VTTREF	VTT
S0	HI	HI	ON	ON
S3	LO	HI	ON	OFF(High-Z)
S4 and S5	LO	LO	OFF(Discharge)	OFF(Discharge)



8 Application and Implementation

NOTE

Information in the following applications sections is not part of the TI component specification, and TI does not warrant its accuracy or completeness. TI's customers are responsible for determining suitability of components for their purposes. Customers should validate and test their design implementation to confirm system functionality.

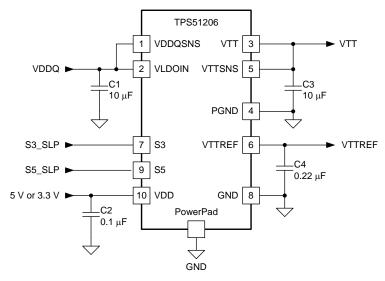
8.1 Application Information

The TPS51206 is typically used as a sink and source tracking termination regulator which converts a voltage from VTT+0.4 V to 3.5 V

8.2 Typical Applications

8.2.1 VLDOIN = VDDQ Configuration

Figure 24 shows an application diagram for a configuration where VLDOIN and VDDQ are connected.



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Figure 24. VLDOIN = VDDQ Configuration

8.2.1.1 Design Requirements

Table 2. Design Parameters

PARAMETER	EXAMPLE VALUE
Supply Voltage (VDD)	3.3 V or 5 V
VLDOIN = VDDQ	1.5 V
Output Current	±2 A



8.2.1.2 Detailed Design Procedure

Table 3. VLDOIN = VDDQ Configuration Components

REFERENCE DESIGNATOR	SPECIFICATION	MANUFACTURER	PART NUMBER
C1, C3	10 μF, 6.3 V, X5R, 1608 (0603)	Taiyo Yuden	JMK107BJ106MA
C2	0.1 μF, 6.3 V, X5R, 1005 (0402)	Taiyo Yuden	JWK105BJ104MP
C4	0.22 μF, 6.3 V, X5R, 1005 (0402)	Taiyo Yuden	JMK105BJ224KV

8.2.1.2.1 VDD Capacitor

Add a ceramic capacitor, with a value 0.1 µF (or greater) and X5R grade (or better), placed close to the VDD terminal, to stabilize the bias supply voltage from any parasitic impedance from the power supply rail.

8.2.1.2.2 VLDOIN Capacitor

Depending on the trace impedance between the VLDOIN bulk power supply to the device, a transient increase of source current is supplied mostly by the charge from the VLDOIN input capacitor. Use a 10-µF (or greater) and X5R grade (or better) ceramic capacitor to supply this transient charge.

8.2.1.2.3 VTTREF Capacitor

Add a ceramic capacitor, with a value $0.22~\mu F$ and X5R grade (or better), placed close to the VTTREF terminal for stable operation.

8.2.1.2.4 VTT Capacitor

For stable operation, a $10-\mu\text{F}$ (or greater) and X5R (or better) grade ceramic capacitor(s) need to be attached close to the VTT terminal. This capacitor is recommended to minimize any additional equivalent series resistance (ESR) and/or equivalent series inductance (ESL) of ground trace between the PGND terminal and the VTT capacitor(s).

8.2.1.2.5 VTTSNS Connection

To achieve tight regulation with minimum effect of trace resistance, a remote sensing terminal, the VTTSNS pin should be connected to the positive terminal of the VTT pin output capacitor(s) as a separate trace from the high-current path from VTT. Consider adding a low-pass R-C filter at the VTTSNS pin in case the ESR of the VTT output capacitor(s) is larger than 2 m Ω . The R-C filter time constant should be approximately the same or slightly lower than the time constant of the VTT output capacitance and ESR.

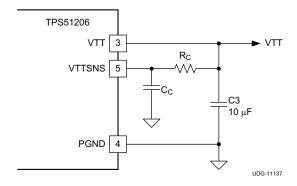


Figure 25. R-C Filter for VTTSNS

8.2.1.2.6 VDDQSNS Connection

VDDQSNS is a reference input of the VTTREF and VTT. Trace should be routed away from noise-generating lines.

8.2.1.3 Application Curves

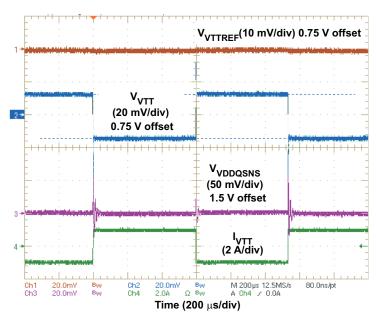


Figure 26. VTT Load Transient Response (0.75 V)

8.2.2 VLDOIN Separated from VDDQ Configuration

Figure 27 shows an application diagram for a configuration where VLDOIN and VDDQ are separated.

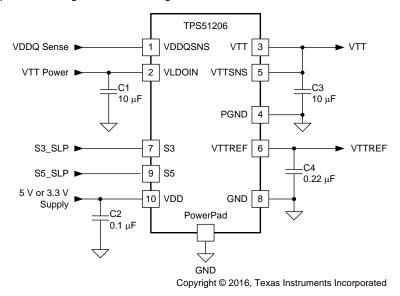


Figure 27. VLDOIN Separated from VDDQ Configuration



8.2.2.1 Design Requirements

Table 4. Design Parameters

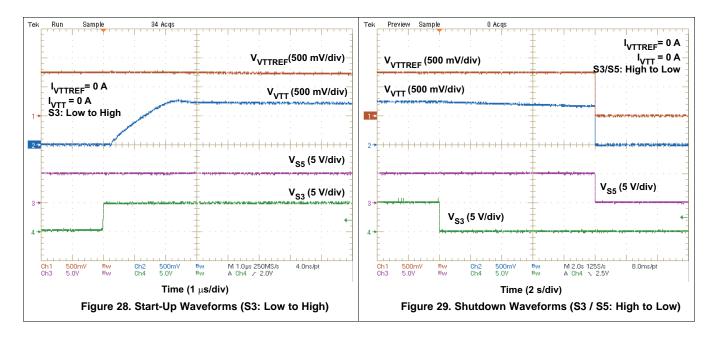
PARAMETER	EXAMPLE VALUE
Supply Voltage (VDD)	3.3 V or 5 V
VLDOIN = VDDQ	1.5 V
Output Current	±2 A

8.2.2.2 Detailed Design Procedure

Table 5. VLDOIN Separated from VDDQ Configuration Components

REFERENCE DESIGNATOR	SPECIFICATION	MANUFACTURER	PART NUMBER
C1, C3	10 μF, 6.3V, X5R, 1608 (0603)	Taiyo Yuden	JMK107BJ106MA
C2	0.1 μF, 6.3V, X5R, 1005 (0402)	Taiyo Yuden	JWK105BJ104MP
C3	10 μF, 6.3V, X5R, 1608 (0603)	Taiyo Yuden	JMK107BJ106MA
C4	0.22 μF, 6.3V, X5R, 1005 (0402)	Taiyo Yuden	JMK105BJ224KV

8.2.2.3 Application Curves





9 Power Supply Recommendations

TPS51206 is designed for a sink / source double date rate (DDR) termination regulator with VTTREF buffered reference output. Supply input voltage (VDD) supports 3.3-V rail and 5-V rail; VLDOIN input voltage supports VTT+0.4 V to 3.5 V.

10 Layout

10.1 Layout Guidelines

Consider the following before beginning a TPS51206 layout design.

- The input bypass capacitor for VLDOIN should be placed as close as possible to the terminal with short and wide connections.
- The output capacitor for VTT should be placed close to the terminals (VTT and PGND) with short and wide connection in order to avoid additional ESR and/or ESL trace inductance.
- VTTSNS should be connected to the positive node of VTT output capacitor(s) as a separate trace from the
 high current VTT power trace. In addition, VTTSNS trace should be routed away from high current trace, on
 the separate layer is recommended. This configuration is strongly recommended to avoid additional ESR
 and/or ESL. If sensing the voltage at the point of the load is required, it is recommended to attach the output
 capacitor(s) at that point. In addition, it is recommended to minimize any additional ESR and/or ESL of ground
 trace between the GND pin and the VTT capacitor(s).
- The GND pin (and the negative node of the VTTREF output capacitor) and PGND pins (and the negative node of the VTT output capacitor) should be connected to the internal system ground planes (for better result, use at least two internal ground planes) with multiple vias. Use as many vias as possible to reduce the impedance between GND pin or PGND pin and the system ground plane.
- In order to effectively remove heat from the package, properly prepare the thermal land. Apply solder directly
 to the package thermal pad. The wide traces of the component and the side copper connected to the thermal
 land pad help to dissipate heat. Numerous vias 0.33 mm in diameter connected from the thermal land to the
 internal/solder side ground plane(s) should also be used to help dissipation. Consult the TPS51206-EVM
 User's Guide for more detailed layout recommendations.



10.2 Layout Example

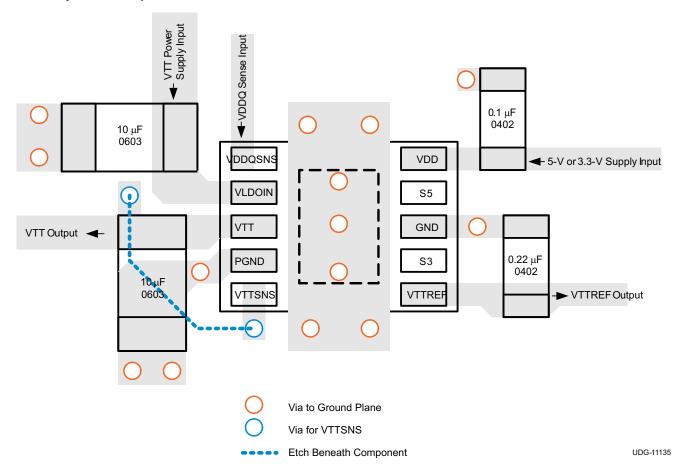


Figure 30. PCB Layout Guideline

10.3 Thermal Considerations

Because the TPS51206 is a linear regulator, the VTT current flows in both source and sink directions, thereby dissipating power from the device. When the device is sourcing current, the voltage difference between V_{VLDOIN} and V_{VTT} times I_{VTT} (VTT current) current becomes the power dissipation as shown in Equation 1.

$$P_{DISS(src)} = (V_{VLDOIN} - V_{VTT}) \times I_{VTT(src)}$$
(1)

In this case, if the VLDOIN pin is connected to an alternative power supply lower than the VDDQ voltage, overall power loss can be reduced. For the sink phase, VTT voltage is applied across the internal LDO regulator, and the power dissipation can be calculated by Equation 2.

$$P_{DISS(snk)} = V_{VTT} \times I_{VTT(snk)}$$
(2)

Maximum power dissipation allowed by the package is calculated by Equation 3.

$$P_{PKG} = \frac{T_{J(max)} - T_{A(max)}}{\theta_{JA}}$$

where

- T_{J(max)} is 125°C
- T_{A(max)} is the maximum ambient temperature in the system
- θ_{JA} is the thermal resistance from junction to ambient

(3)



11 Device and Documentation Support

11.1 Device Support

11.1.1 Third-Party Products Disclaimer

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11.2 Receiving Notification of Documentation Updates

To receive notification of documentation updates, navigate to the device product folder on ti.com. In the upper right corner, click on *Alert me* to register and receive a weekly digest of any product information that has changed. For change details, review the revision history included in any revised document.

11.3 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

TI E2E™ Online Community TI's Engineer-to-Engineer (E2E) Community. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas and help solve problems with fellow engineers.

Design Support *TI's Design Support* Quickly find helpful E2E forums along with design support tools and contact information for technical support.

11.4 Trademarks

PowerPAD, E2E are trademarks of Texas Instruments.
All other trademarks are the property of their respective owners.

11.5 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

11.6 Glossary

SLYZ022 — TI Glossary.

This glossary lists and explains terms, acronyms, and definitions.

12 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.



PACKAGE OPTION ADDENDUM

27-Jun-2016

PACKAGING INFORMATION

Orderable Device	Status	Package Type	_	Pins	_	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
	(1)		Drawing		Qty	(2)	(6)	(3)		(4/5)	
TPS51206DSQR	ACTIVE	WSON	DSQ	10	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	1206	Samples
TPS51206DSQT	ACTIVE	WSON	DSQ	10	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	1206	Samples

(1) The marketing status values are defined as follows:

ACTIVE: Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

PREVIEW: Device has been announced but is not in production. Samples may or may not be available.

OBSOLETE: TI has discontinued the production of the device.

(2) Eco Plan - The planned eco-friendly classification: Pb-Free (RoHS), Pb-Free (RoHS Exempt), or Green (RoHS & no Sb/Br) - please check http://www.ti.com/productcontent for the latest availability information and additional product content details.

TBD: The Pb-Free/Green conversion plan has not been defined.

Pb-Free (RoHS): TI's terms "Lead-Free" or "Pb-Free" mean semiconductor products that are compatible with the current RoHS requirements for all 6 substances, including the requirement that lead not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, TI Pb-Free products are suitable for use in specified lead-free processes.

Pb-Free (RoHS Exempt): This component has a RoHS exemption for either 1) lead-based flip-chip solder bumps used between the die and package, or 2) lead-based die adhesive used between the die and leadframe. The component is otherwise considered Pb-Free (RoHS compatible) as defined above.

Green (RoHS & no Sb/Br): TI defines "Green" to mean Pb-Free (RoHS compatible), and free of Bromine (Br) and Antimony (Sb) based flame retardants (Br or Sb do not exceed 0.1% by weight in homogeneous material)

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead/Ball Finish Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

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PACKAGE OPTION ADDENDUM

27-Jun-2016

In no event shall TI's liability arising out of suc	ch information exceed the total purchase price	of the TI part(s) at issue in this document sole	d by TI to Customer on an annual basis.

PACKAGE MATERIALS INFORMATION

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TAPE AND REEL INFORMATION





	Dimension designed to accommodate the component width
	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



*All dimensions are nominal

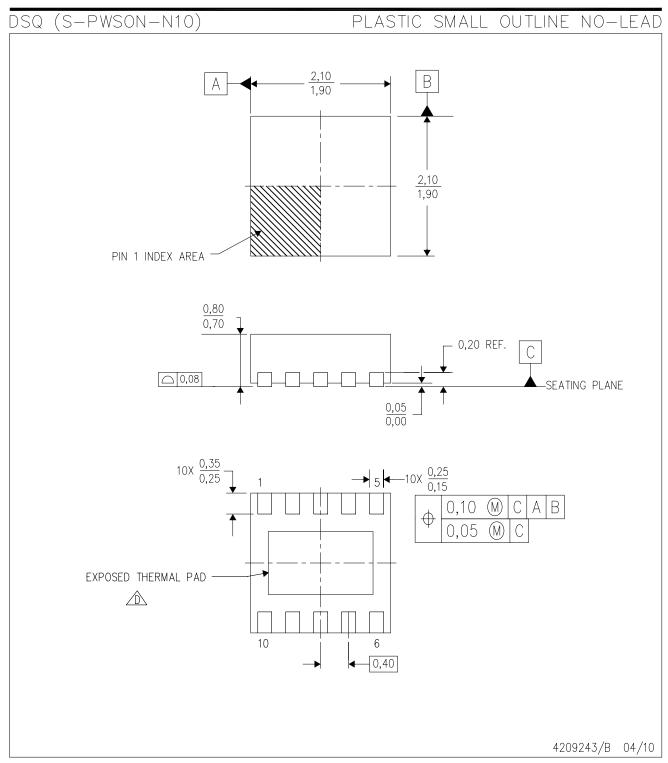
All differsions are nominal												
Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
TPS51206DSQR	WSON	DSQ	10	3000	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2
TPS51206DSQR	WSON	DSQ	10	3000	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2
TPS51206DSQT	WSON	DSQ	10	250	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2
TPS51206DSQT	WSON	DSQ	10	250	180.0	8.4	2.3	2.3	1.15	4.0	8.0	Q2

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*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
TPS51206DSQR	WSON	DSQ	10	3000	210.0	185.0	35.0
TPS51206DSQR	WSON	DSQ	10	3000	210.0	185.0	35.0
TPS51206DSQT	WSON	DSQ	10	250	210.0	185.0	35.0
TPS51206DSQT	WSON	DSQ	10	250	210.0	185.0	35.0



NOTES: A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M-1994.

- B. This drawing is subject to change without notice.
- C. Small Outline No-Lead (SON) package configuration.
- The package thermal pad must be soldered to the board for thermal and mechanical performance. See the Product Data Sheet for details regarding the exposed thermal pad dimensions.



DSQ (R-PWSON-N10)

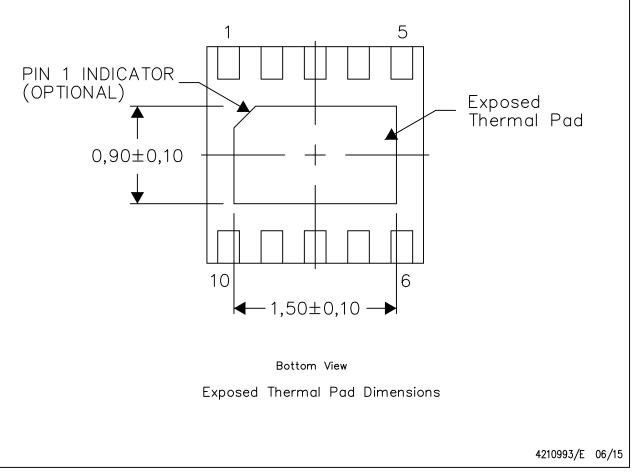
PLASTIC SMALL OUTLINE NO-LEAD

THERMAL INFORMATION

This package incorporates an exposed thermal pad that is designed to be attached directly to an external heatsink. The thermal pad must be soldered directly to the printed circuit board (PCB). After soldering, the PCB can be used as a heatsink. In addition, through the use of thermal vias, the thermal pad can be attached directly to the appropriate copper plane shown in the electrical schematic for the device, or alternatively, can be attached to a special heatsink structure designed into the PCB. This design optimizes the heat transfer from the integrated circuit (IC).

For information on the Quad Flatpack No—Lead (QFN) package and its advantages, refer to Application Report, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271. This document is available at www.ti.com.

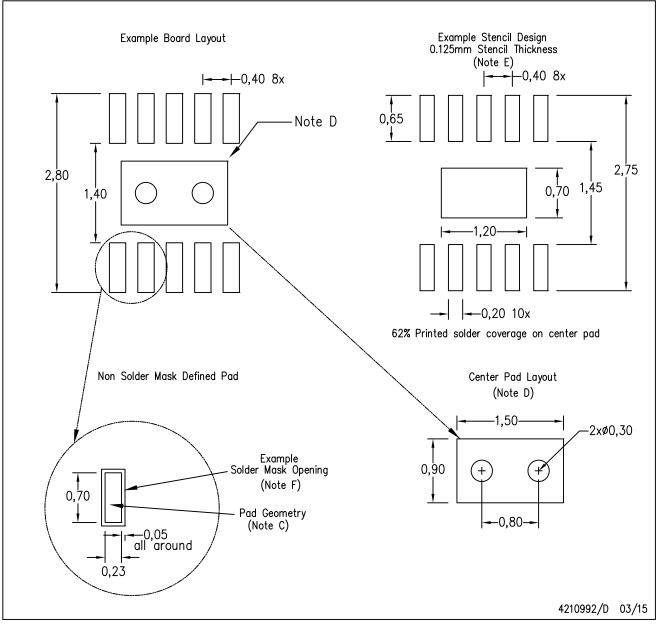
The exposed thermal pad dimensions for this package are shown in the following illustration.



NOTES: A. All linear dimensions are in millimeters

DSQ (R-PWSON-N10)

PLASTIC SMALL OUTLINE NO-LEAD



NOTES: A. All linear dimensions are in millimeters.

- B. This drawing is subject to change without notice.
- C. Publication IPC-7351 is recommended for alternate designs.
- D. This package is designed to be soldered to a thermal pad on the board. Refer to Application Note, QFN/SON PCB Attachment, Texas Instruments Literature No. SLUA271, and also the Product Data Sheets for specific thermal information, via requirements, and recommended board layout. These documents are available at www.ti.com www.ti.com.
- E. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC 7525 for stencil design considerations.
- F. Customers should contact their board fabrication site for solder mask tolerances.



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