













#### MSP430G2533, MSP430G2433, MSP430G2333, MSP430G2233 MSP430G2403, MSP430G2303, MSP430G2203

SLAS734G -APRIL 2011-REVISED APRIL 2016

# MSP430G2x33, MSP430G2x03 Mixed-Signal Microcontrollers

#### **Device Overview**

#### 1.1 **Features**

Low Supply-Voltage Range: 1.8 V to 3.6 V

Ultra-Low Power Consumption

Active Mode: 230 µA at 1 MHz, 2.2 V

Standby Mode: 0.5 μA

Off Mode (RAM Retention): 0.1 µA

Five Power-Saving Modes

Ultra-Fast Wake up From Standby Mode in Less Than 1 µs

16-Bit RISC Architecture, 62.5-ns Instruction Cycle Time

**Basic Clock Module Configurations** 

 Internal Frequencies up to 16 MHz With Four Calibrated Frequencies

 Internal Very-Low-Power Low-Frequency (LF) Oscillator

32-kHz Crvstal

- External Digital Clock Source

Two 16-Bit Timer A With Three Capture/Compare Registers

Up to 24 Capacitive-Touch Enabled I/O Pins

- Universal Serial Communication Interface (USCI)
  - Enhanced UART Supports Automatic Baud-Rate Detection (LIN)
  - IrDA Encoder and Decoder
  - Synchronous SPI
  - I<sup>2</sup>C
- 10-Bit 200-ksps Analog-to-Digital Converter (ADC) With Internal Reference, Sample-and-Hold, and Autoscan (See Table 3-1)
- Brownout Detector
- Serial Onboard Programming, No External Programming Voltage Needed, Programmable Code Protection by Security Fuse
- On-Chip Emulation Logic With Spy-Bi-Wire Interface
- Section 3 Summarizes Available Family Members
- Package Options

- TSSOP: 20 Pin, 28 Pin

- PDIP: 20 Pin QFN: 32 Pin

For Complete Module Descriptions, See the MSP430x2xx Family User's Guide (SLAU144)

#### **Applications** 1.2

- **Power Management**
- Sensor Interface

Capacitive Touch

#### 1.3 **Description**

The TI MSP family of ultra-low-power microcontrollers consists of several devices that feature different sets of peripherals targeted for various applications. The architecture, combined with five low-power modes, is optimized to achieve extended battery life in portable measurement applications. The device features a powerful 16-bit RISC CPU, 16-bit registers, and constant generators that contribute to maximum code efficiency. The digitally controlled oscillator (DCO) allows the device to wake up from lowpower modes to active mode in less than 1 µs.

The MSP430G2x03 and MSP430G2x33 devices are ultra-low-power mixed-signal microcontrollers with built-in 16-bit timers, up to 24 I/O capacitive-touch enabled pins, and built-in communication capability using the USCI. In addition, the MSP430G2x33 family members have a 10-bit ADC. See Section 3 for configuration details.

Typical applications include low-cost sensor systems that capture analog signals, convert them to digital values, and then process the data for display or for transmission to a host system.



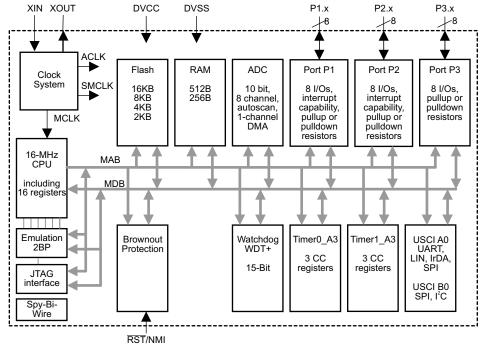
## Device Information<sup>(1)</sup>

PART NUMBER	PACKAGE	BODY SIZE (2)
MSP430G2533IRHB	VQFN (32)	5 mm × 5 mm
MOD 400 COCOOLDIA	TSSOP (28)	9.7 mm × 4.4 mm
MSP430G2533IPW	TSSOP (20)	6.5 mm × 4.4 mm
MSP430G2533IN	PDIP (20)	24.33 mm × 6.35 mm

- (1) For the most current part, package, and ordering information, see the *Package Option Addendum* in Section 8, or see the TI website at www.ti.com.
- (2) The sizes shown here are approximations. For the package dimensions with tolerances, see the *Mechanical Data* in Section 8.

#### 1.4 Functional Block Diagrams

Figure 1-1 shows the functional block diagram of the MSP430G2x33 MCUs.



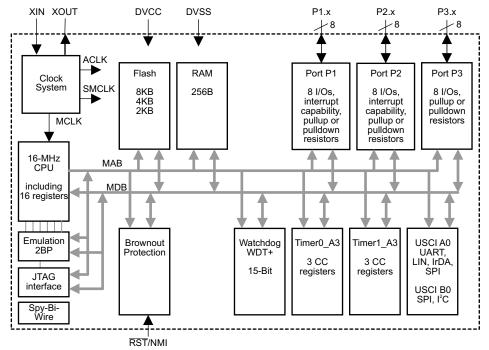
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NOTE: Port P3 is available on 28-pin and 32-pin devices only.

Figure 1-1. Functional Block Diagram, MSP430G2x33



Figure 1-2 shows the functional block diagram of the MSP430G2x03 MCUs.



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NOTE: Port P3 is available on 28-pin and 32-pin devices only.

Figure 1-2. Functional Block Diagram, MSP430G2x03



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# 2 Revision History

NOTE: Page numbers for previous revisions may differ from page numbers in the current version.

Chang	ges from May 2, 2013 to April 27, 2016	Page
•	Document format and organization changes throughout, including addition of section numbering.  Added Device Information table  Added Section 3.1, Related Products  Moved Section 5, Specifications  Added Section 5.2, ESD Ratings.  Added Section 5.8, Thermal Resistance Characteristics	<u>1</u>
•	Throughout document, changed all instances of "bootstrap loader" to "bootloader"	
•	Moved and renamed Section 6.10, I/O Port Diagrams.  Added notes to UCB0STE and UCA0CLK in Table 6-18	48
•	Added notes to UCB0CLK and UCA0STE in Table 6-19  Added "and PW28" to title of Section 6.10.8	62
•	Added "and PW28" to title of Table 6-23	64



## 3 Device Comparison

Table 3-1 compares the available family members.

Table 3-1. Device Comparison<sup>(1)(2)</sup>

DEVICE	BSL	EEM	FLASH (KB)	RAM (B)	Timer_A	ADC10 CHANNELS	USCI_A0, USCI_B0	СГОСК	1/0	PACKAGE	
									24	32-QFN	
MSP430G2533	1	1	16	512	2x TA3	8	1	LF, DCO,	24	28-TSSOP	
WSP430G2533	1	ı	16	312	2X 1A3	0	I	VLO	16	20-TSSOP	
									16	20-PDIP	
									24	32-QFN	
MSP430G2433	1	1	8	512	2x TA3	8	1	LF, DCO,	24	28-TSSOP	
W3F430G2433	'	ı	0	312	2X 1A3	0	ı	VLO	16	20-TSSOP	
									16	20-PDIP	
									24	32-QFN	
MSP430G2333	1	1	4	256	2x TA3	8	1	LF, DCO,	24	28-TSSOP	
WISF430G2333	'	ı	4	236	2X 1A3	0	ı	VLO	16	20-TSSOP	
									16	20-PDIP	
								LF, DCO,		24	32-QFN
MSP430G2233	1	1	2	256	2x TA3	8	4		24	28-TSSOP	
WISF430G2233	'	ı	2	236	2X 1A3	0	1	VLO	16	20-TSSOP	
									16	20-PDIP	
									24	32-QFN	
MSP430G2403	1	1	8	512	2x TA3		1	LF, DCO,	24	28-TSSOP	
MSP430G2403	1	ı	0	312	2X 1A3	_	I	VLO	16	20-TSSOP	
									16	20-PDIP	
									24	32-QFN	
MSP430G2303	4	1	4	256	2x TA3		4	LF, DCO,	24	28-TSSOP	
WSP430G2303	1	ı	4	256	2X 1A3	_	1	VLO	16	20-TSSOP	
									16	20-PDIP	
									24	32-QFN	
MSP430G2203	1	1	2	256	2x TA3		1	LF, DCO,	24	28-TSSOP	
IVISP430G2203	'	ı		200	2X 1A3	-	1	- 1	VLO	16	20-TSSOP
									16	20-PDIP	

<sup>(1)</sup> For the most current device, package, and ordering information, see the *Package Option Addendum* in Section 8, or see the TI website at www.ti.com.

<sup>(2)</sup> Package drawings, thermal data, and symbolization are available at www.ti.com/packaging.





#### 3.1 Related Products

For information about other devices in this family of products or related products, see the following links.

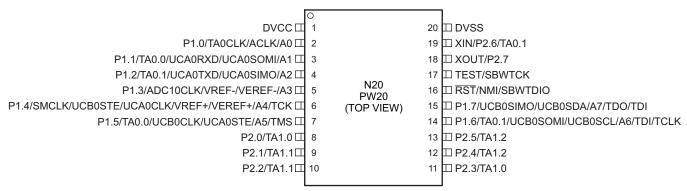
- **Products for MSP 16-Bit and 32-Bit MCUs** Low-power mixed-signal processors with smart analog and digital peripherals for a wide range of industrial and consumer applications.
- Products for Ultra-low Power MCUs MSP Ultra-Low-Power microcontrollers (MCUs) from Texas Instruments (TI) offer the lowest power consumption and the perfect mix of integrated peripherals for a wide range of low-power and portable applications.
- Products for MSP430G2x/i2x Low-Cost Industrial MCUs MSP430G2x microcontrollers (MCUs) from the MSP ultra-low-power MCU series, offers the low power and performance of 16-bit MSP microcontrollers with a feature set targeted at cost sensitive applications.
- Companion Products for MSP430G2533 Review products that are frequently purchased or used in conjunction with this product.
- Reference Designs for MSP430G2533 TI Designs Reference Design Library is a robust reference design library that spans analog, embedded processor, and connectivity. Created by TI experts to help you jump start your system design, all TI Designs include schematic or block diagrams, BOMs, and design files to speed your time to market. Search and download designs at ti.com/tidesigns.



## 4 Terminal Configuration and Functions

#### 4.1 Pin Diagrams

Figure 4-1 shows the pinout for the MSP430G2x03 and MSP430G2x33 devices in the 20-pin N or PW package.

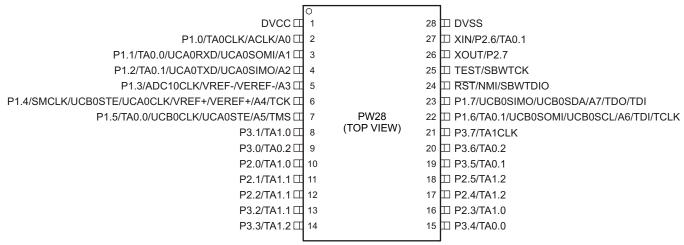


NOTE: ADC10 is available on MSP430G2x33 devices only.

NOTE: The pulldown resistors of port P3 should be enabled by setting P3REN.x = 1.

Figure 4-1. 20-Pin N or PW Package (Top View), MSP430G2x03 and MSP430G2x33

Figure 4-2 shows the pinout for the MSP430G2x03 and MSP430G2x33 devices in the 28-pin PW package.

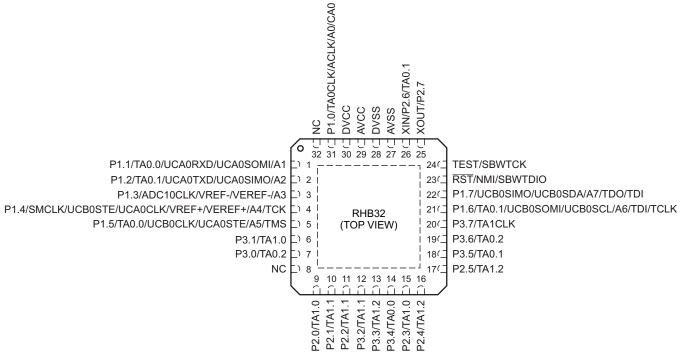


NOTE: ADC10 is available on MSP430G2x33 devices only.

Figure 4-2. 28-Pin PW Package (Top View), MSP430G2x03 and MSP430G2x33



Figure 4-3 shows the pinout for the MSP430G2x03 and MSP430G2x33 devices in the 32-pin RHB package.



NOTE: ADC10 is available on MSP430G2x33 devices only.

Figure 4-3. 32-Pin RHB Package (Top View), MSP430G2x03 and MSP430G2x33



## 4.2 Signal Descriptions

Table 4-1 describes the signals.

**Table 4-1. Terminal Functions** 

	TERMINAL								
		NO.		1/0	DESCRIPTION				
NAME	PW20, N20	PW28	RHB32	1/0	DESCRIPTION				
P1.0/					General-purpose digital I/O pin				
TA0CLK/	2	2	31	I/O	Timer0_A, clock signal TACLK input				
ACLK/		2	31	1/0	ACLK signal output				
A0					ADC10 analog input A0 <sup>(1)</sup>				
P1.1/					General-purpose digital I/O pin				
TA0.0/					Timer0_A, capture: CCI0A input, compare: Out0 output / BSL transmit				
UCA0RXD/	3	3	1	I/O	USCI_A0 receive data input in UART mode				
UCA0SOMI/					USCI_A0 slave data out/master in SPI mode				
A1					ADC10 analog input A1 <sup>(1)</sup> General-purpose digital I/O pin				
P1.2/					General-purpose digital I/O pin				
TA0.1/					Timer0_A, capture: CCI1A input, compare: Out1 output				
UCA0TXD/	4	4	2	I/O	USCI_A0 transmit data output in UART mode				
UCA0SIMO/					USCI_A0 slave data in/master out in SPI mode				
A2					ADC10 analog input A2 <sup>(1)</sup>				
P1.3/					General-purpose digital I/O pin				
ADC10CLK/	5	5 5		I/O	ADC10, conversion clock output <sup>(1)</sup>				
A3/	5	5	3	1/0	ADC10 analog input A3 <sup>(1)</sup>				
VREF-/VEREF-					ADC10 negative reference voltage <sup>(1)</sup>				
P1.4/					General-purpose digital I/O pin				
SMCLK/					SMCLK signal output				
UCB0STE/					USCI_B0 slave transmit enable				
UCA0CLK/	6	6	4	I/O	USCI_A0 clock input/output				
A4/					ADC10 analog input A4 <sup>(1)</sup>				
VREF+/VEREF+					ADC10 positive reference voltage <sup>(1)</sup>				
TCK					JTAG test clock, input terminal for device programming and test				
P1.5/					General-purpose digital I/O pin				
TA0.0/					Timer0_A, compare: Out0 output / BSL receive				
UCB0CLK/	7	7	5	I/O	USCI_B0 clock input/output				
UCA0STE/	,	1	5	1/0	USCI_A0 slave transmit enable				
A5/					ADC10 analog input A5 <sup>(1)</sup>				
TMS					JTAG test mode select, input terminal for device programming and test				
P1.6/					General-purpose digital I/O pin				
TA0.1/					Timer0_A, compare: Out1 output				
A6/	1.1	22	24	1/0	ADC10 analog input A6 <sup>(1)</sup>				
UCB0SOMI/	14	22	21	I/O	USCI_B0 slave out/master in SPI mode,				
UCB0SCL/					USCI_B0 SCL I <sup>2</sup> C clock in I <sup>2</sup> C mode				
TDI/TCLK					JTAG test data input or test clock input during programming and test				



#### **Table 4-1. Terminal Functions (continued)**

	TERMINA	۱L						
		NO.		1/0	DESCRIPTION			
NAME	PW20, N20	PW28	RHB32	I/O	DESCRIPTION			
P1.7/					General-purpose digital I/O pin			
A7/					ADC10 analog input A7 <sup>(1)</sup>			
UCB0SIMO/	15	23	22	I/O	USCI_B0 slave in/master out in SPI mode			
UCB0SDA/	15	23	22	1/0	USCI_B0 SDA I <sup>2</sup> C data in I <sup>2</sup> C mode			
TDO/TDI					JTAG test data output terminal or test data input during programming and test <sup>(2)</sup>			
P2.0/	0	40	0	2	General-purpose digital I/O pin			
TA1.0	8	10	9	I/O	Timer1_A, capture: CCI0A input, compare: Out0 output			
P2.1/	9	11	10	I/O	General-purpose digital I/O pin			
TA1.1	9	11	10	1/0	Timer1_A, capture: CCl1A input, compare: Out1 output			
P2.2/	10	12	11	1/0	General-purpose digital I/O pin			
TA1.1	10	12	11	1/0	Timer1_A, capture: CCI1B input, compare: Out1 output			
P2.3/	11	16	15	I/O	General-purpose digital I/O pin			
TA1.0	11	10	13	1/0	Timer1_A, capture: CCI0B input, compare: Out0 output			
P2.4/	12	17	16	I/O	General-purpose digital I/O pin			
TA1.2	12	17	10	2	Timer1_A, capture: CCI2A input, compare: Out2 output			
P2.5/	13	18	17	I/O	General-purpose digital I/O pin			
TA1.2	13	10	17	2	Timer1_A, capture: CCI2B input, compare: Out2 output			
XIN/					Input terminal of crystal oscillator			
P2.6/	19	27	26	I/O	General-purpose digital I/O pin			
TA0.1					Timer0_A, compare: Out1 output			
XOUT/	18	26	25	I/O	Output terminal of crystal oscillator (3)			
P2.7	10	20	25	1/0	General-purpose digital I/O pin			
P3.0/	_	9	7	I/O	General-purpose digital I/O pin			
TA0.2		9	,	1/0	Timer0_A, capture: CCI2A input, compare: Out2 output			
P3.1/	_	8	6	I/O	General-purpose digital I/O pin			
TA1.0		0	0	1/0	Timer1_A, compare: Out0 output			
P3.2/	_	13	12	I/O	General-purpose digital I/O pin			
TA1.1		13	12	1/0	Timer1_A, compare: Out1 output			
P3.3/	_	14	13	I/O	General-purpose digital I/O			
TA1.2		14	13	1/0	Timer1_A, compare: Out2 output			
P3.4/	_	15	14	I/O	General-purpose digital I/O			
TA0.0		13	14	1/0	Timer0_A, compare: Out0 output			
P3.5/	_	19	18	I/O	General-purpose digital I/O			
TA0.1		19	10	1/0	Timer0_A, compare: Out1 output			
P3.6/	_	20	19	I/O	General-purpose digital I/O			
TA0.2	_	20	13	1/0	Timer0_A, compare: Out2 output			
P3.7/	_	21	20	I/O	General-purpose digital I/O			
TA1CLK	_	۷.	20	1,0	Timer1_A, clock signal TACLK input			
RST/					Reset			
NMI/	16	24	23	- 1	Nonmaskable interrupt input			
SBWTDIO					Spy-Bi-Wire test data input/output during programming and test			

<sup>(2)</sup> TDO or TDI is selected by JTAG instruction.

<sup>(3)</sup> If XOUT/P2.7 is used as an input, excess current flows until P2SEL.7 is cleared. This is due to the oscillator output driver connection to this pad after reset.



## Table 4-1. Terminal Functions (continued)

	TERMINA	L						
		NO.			DESCRIPTION			
NAME	PW20, N20 PW28 RHB32		I/O					
TEST/	17	25	24	ı	Selects test mode for JTAG pins on Port 1. The device protection fuse is connected to TEST.			
SBWTCK					Spy-Bi-Wire test clock input during programming and test			
AVCC	NA	NA	29	NA	Analog supply voltage			
DVCC	1	1	30	NA	Digital supply voltage			
DVSS	20	28	27, 28	NA	Ground reference			
NC	NA	NA	8, 32	NA	Not connected			
QFN Pad	NA	NA	Pad	NA	QFN package pad connection to VSS recommended.			



## 5 Specifications

## 5.1 Absolute Maximum Ratings<sup>(1)</sup>

over operating free-air temperature range (unless otherwise noted)

		MIN	MAX	UNIT	
Voltage applied at V <sub>CC</sub> to V <sub>SS</sub>		-0.3	4.1	V	
Voltage applied to any pin <sup>(2)</sup>		-0.3	$V_{CC} + 0.3$	V	
Diode current at any device pin			±2	mA	
Character T (3)	Unprogrammed device		150	00	
Storage temperature, T <sub>stg</sub> (3)	Programmed device	-55	150	°C	

<sup>(1)</sup> Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions is not implied. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability.

## 5.2 ESD Ratings

			VALUE	UNIT
.,		Human-body model (HBM), per ANSI/ESDA/JEDEC JS-001 <sup>(1)</sup>	±1000	V
V <sub>(ESD)</sub>	Electrostatic discharge	Charged-device model (CDM), per JEDEC specification JESD22-C101 <sup>(2)</sup>	±250	V

<sup>(1)</sup> JEDEC document JEP155 states that 500-V HBM allows safe manufacturing with a standard ESD control process. Pins listed as ±1000 V may actually have higher performance.

## 5.3 Recommended Operating Conditions

Typical values are specified at  $V_{CC} = 3.3 \text{ V}$  and  $T_A = 25^{\circ}\text{C}$  (unless otherwise noted)

			MIN	NOM	MAX	UNIT
V	O made coella ma	During program execution	1.8		3.6	V
V <sub>CC</sub>	Supply voltage	During flash programming or erase	2.2		3.6	<b>v</b>
$V_{SS}$	Supply voltage			0		V
T <sub>A</sub>	Operating free-air temperature	-40		85	°C	
		V <sub>CC</sub> = 1.8 V, Duty cycle = 50% ±10%	DC		6	
f <sub>SYSTEM</sub>	Processor frequency (maximum MCLK frequency using the USART module) (1)(2)	V <sub>CC</sub> = 2.7 V, Duty cycle = 50% ±10%	DC		12	MHz
		V <sub>CC</sub> = 3.3 V, Duty cycle = 50% ±10%	DC		16	

<sup>(1)</sup> The MSP430 CPU is clocked directly with MCLK. Both the high and low phases of MCLK must not exceed the pulse duration of the specified maximum frequency.

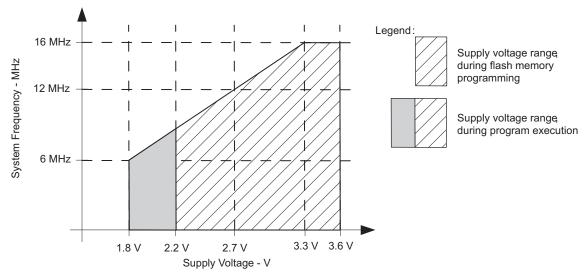
<sup>(2)</sup> All voltages referenced to V<sub>SS</sub>. The JTAG fuse-blow voltage, V<sub>FB</sub>, is allowed to exceed the absolute maximum rating. The voltage is applied to the TEST pin when blowing the JTAG fuse.

<sup>(3)</sup> Higher temperature may be applied during board soldering according to the current JEDEC J-STD-020 specification with peak reflow temperatures not higher than classified on the device label on the shipping boxes or reels.

<sup>(2)</sup> JEDEC document JÉP157 states that 250-V CDM allows safe manufacturing with a standard ESD control process. Pins listed as ±250 V may actually have higher performance.

<sup>(2)</sup> Modules might have a different maximum input clock specification. See the specification of the respective module in this data sheet.





Note: Minimum processor frequency is defined by system clock. Flash program or erase operations require a minimum V<sub>CC</sub> of 2.2 V.

Figure 5-1. Safe Operating Area

## 5.4 Active Mode Supply Current Into V<sub>CC</sub> Excluding External Current

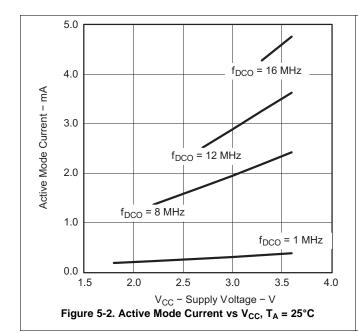
PARAME	TER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
	$f_{DCO} = f_{MCLK} = f_{SMCLK} = 1 \text{ MHz},$	2.2 V		230			
	mode (AM) t at 1 MHz	f <sub>ACLK</sub> = 0 Hz, Program executes in flash, BCSCTL1 = CALBC1_1MHZ, DCOCTL = CALDCO_1MHZ, CPUOFF = 0, SCG0 = 0, SCG1 = 0, OSCOFF = 0	3 V		330	420	μΑ

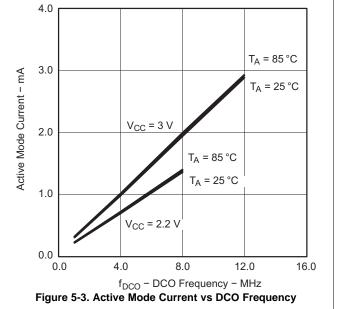
<sup>(1)</sup> All inputs are tied to 0 V or to  $V_{CC}$ . Outputs do not source or sink any current.

<sup>(2)</sup> The currents are characterized with a Micro Crystal CC4V-T1A SMD crystal with a load capacitance of 9 pF. The internal and external load capacitance is chosen to closely match the required 9 pF.



## 5.5 Typical Characteristics, Active Mode Supply Current (Into V<sub>cc</sub>)







## 5.6 Low-Power Mode Supply Currents (Into V<sub>CC</sub>) Excluding External Current

Р	ARAMETER	TEST CONDITIONS	T <sub>A</sub>	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
I <sub>LPM0,1MHz</sub>	Low-power mode 0 (LPM0) current (3)	$\begin{array}{l} f_{MCLK} = 0 \text{ MHz}, \\ f_{SMCLK} = f_{DCO} = 1 \text{ MHz}, \\ f_{ACLK} = 32768 \text{ Hz}, \\ BCSCTL1 = CALBC1\_1MHZ, \\ DCOCTL = CALDCO\_1MHZ, \\ CPUOFF = 1, SCG0 = 0, SCG1 = 0, \\ OSCOFF = 0 \end{array}$	25°C	2.2 V		56		μΑ
I <sub>LPM2</sub>	Low-power mode 2 (LPM2) current (4)	$\begin{split} &f_{\text{MCLK}} = f_{\text{SMCLK}} = 0 \text{ MHz}, \\ &f_{\text{DCO}} = 1 \text{ MHz}, \\ &f_{\text{ACLK}} = 32768 \text{ Hz}, \\ &\text{BCSCTL1} = \text{CALBC1\_1MHZ}, \\ &\text{DCOCTL} = \text{CALDCO\_1MHZ}, \\ &\text{CPUOFF} = 1, \text{SCG0} = 0, \text{SCG1} = 1, \\ &\text{OSCOFF} = 0 \end{split}$	25°C	2.2 V		22		μΑ
I <sub>LPM3,LFXT1</sub>	Low-power mode 3 (LPM3) current <sup>(4)</sup>	$ \begin{aligned} f_{DCO} &= f_{MCLK} = f_{SMCLK} = 0 \text{ MHz}, \\ f_{ACLK} &= 32768 \text{ Hz}, \\ CPUOFF &= 1, SCG0 = 1, SCG1 = 1, \\ OSCOFF &= 0 \end{aligned} $	25°C	2.2 V		0.7	1.5	μΑ
I <sub>LPM3,VLO</sub>	Low-power mode 3 current, (LPM3) <sup>(4)</sup>	$ \begin{aligned} &f_{DCO} = f_{MCLK} = f_{SMCLK} = 0 \text{ MHz}, \\ &f_{ACLK} \text{ from internal LF oscillator (VLO),} \\ &CPUOFF = 1, SCG0 = 1, SCG1 = 1, \\ &OSCOFF = 0 \end{aligned} $	25°C	2.2 V		0.5	0.7	μΑ
		$f_{DCO} = f_{MCLK} = f_{SMCLK} = 0 \text{ MHz},$	25°C			0.1	0.5	
I <sub>LPM4</sub>	Low-power mode 4 (LPM4) current <sup>(5)</sup>	f <sub>ACLK</sub> = 0 Hz, CPUOFF = 1, SCG0 = 1, SCG1 = 1, OSCOFF = 1	85°C	2.2 V		0.8	1.7	μΑ

<sup>(1)</sup> All inputs are tied to 0 V or to  $V_{CC}$ . Outputs do not source or sink any current.

<sup>(2)</sup> The currents are characterized with a Micro Crystal CC4V-T1A SMD crystal with a load capacitance of 9 pF. The internal and external load capacitance is chosen to closely match the required 9 pF.

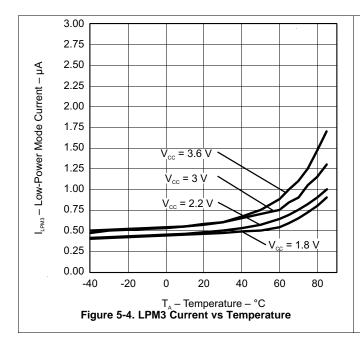
<sup>(3)</sup> Current for brownout and WDT clocked by SMCLK included.

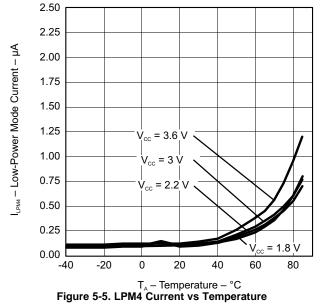
<sup>(4)</sup> Current for brownout and WDT clocked by ACLK included.

<sup>(5)</sup> Current for brownout included.



## 5.7 Typical Characteristics, Low-Power Mode Supply Currents







#### 5.8 Thermal Resistance Characteristics

	PARAMETER		VALUE (1)	UNIT	
		VQFN (RHB-32)	32.1		
DO.	Junction-to-ambient thermal resistance, still air (2)	TSSOP (PW-28)	72.2	°C/W	
$R\theta_{JA}$	Junction-to-ambient thermal resistance, still all V	TSSOP (PW-20)	86.5	*C/vv	
		PDIP (N-20)	49.3		
		VQFN (RHB-32)	22.3		
DO	lunction to some (top) the sunch resistance (3)	TSSOP (PW-28)	18.3	°C/W	
$R\theta_{JC(TOP)}$	Junction-to-case (top) thermal resistance (3)	TSSOP (PW-20)	20.8	3C/VV	
		PDIP (N-20)	41		
		VQFN (RHB-32)	1.4		
D0	handler to see the three the see	TSSOP (PW-28)	N/A	00.004	
$R\theta_{JC(BOTTOM)}$	Junction-to-case (bottom) thermal resistance	TSSOP (PW-20)	N/A	°C/W	
		PDIP (N-20)	N/A		
		VQFN (RHB-32)	6.1		
	Junction-to-board thermal resistance (4)	TSSOP (PW-28)	30.4	90.44	
$\theta_{JB}$	Junction-to-board thermal resistance V	TSSOP (PW-20)	39	°C/W	
		PDIP (N-20)	30.2		
		VQFN (RHB-32)	0.3		
		TSSOP (PW-28)	0.7	90.44	
$\Psi_{ m JT}$	Junction-to-package-top characterization parameter	TSSOP (PW-20)	0.8	°C/W	
		PDIP (N-20)	18.1		
		VQFN (RHB-32)	6.1		
	lunction to be and about station and an	TSSOP (PW-28)	29.9	1	
$\Psi_{JB}$	Junction-to-board characterization parameter	TSSOP (PW-20)	38.1	°C/W	
		PDIP (N-20)	30.1		

<sup>(1)</sup> These values are based on a JEDEC-defined 2S2P system (with the exception of the Theta JC (Rθ<sub>JC</sub>) value, which is based on a JEDEC-defined 1S0P system) and will change based on environment and application. For more information, see these EIA/JEDEC standards:

- JESD51-2, Integrated Circuits Thermal Test Method Environmental Conditions Natural Convection (Still Air)
- JESD51-3, Low Effective Thermal Conductivity Test Board for Leaded Surface Mount Packages
- JESD51-7, High Effective Thermal Conductivity Test Board for Leaded Surface Mount Packages
- JESD51-9, Test Boards for Area Array Surface Mount Package Thermal Measurements
- (2) The junction-to-ambient thermal resistance under natural convection is obtained in a simulation on a JEDEC-standard, High-K board, as specified in JESD51-7, in an environment described in JESD51-2a.
- (3) The junction-to-case (top) thermal resistance is obtained by simulating a cold plate test on the package top. No specific JEDEC-standard test exists, but a close description can be found in the ANSI SEMI standard G30-88.
- (4) The junction-to-board thermal resistance is obtained by simulating in an environment with a ring cold plate fixture to control the PCB temperature, as described in JESD51-8.



#### 5.9 Schmitt-Trigger Inputs, Ports Px

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
V	Desitive gains input threshold valtage			0.45 V <sub>CC</sub>		0.75 V <sub>CC</sub>	V
V <sub>IT+</sub>	Positive-going input threshold voltage		3 V	1.35		2.25	V
.,	No setti o maioni innut thuseheld veltare			0.25 V <sub>CC</sub>		0.55 V <sub>CC</sub>	V
V <sub>IT</sub> _	/ <sub>IT</sub> Negative-going input threshold voltage		3 V	0.75		1.65	V
$V_{hys}$	Input voltage hysteresis (V <sub>IT+</sub> – V <sub>IT-</sub> )		3 V	0.3		1	V
R <sub>Pull</sub>	Pullup or pulldown resistor	For pullup: V <sub>IN</sub> = V <sub>SS</sub> For pulldown: V <sub>IN</sub> = V <sub>CC</sub>	3 V	20	35	50	kΩ
Cı	Input capacitance	$V_{IN} = V_{SS}$ or $V_{CC}$			5		рF

#### 5.10 Leakage Current, Ports Px

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN MAX	UNIT
I <sub>lkg(Px.y)</sub> High-impedance leakage current	See (1) (2)	3 V	±50	nΑ

The leakage current is measured with V<sub>SS</sub> or V<sub>CC</sub> applied to the corresponding pins, unless otherwise noted.

#### 5.11 Outputs, Ports Px

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
$V_{OH}$	High-level output voltage	$I_{(OHmax)} = -6 \text{ mA}^{(1)}$	3 V	V	<sub>CC</sub> – 0.3		V
$V_{OL}$	Low-level output voltage	$I_{(OLmax)} = 6 \text{ mA}^{(1)}$	3 V	V	SS + 0.3		V

The maximum total current, I<sub>(OHmax)</sub> and I<sub>(OLmax)</sub>, for all outputs combined should not exceed ±48 mA to hold the maximum voltage drop specified.

#### 5.12 Output Frequency, Ports Px

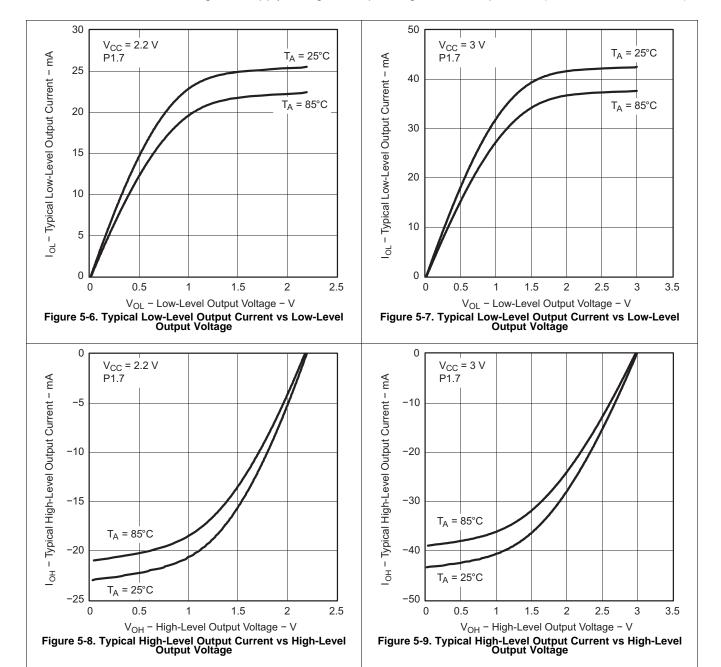
	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
$f_{Px.y}$	Port output frequency (with load)	Px.y, $C_L = 20 \text{ pF}$ , $R_L = 1 \text{ k}\Omega^{(1)}$ (2)	3 V		12		MHz
f <sub>Port_CLK</sub>	Clock output frequency	$Px.y, C_L = 20 pF^{(2)}$	3 V		16		MHz

A resistive divider with two 50-kΩ resistors between V<sub>CC</sub> and V<sub>SS</sub> is used as load. The output is connected to the center tap of the divider.

<sup>(2)</sup> The leakage of the digital port pins is measured individually. The port pin is selected for input and the pullup/pulldown resistor is disabled.

<sup>(2)</sup> The output voltage reaches at least 10% and 90%  $V_{CC}$  at the specified toggle frequency.

#### 5.13 Typical Characteristics – Outputs



V<sub>cc</sub> = 2.2 V



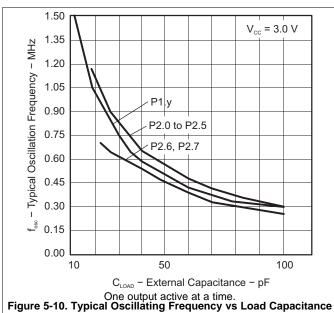
## 5.14 Pin-Oscillator Frequency – Ports Px

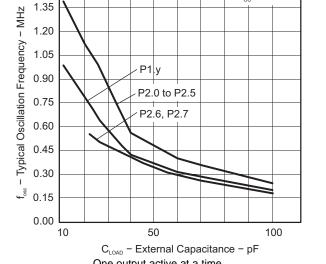
over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT	
40	Port output oscillation frequency	P1.y, $C_L = 10 \text{ pF}$ , $R_L = 100 \text{ k}\Omega^{(1)(2)}$	3 V	1	1400		1-11-	
fo <sub>P1.x</sub>		P1.y, $C_L = 20 \text{ pF}$ , $R_L = 100 \text{ k}\Omega^{(1)(2)}$	3 V		900		kHz	
	Dest autout assillation from the control	P2.0 to P2.5, $C_L = 10 \text{ pF}$ , $R_L = 100 \text{ k}\Omega^{(1)(2)}$	2.1/	1	1800		kHz	
fo <sub>P2.x</sub>	Port output oscillation frequency	P2.0 to P2.5, $C_L = 20 \text{ pF}$ , $R_L = 100 \text{ k}\Omega^{(1)(2)}$	3 V	1	1000		Kr1Z	
fo <sub>P2.6/7</sub>	Port output oscillation frequency	P2.6 and P2.7, $C_L = 20 \text{ pF}$ , $R_L = 100 \text{ k}\Omega^{(1)(2)}$	3 V		700		kHz	
40	Port dutant acciliation traditional	P3.y, $C_L = 10 \text{ pF}$ , $R_L = 100 \text{ k}\Omega^{(1)(2)}$	2.1/	1	1800		1.11=	
fo <sub>P3.x</sub>		P3.y, $C_L = 20 \text{ pF}$ , $R_L = 100 \text{ k}\Omega^{(1)(2)}$	3 V	1	1000		kHz	

<sup>(1)</sup> A resistive divider with two 50-kΩ resistors between V<sub>CC</sub> and V<sub>SS</sub> is used as load. The output is connected to the center tap of the

## 5.15 Typical Characteristics - Pin-Oscillator Frequency





One output active at a time.
Figure 5-11. Typical Oscillating Frequency vs Load Capacitance

The output voltage reaches at least 10% and 90% V<sub>CC</sub> at the specified toggle frequency.



#### 5.16 POR, BOR<sup>(1)(2)</sup>

	PARAMETER		V <sub>CC</sub>	MIN	TYP	MAX	UNIT
V <sub>CC(start)</sub>	See Figure 5-12	dV <sub>CC</sub> /dt ≤ 3 V/s		0.7	V <sub>(B_IT)</sub>		V
V <sub>(B_IT-)</sub>	See Figure 5-12 through Figure 5-14	dV <sub>CC</sub> /dt ≤ 3 V/s		1.35		V	
V <sub>hys(B_IT-)</sub>	See Figure 5-12	dV <sub>CC</sub> /dt ≤ 3 V/s		140			mV
t <sub>d(BOR)</sub>	See Figure 5-12				2000		μs
t <sub>(reset)</sub>	Pulse duration needed at RST/NMI pin to accepted reset internally		2.2 V	2			μs

- The current consumption of the brownout module is already included in the I<sub>CC</sub> current consumption data. The voltage level V<sub>(B\_IT-)</sub> + V<sub>hys(B\_IT-)</sub> is ≤ 1.8 V.
- (2) During power up, the CPU begins code execution following a period of  $t_{d(BOR)}$  after  $V_{CC} = V_{(B\_IT-)} + V_{hys(B\_IT-)}$ . The default DCO settings must not be changed until  $V_{CC} \ge V_{CC(min)}$ , where  $V_{CC(min)}$  is the minimum supply voltage for the desired operating frequency.

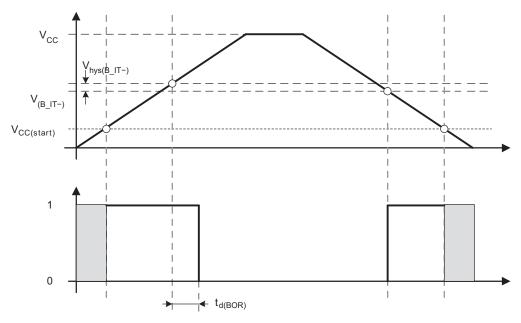


Figure 5-12. POR and BOR vs Supply Voltage

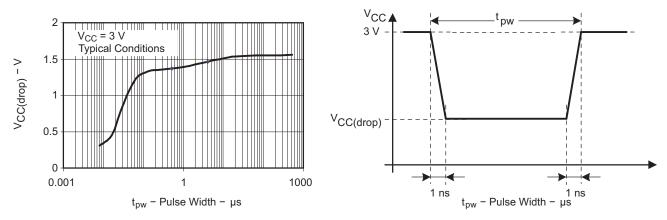


Figure 5-13. V<sub>CC(drop)</sub> Level With a Square Voltage Drop to Generate a POR or BOR Signal



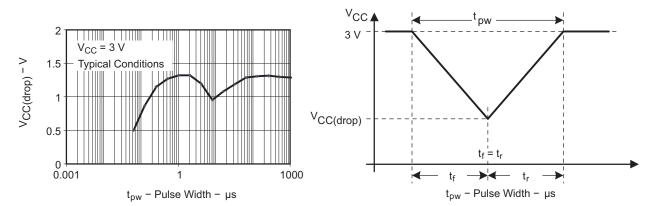


Figure 5-14. V<sub>CC(drop)</sub> Level With a Triangle Voltage Drop to Generate a POR or BOR Signal



#### 5.17 Main DCO Characteristics

- All ranges selected by RSELx overlap with RSELx + 1: RSELx = 0 overlaps RSELx = 1, ... RSELx = 14 overlaps RSELx = 15.
- DCO control bits DCOx have a step size as defined by parameter S<sub>DCO</sub>.
- Modulation control bits MODx select how often f<sub>DCO(RSEL,DCO+1)</sub> is used within the period of 32 DCOCLK cycles. The frequency f<sub>DCO(RSEL,DCO)</sub> is used for the remaining cycles. The frequency is an average equal to:

$$f_{average} = \frac{32 \times f_{DCO(RSEL,DCO)} \times f_{DCO(RSEL,DCO+1)}}{MOD \times f_{DCO(RSEL,DCO)} + (32 - MOD) \times f_{DCO(RSEL,DCO+1)}}$$

#### 5.18 DCO Frequency

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
		RSELx < 14		1.8		3.6	
$V_{CC}$	Supply voltage	RSELx = 14		2.2		3.6	V
		RSELx = 15		3		3.6	
f <sub>DCO(0,0)</sub>	DCO frequency (0, 0)	RSELx = 0, $DCOx = 0$ , $MODx = 0$	3 V	0.06		0.14	MHz
f <sub>DCO(0,3)</sub>	DCO frequency (0, 3)	RSELx = 0, $DCOx = 3$ , $MODx = 0$	3 V	0.07		0.17	MHz
f <sub>DCO(1,3)</sub>	DCO frequency (1, 3)	RSELx = 1, $DCOx = 3$ , $MODx = 0$	3 V		0.15		MHz
f <sub>DCO(2,3)</sub>	DCO frequency (2, 3)	RSELx = 2, $DCOx = 3$ , $MODx = 0$	3 V		0.21		MHz
$f_{DCO(3,3)}$	DCO frequency (3, 3)	RSELx = 3, $DCOx = 3$ , $MODx = 0$	3 V		0.30		MHz
f <sub>DCO(4,3)</sub>	DCO frequency (4, 3)	RSELx = 4, $DCOx = 3$ , $MODx = 0$	3 V		0.41		MHz
f <sub>DCO(5,3)</sub>	DCO frequency (5, 3)	RSELx = 5, $DCOx = 3$ , $MODx = 0$	3 V		0.58		MHz
f <sub>DCO(6,3)</sub>	DCO frequency (6, 3)	RSELx = 6, $DCOx = 3$ , $MODx = 0$	3 V	0.54		1.06	MHz
f <sub>DCO(7,3)</sub>	DCO frequency (7, 3)	RSELx = 7, $DCOx = 3$ , $MODx = 0$	3 V	0.80		1.50	MHz
f <sub>DCO(8,3)</sub>	DCO frequency (8, 3)	RSELx = 8, $DCOx = 3$ , $MODx = 0$	3 V		1.6		MHz
f <sub>DCO(9,3)</sub>	DCO frequency (9, 3)	RSELx = 9, $DCOx = 3$ , $MODx = 0$	3 V		2.3		MHz
f <sub>DCO(10,3)</sub>	DCO frequency (10, 3)	RSELx = 10, $DCOx = 3$ , $MODx = 0$	3 V		3.4		MHz
f <sub>DCO(11,3)</sub>	DCO frequency (11, 3)	RSELx = 11, $DCOx = 3$ , $MODx = 0$	3 V		4.25		MHz
f <sub>DCO(12,3)</sub>	DCO frequency (12, 3)	RSELx = 12, $DCOx = 3$ , $MODx = 0$	3 V	4.30		7.30	MHz
f <sub>DCO(13,3)</sub>	DCO frequency (13, 3)	RSELx = 13, $DCOx = 3$ , $MODx = 0$	3 V	6.00		9.60	MHz
f <sub>DCO(14,3)</sub>	DCO frequency (14, 3)	RSELx = 14, $DCOx = 3$ , $MODx = 0$	3 V	8.60		13.9	MHz
f <sub>DCO(15,3)</sub>	DCO frequency (15, 3)	RSELx = 15, $DCOx = 3$ , $MODx = 0$	3 V	12.0		18.5	MHz
f <sub>DCO(15,7)</sub>	DCO frequency (15, 7)	RSELx = 15, $DCOx = 7$ , $MODx = 0$	3 V	16.0		26.0	MHz
S <sub>RSEL</sub>	Frequency step between range RSEL and RSEL+1	$S_{RSEL} = f_{DCO(RSEL+1,DCO)}/f_{DCO(RSEL,DCO)}$	3 V		1.35		ratio
S <sub>DCO</sub>	Frequency step between tap DCO and DCO+1	$S_{DCO} = f_{DCO(RSEL,DCO+1)}/f_{DCO(RSEL,DCO)}$	3 V		1.08		ratio
Duty cycle		Measured at SMCLK output	3 V		50%		



## 5.19 Calibrated DCO Frequencies, Tolerance

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	T <sub>A</sub>	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
1-MHz tolerance over temperature <sup>(1)</sup>	BCSCTL1 = CALBC1_1MHZ, DCOCTL = CALDCO_1MHZ, calibrated at 30°C and 3 V	0°C to 85°C	3 V	-3%	±0.5%	+3%	
1-MHz tolerance over V <sub>CC</sub>	BCSCTL1 = CALBC1_1MHZ, DCOCTL = CALDCO_1MHZ, calibrated at 30°C and 3 V	30°C	1.8 V to 3.6 V	-3%	±2%	+3%	
1-MHz tolerance overall	BCSCTL1 = CALBC1_1MHZ, DCOCTL = CALDCO_1MHZ, calibrated at 30°C and 3 V	-40°C to 85°C	1.8 V to 3.6 V	-6%	±3%	+6%	
8-MHz tolerance over temperature <sup>(1)</sup>	BCSCTL1 = CALBC1_8MHZ, DCOCTL = CALDCO_8MHZ, calibrated at 30°C and 3 V	0°C to 85°C	3 V	-3%	±0.5%	+3%	
8-MHz tolerance over V <sub>CC</sub>	BCSCTL1 = CALBC1_8MHZ, DCOCTL = CALDCO_8MHZ, calibrated at 30°C and 3 V	30°C	2.2 V to 3.6 V	-3%	±2%	+3%	
8-MHz tolerance overall	BCSCTL1 = CALBC1_8MHZ, DCOCTL = CALDCO_8MHZ, calibrated at 30°C and 3 V	-40°C to 85°C	2.2 V to 3.6 V	-6%	±3%	+6%	
12-MHz tolerance over temperature <sup>(1)</sup>	BCSCTL1 = CALBC1_12MHZ, DCOCTL = CALDCO_12MHZ, calibrated at 30°C and 3 V	0°C to 85°C	3 V	-3%	±0.5%	+3%	
12-MHz tolerance over V <sub>CC</sub>	BCSCTL1 = CALBC1_12MHZ, DCOCTL = CALDCO_12MHZ, calibrated at 30°C and 3 V	30°C	2.7 V to 3.6 V	-3%	±2%	+3%	
12-MHz tolerance overall	BCSCTL1 = CALBC1_12MHZ, DCOCTL = CALDCO_12MHZ, calibrated at 30°C and 3 V	-40°C to 85°C	2.7 V to 3.6 V	-6%	±3%	+6%	
16-MHz tolerance over temperature <sup>(1)</sup>	BCSCTL1 = CALBC1_16MHZ, DCOCTL = CALDCO_16MHZ, calibrated at 30°C and 3 V	0°C to 85°C	3 V	-3%	±0.5%	+3%	
16-MHz tolerance over V <sub>CC</sub>	BCSCTL1 = CALBC1_16MHZ, DCOCTL = CALDCO_16MHZ, calibrated at 30°C and 3 V	30°C	3.3 V to 3.6 V	-3%	±2%	+3%	
16-MHz tolerance overall	BCSCTL1 = CALBC1_16MHZ, DCOCTL = CALDCO_16MHZ, calibrated at 30°C and 3 V	-40°C to 85°C	3.3 V to 3.6 V	-6%	±3%	+6%	

<sup>(1)</sup> This is the frequency change from the measured frequency at 30°C over temperature.

MSP430G2203



#### 5.20 Wake-up Times From Lower-Power Modes (LPM3, LPM4)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
t <sub>DCO,LPM3/4</sub>	DCO clock wake-up time from LPM3 or LPM4 <sup>(1)</sup>	BCSCTL1 = CALBC1_1MHz, DCOCTL = CALDCO_1MHz	3 V		1.5		μs
t <sub>CPU,LPM3/4</sub>	CPU wake-up time from LPM3 or LPM4 (2)			1	1/f <sub>MCLK</sub> +		

<sup>1)</sup> The DCO clock wake-up time is measured from the edge of an external wake-up signal (for example, port interrupt) to the first clock edge observable externally on a clock pin (MCLK or SMCLK).

## 5.21 Typical Characteristics, DCO Clock Wake-up Time From LPM3 or LPM4

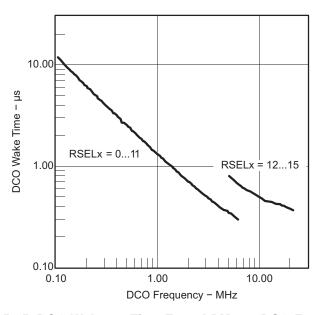


Figure 5-15. DCO Wake-up Time From LPM3 vs DCO Frequency

<sup>(2)</sup> Parameter applicable only if DCOCLK is used for MCLK.



## 5.22 Crystal Oscillator, XT1, Low-Frequency Mode<sup>(1)</sup>

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
f <sub>LFXT1,LF</sub>	LFXT1 oscillator crystal frequency, LF mode 0, 1	XTS = 0, LFXT1Sx = 0 or 1	1.8 V to 3.6 V		32768		Hz
f <sub>LFXT1,LF,logic</sub>	LFXT1 oscillator logic level square-wave input frequency, LF mode	XTS = 0, XCAPx = 0, LFXT1Sx = 3	1.8 V to 3.6 V	10000	32768	50000	Hz
04	Oscillation allowance for	XTS = 0, LFXT1Sx = 0, f <sub>LFXT1,LF</sub> = 32768 Hz, C <sub>L,eff</sub> = 6 pF			500		kΩ
OA <sub>LF</sub>	LF crystals	$XTS = 0$ , $LFXT1Sx = 0$ , $f_{LFXT1,LF} = 32768$ Hz, $C_{L,eff} = 12$ pF			200		K12
		XTS = 0, $XCAPx = 0$			1		
0	Integrated effective load	XTS = 0, $XCAPx = 1$		5.5			
$C_{L,eff}$	capacitance, LF mode <sup>(2)</sup>	XTS = 0, $XCAPx = 2$			8.5		pF
		XTS = 0, $XCAPx = 3$			11		
	Duty cycle, LF mode	XTS = 0, Measured at P2.0/ACLK, f <sub>LFXT1,LF</sub> = 32768 Hz	2.2 V	30%	50%	70%	
f <sub>Fault,LF</sub>	Oscillator fault frequency, LF mode <sup>(3)</sup>	XTS = 0, XCAPx = 0, LFXT1Sx = 3 <sup>(4)</sup>	2.2 V	10		10000	Hz

- (1) To improve EMI on the XT1 oscillator, the following guidelines should be observed.
  - Keep the trace between the device and the crystal as short as possible.
  - Design a good ground plane around the oscillator pins.
  - Prevent crosstalk from other clock or data lines into oscillator pins XIN and XOUT.
  - Avoid running PCB traces underneath or adjacent to the XIN and XOUT pins.
  - Use assembly materials and processes that avoid any parasitic load on the oscillator XIN and XOUT pins.
  - If conformal coating is used, make sure that it does not induce capacitive or resistive leakage between the oscillator pins.
  - Do not route the XOUT line to the JTAG header to support the serial programming adapter as shown in other documentation. This
    signal is no longer required for the serial programming adapter.
- (2) Includes parasitic bond and package capacitance (approximately 2 pF per pin). Because the PCB adds additional capacitance, verify the correct load by measuring the ACLK frequency. For a correct setup, the effective load capacitance should always match the specification of the used crystal.
- (3) Frequencies below the MIN specification set the fault flag. Frequencies above the MAX specification do not set the fault flag. Frequencies in between might set the flag.
- (4) Measured with logic-level input frequency but also applies to operation with crystals.

#### 5.23 Internal Very-Low-Power Low-Frequency Oscillator (VLO)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	T <sub>A</sub>	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
$f_{VLO}$	VLO frequency	–40°C to 85°C	3 V	4	12	20	kHz
$df_{VLO}/d_{T}$	VLO frequency temperature drift	–40°C to 85°C	3 V		0.5		%/°C
$\mathrm{df_{VLO}}/\mathrm{dV_{CC}}$	VLO frequency supply voltage drift	25°C	1.8 V to 3.6 V		4		%/V

### 5.24 Timer A

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
$f_{TA}$	Timer_A input clock frequency	SMCLK, duty cycle = 50% ±10%			f <sub>SYSTEM</sub>		MHz
t <sub>TA,cap</sub>	Timer_A capture timing	TA0, TA1	3 V	20			ns



#### 5.25 USCI (UART Mode)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
fusci	USCI input clock frequency	SMCLK, duty cycle = 50% ±10%			f <sub>SYSTEM</sub>		MHz
f <sub>max,BITCLK</sub>	Maximum BITCLK clock frequency (equals baud rate in MBaud) (1)		3 V	2			MHz
t <sub>T</sub>	UART receive deglitch time <sup>(2)</sup>		3 V	50	100	600	ns

<sup>(1)</sup> The DCO wake-up time must be considered in LPM3 and LPM4 for baud rates above 1 MHz.

#### 5.26 USCI (SPI Master Mode)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted) (see Figure 5-16 and Figure 5-17)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	MAX	UNIT
f <sub>USCI</sub>	USCI input clock frequency	SMCLK, duty cycle = 50% ±10%			f <sub>SYSTEM</sub>	MHz
t <sub>SU,MI</sub>	SOMI input data setup time		3 V	75		ns
t <sub>HD,MI</sub>	SOMI input data hold time		3 V	0		ns
t <sub>VALID,MO</sub>	SIMO output data valid time	UCLK edge to SIMO valid, C <sub>L</sub> = 20 pF	3 V		20	ns

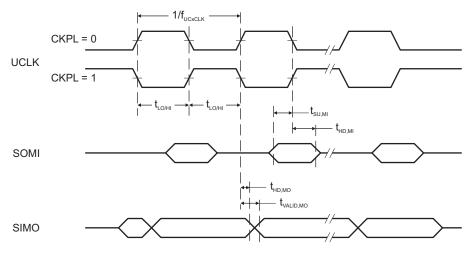


Figure 5-16. SPI Master Mode, CKPH = 0

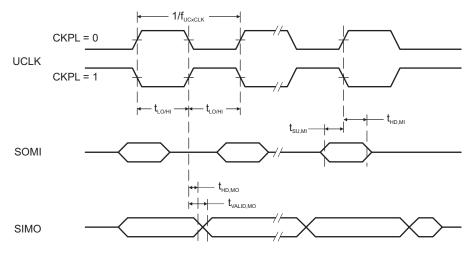


Figure 5-17. SPI Master Mode, CKPH = 1

<sup>(2)</sup> Pulses on the UART receive input (UCxRX) shorter than the UART receive deglitch time are suppressed. To ensure that pulses are correctly recognized, their duration should exceed the maximum specification of the deglitch time.



# 5.27 USCI (SPI Slave Mode)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted) (see Figure 5-18 and Figure 5-19)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
t <sub>STE,LEAD</sub>	STE lead time, STE low to clock		3 V		50		ns
t <sub>STE,LAG</sub>	STE lag time, Last clock to STE high		3 V	10			ns
t <sub>STE,ACC</sub>	STE access time, STE low to SOMI data out		3 V		50		ns
t <sub>STE,DIS</sub>	STE disable time, STE high to SOMI high impedance		3 V		50		ns
t <sub>SU,SI</sub>	SIMO input data setup time		3 V	15			ns
t <sub>HD,SI</sub>	SIMO input data hold time		3 V	10			ns
t <sub>VALID,SO</sub>	SOMI output data valid time	UCLK edge to SOMI valid, C <sub>L</sub> = 20 pF	3 V		50	75	ns

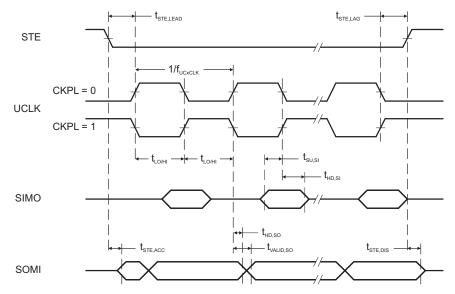


Figure 5-18. SPI Slave Mode, CKPH = 0

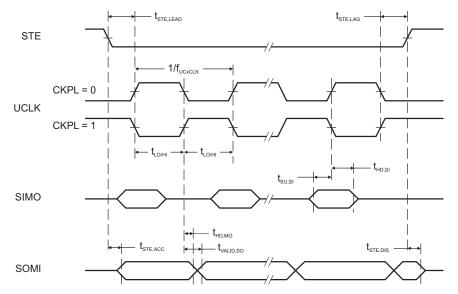


Figure 5-19. SPI Slave Mode, CKPH = 1

# 5.28 USCI (I<sup>2</sup>C Mode)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
f <sub>USCI</sub>	USCI input clock frequency	SMCLK, duty cycle = 50% ±10%				f <sub>SYSTEM</sub>	MHz
f <sub>SCL</sub>	SCL clock frequency		3 V	0		400	kHz
	Hold time (repeated) CTART	f <sub>SCL</sub> ≤ 100 kHz	3 V	4.0			
t <sub>HD,STA</sub>	Hold time (repeated) START	f <sub>SCL</sub> > 100 kHz	3 V	0.6			μs
	Catual time for a repeated START	f <sub>SCL</sub> ≤ 100 kHz	3 V	4.7			
t <sub>SU,STA</sub>	Setup time for a repeated START	f <sub>SCL</sub> > 100 kHz	3 V	0.6			μs
t <sub>HD,DAT</sub>	Data hold time		3 V	0			ns
t <sub>SU,DAT</sub>	Data setup time		3 V	250			ns
t <sub>SU,STO</sub>	Setup time for STOP		3 V	4.0			μs
t <sub>SP</sub>	Pulse duration of spikes suppressed by input filter		3 V	50	100	600	ns

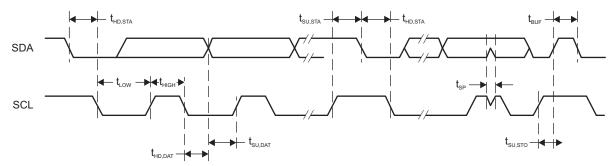


Figure 5-20. I<sup>2</sup>C Mode Timing



## 5.29 10-Bit ADC, Power Supply and Input Range Conditions (MSP430G2x33 Only)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)<sup>(1)</sup>

	PARAMETER	TEST CONDITIONS	T <sub>A</sub>	V <sub>CC</sub>	MIN	TYP	MAX	UNIT
V <sub>CC</sub>	Analog supply voltage	V <sub>SS</sub> = 0 V			2.2		3.6	V
V <sub>Ax</sub>	Analog input voltage (2)	All Ax terminals, Analog inputs selected in ADC10AE register		3 V	0		$V_{CC}$	V
I <sub>ADC10</sub>	ADC10 supply current <sup>(3)</sup>	$ \begin{aligned} &f_{ADC10CLK} = 5.0 \text{ MHz}, \\ &ADC10ON = 1, \text{ REFON} = 0, \\ &ADC10SHT0 = 1, \text{ ADC10SHT1} = 0, \\ &ADC10DIV = 0 \end{aligned} $	25°C	3 V		0.6		mA
	Reference supply current,	f <sub>ADC10CLK</sub> = 5.0 MHz, ADC10ON = 0, REF2_5V = 0, REFON = 1, REFOUT = 0	0500	3 V		0.25		
I <sub>REF+</sub>	Reference supply current, reference buffer disabled (4)	f <sub>ADC10CLK</sub> = 5.0 MHz, ADC10ON = 0, REF2_5V = 1, REFON = 1, REFOUT = 0	25°C	3 V		0.25		mA
I <sub>REFB,0</sub>	Reference buffer supply current with ADC10SR = $0^{(4)}$	f <sub>ADC10CLK</sub> = 5.0 MHz, ADC10ON = 0, REFON = 1, REF2_5V = 0, REFOUT = 1, ADC10SR = 0	25°C	3 V		1.1		mA
I <sub>REFB,1</sub>	Reference buffer supply current with ADC10SR = 1 <sup>(4)</sup>	f <sub>ADC10CLK</sub> = 5.0 MHz, ADC10ON = 0, REFON = 1, REF2_5V = 0, REFOUT = 1, ADC10SR = 1	25°C	3 V		0.5		mA
Cı	Input capacitance	Only one terminal Ax can be selected at one time	25°C	3 V			27	pF
R <sub>I</sub>	Input MUX ON resistance	$0 \text{ V} \leq \text{V}_{Ax} \leq \text{V}_{CC}$	25°C	3 V		1000		Ω

<sup>(1)</sup> The leakage current is defined in the leakage current table with Px.y/Ax parameter.

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<sup>2)</sup> The analog input voltage range must be within the selected reference voltage range  $V_{R+}$  to  $V_{R-}$  for valid conversion results.

<sup>(3)</sup> The internal reference supply current is not included in current consumption parameter I<sub>ADC10</sub>.

<sup>(4)</sup> The internal reference current is supplied through terminal V<sub>CC</sub>. Consumption is independent of the ADC10ON control bit, unless a conversion is active. The REFON bit enables the built-in reference to settle before starting an A/D conversion.



## 5.30 10-Bit ADC, Built-In Voltage Reference (MSP430G2x33 Only)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
V	Positive built-in reference	$I_{VREF+} \le 1 \text{ mA}, REF2\_5V = 0$		2.2			<b>\</b>
$V_{CC,REF+}$	analog supply voltage range	I <sub>VREF+</sub> ≤ 1 mA, REF2_5V = 1		2.9			V
V	Positive built-in reference	$I_{VREF+} \le I_{VREF+}$ max, REF2_5V = 0	3 V	1.41	1.5	1.59	<b>\</b>
$V_{REF+}$	voltage	$I_{VREF+} \le I_{VREF+}$ max, REF2_5V = 1	3 V	2.35	2.5	2.65	V
I <sub>LD,VREF+</sub>	Maximum VREF+ load current		3 V			±1	mA
	VDEE , lood roundation	$I_{VREF+}$ = 500 μA ±100 μA, Analog input voltage $V_{Ax}$ ≈ 0.75 V, REF2_5V = 0	- 3 V			±2	5
	VREF+ load regulation	$I_{VREF+}$ = 500 μA ±100 μA, Analog input voltage $V_{AX}$ ≈ 1.25 V, REF2_5V = 1	3 V			±2	LSB
	V <sub>REF+</sub> load regulation response time	$I_{VREF+}$ = 100 μA $\rightarrow$ 900 μA, $V_{Ax}$ ≈ 0.5 × VREF+, Error of conversion result ≤ 1 LSB, ADC10SR = 0	3 V			400	ns
C <sub>VREF+</sub>	Maximum capacitance at pin VREF+	I <sub>VREF+</sub> ≤ ±1 mA, REFON = 1, REFOUT = 1	3 V			100	pF
TC <sub>REF+</sub>	Temperature coefficient	I <sub>VREF+</sub> = const with 0 mA ≤ I <sub>VREF+</sub> ≤ 1 mA	3 V			±100	°C
t <sub>REFON</sub>	Settling time of internal reference voltage to 99.9% VREF	$I_{VREF+}$ = 0.5 mA, REF2_5V = 0, REFON = 0 $\rightarrow$ 1	3.6 V			30	μs
t <sub>REFBURST</sub>	Settling time of reference buffer to 99.9% VREF	$I_{VREF+}$ = 0.5 mA, REF2_5V = 1, REFON = 1, REFBURST = 1, ADC10SR = 0	3 V			2	μs



# 5.31 10-Bit ADC, External Reference<sup>(1)</sup> (MSP430G2x33 Only)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP MAX	UNIT
VEREF+	Positive external reference input	VEREF+ > VEREF-, SREF1 = 1, SREF0 = 0		1.4	V <sub>CC</sub>	V
VEREF+	voltage range (2)	VEREF- $\leq$ VEREF+ $\leq$ V <sub>CC</sub> - 0.15 V, SREF1 = 1, SREF0 = 1 (3)		1.4	3	V
VEREF-	Negative external reference input voltage range <sup>(4)</sup>	VEREF+ > VEREF-		0	1.2	>
ΔVEREF	Differential external reference input voltage range, ΔVEREF = VEREF+ – VEREF-	VEREF+ > VEREF- (5)		1.4	V <sub>CC</sub>	>
	Static input current into VEREE	$0 \text{ V} \leq \text{VEREF+} \leq \text{V}_{CC}$ , SREF1 = 1, SREF0 = 0	3 V		±1	
I <sub>VEREF+</sub>	Static input current into VEREF+	$0 \text{ V} \le \text{VEREF+} \le \text{V}_{\text{CC}} - 0.15 \text{ V} \le 3 \text{ V},$ SREF1 = 1, SREF0 = $1^{(3)}$	3 V		0	μA
I <sub>VEREF</sub>	Static input current into VEREF-	0 V ≤ VEREF- ≤ V <sub>CC</sub>	3 V		±1	μΑ

<sup>(1)</sup> The external reference is used during conversion to charge and discharge the capacitance array. The input capacitance, C<sub>I</sub>, is also the dynamic load for an external reference during conversion. The dynamic impedance of the reference supply should follow the recommendations on analog-source impedance to allow the charge to settle for 10-bit accuracy.

## 5.32 10-Bit ADC, Timing Parameters (MSP430G2x33 Only)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

P	ARAMETER	TEST CONDITIO	NS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
£	ADC10 input clock	For specified performance of	ADC10SR = 0	3 V	0.45		6.3	MHz
†ADC10CLK	frequency	ADC10 linearity parameters	ADC10SR = 1	3 V	0.45		1.5	IVITZ
f <sub>ADC10OSC</sub>	ADC10 built-in oscillator frequency	ADC10DIVx = 0, ADC10SSELX f <sub>ADC10CLK</sub> = f <sub>ADC10OSC</sub>	ζ = 0,	3 V	3.7		6.3	MHz
		ADC10 built-in oscillator, ADC1 fADC10CLK = fADC10OSC	0SSELx = 0,	3 V	2.06		3.51	
tCONVERT	Conversion time	f <sub>ADC10CLK</sub> from ACLK, MCLK, c ADC10SSELx ≠ 0	r SMCLK:			13 x ADC10DIV x 1 / f <sub>ADC10CLK</sub>		μs
t <sub>ADC10ON</sub>	Turnon settling time of the ADC	(1)					100	ns

The condition is that the error in a conversion started after t<sub>ADC10ON</sub> is less than ±0.5 LSB. The reference and input signal are already settled.

#### 5.33 10-Bit ADC, Linearity Parameters (MSP430G2x33 Only)

	PARAMETER	TEST CONDITIONS	V <sub>CC</sub>	MIN TYP	MAX	UNIT
Eı	Integral linearity error		3 V		±1	LSB
E <sub>D</sub>	Differential linearity error		3 V		±1	LSB
Eo	Offset error	Source impedance $R_S$ < 100 $\Omega$	3 V		±1	LSB
E <sub>G</sub>	Gain error		3 V	±1.1	±2	LSB
E <sub>T</sub>	Total unadjusted error		3 V	±2	±5	LSB

<sup>(2)</sup> The accuracy limits the minimum positive external reference voltage. Lower reference voltage levels may be applied with reduced accuracy requirements.

<sup>(3)</sup> Under this condition, the external reference is internally buffered. The reference buffer is active and requires the reference buffer supply current I<sub>REFB</sub>. The current consumption can be limited to the sample and conversion period with REBURST = 1.

<sup>(4)</sup> The accuracy limits the maximum negative external reference voltage. Higher reference voltage levels may be applied with reduced accuracy requirements.

<sup>(5)</sup> The accuracy limits the minimum external differential reference voltage. Lower differential reference voltage levels may be applied with reduced accuracy requirements.



#### 5.34 10-Bit ADC, Temperature Sensor and Built-In V<sub>MID</sub> (MSP430G2x33 Only)

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
I <sub>SENSOR</sub>	Temperature sensor supply current <sup>(1)</sup>	REFON = 0, INCHx = 0Ah, T <sub>A</sub> = 25°C	3 V		60		μΑ
TC <sub>SENSOR</sub>		ADC10ON = 1, INCHx = 0Ah (2)	3 V		3.55		mV/°C
t <sub>Sensor(sample)</sub>	Sample time required if channel 10 is selected <sup>(3)</sup>	ADC10ON = 1, INCHx = 0Ah, Error of conversion result ≤ 1 LSB	3 V	30			μs
I <sub>VMID</sub>	Current into divider at channel 11	ADC10ON = 1, INCHx = 0Bh	3 V			(4)	μΑ
V <sub>MID</sub>	V <sub>CC</sub> divider at channel 11	ADC10ON = 1, INCHx = 0Bh, $V_{MID} \approx 0.5 \times V_{CC}$	3 V		1.5		V
t <sub>VMID(sample)</sub>	Sample time required if channel 11 is selected <sup>(5)</sup>	ADC10ON = 1, INCHx = 0Bh, Error of conversion result ≤ 1 LSB	3 V	1220			ns

The sensor current  $I_{SENSOR}$  is consumed if (ADC10ON = 1 and REFON = 1) or (ADC10ON = 1 and INCH = 0Ah and sample signal is high). When REFON = 1, I<sub>SENSOR</sub> is included in I<sub>REF+</sub>. When REFON = 0, I<sub>SENSOR</sub> applies during conversion of the temperature sensor input (INCH = 0Ah).

The following formula can be used to calculate the temperature sensor output voltage:

- $V_{Sensor,typ} = \overrightarrow{TC}_{Sensor} \ (273 + T \ [^{\circ}C] \ ) + V_{Offset,sensor} \ [mV] \ or \\ V_{Sensor,typ} = TC_{Sensor} \ T \ [^{\circ}C] + V_{Sensor} (T_A = 0^{\circ}C) \ [mV] \\ The typical equivalent impedance of the sensor is 51 k<math>\Omega$ . The sample time required includes the sensor-on time  $t_{SENSOR(on)}$ .
- No additional current is needed. The V<sub>MID</sub> is used during sampling.
- The on-time  $t_{VMID(on)}$  is included in the sampling time  $t_{VMID(sample)}$ ; no additional on time is needed.

#### 5.35 Flash Memory

	PARAMETER	TEST CONDITIONS	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
V <sub>CC(PGM/ERASE)</sub>	Program and erase supply voltage			2.2		3.6	V
f <sub>FTG</sub>	Flash timing generator frequency			257		476	kHz
I <sub>PGM</sub>	Supply current from V <sub>CC</sub> during program		2.2 V, 3.6 V		1	5	mA
I <sub>ERASE</sub>	Supply current from V <sub>CC</sub> during erase		2.2 V, 3.6 V		1	7	mA
t <sub>CPT</sub>	Cumulative program time <sup>(1)</sup>		2.2 V, 3.6 V			10	ms
t <sub>CMErase</sub>	Cumulative mass erase time		2.2 V, 3.6 V	20			ms
	Program and erase endurance			10 <sup>4</sup>	10 <sup>5</sup>		cycles
t <sub>Retention</sub>	Data retention duration	$T_J = 25^{\circ}C$		100			years
t <sub>Word</sub>	Word or byte program time	See (2)			30		t <sub>FTG</sub>
t <sub>Block, 0</sub>	Block program time for first byte or word	See (2)			25		t <sub>FTG</sub>
t <sub>Block, 1-63</sub>	Block program time for each additional byte or word	See (2)			18		t <sub>FTG</sub>
t <sub>Block, End</sub>	Block program end-sequence wait time	See (2)			6		t <sub>FTG</sub>
t <sub>Mass Erase</sub>	Mass erase time	See (2)			10593		t <sub>FTG</sub>
t <sub>Seg Erase</sub>	Segment erase time	See (2)			4819		t <sub>FTG</sub>

Do not exceed the cumulative program time when writing to a 64-byte flash block. This parameter applies to all programming methods: individual word or byte write and block write modes.

These values are hardwired into the state machine of the flash controller ( $t_{FTG} = 1/f_{FTG}$ ).



#### 5.36 RAM

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

PARAMETER	TEST CONDITIONS	MIN MAX	UNIT
V <sub>(RAMh)</sub> RAM retention supply voltage <sup>(1)</sup>	CPU halted	1.6	V

<sup>(1)</sup> This parameter defines the minimum supply voltage V<sub>CC</sub> when the data in RAM remains unchanged. No program execution should happen during this supply voltage condition.

#### 5.37 JTAG and Spy-Bi-Wire Interface

over recommended ranges of supply voltage and operating free-air temperature (unless otherwise noted)

	0 117 0 1 0	•		,		
	PARAMETER	V <sub>cc</sub>	MIN	TYP	MAX	UNIT
f <sub>SBW</sub>	Spy-Bi-Wire input frequency	2.2 V	0		20	MHz
t <sub>SBW,Low</sub>	Spy-Bi-Wire low clock pulse duration	2.2 V	0.025		15	μs
t <sub>SBW,En</sub>	Spy-Bi-Wire enable time (TEST high to acceptance of first clock edge <sup>(1)</sup> )	2.2 V			1	μs
t <sub>SBW,Ret</sub>	Spy-Bi-Wire return to normal operation time	2.2 V	15		100	μs
f <sub>TCK</sub>	TCK input frequency <sup>(2)</sup>	2.2 V	0		5	MHz
R <sub>Internal</sub>	Internal pulldown resistance on TEST	2.2 V	25	60	90	kΩ

<sup>(1)</sup> Tools that access the Spy-Bi-Wire interface must wait for the maximum t<sub>SBW,En</sub> time after pulling the TEST/SBWCLK pin high before applying the first SBWCLK clock edge.

# 5.38 JTAG Fuse<sup>(1)</sup>

	PARAMETER	TEST CONDITIONS	MIN	MAX	UNIT
V <sub>CC(FB)</sub>	Supply voltage during fuse-blow condition	T <sub>A</sub> = 25°C	2.5		V
V <sub>FB</sub>	Voltage level on TEST for fuse blow		6	7	V
I <sub>FB</sub>	Supply current into TEST during fuse blow			100	mA
t <sub>FB</sub>	Time to blow fuse			1	ms

<sup>(1)</sup> After the fuse is blown, no further access to the JTAG/Test, Spy-Bi-Wire, and emulation features is possible, and JTAG is switched to bypass mode.

<sup>(2)</sup> f<sub>TCK</sub> may be restricted to meet the timing requirements of the module selected.



#### 6 Detailed Description

#### 6.1 CPU

The MSP430 CPU has a 16-bit RISC architecture that is highly transparent to the application. All operations, other than program-flow instructions, are performed as register operations in conjunction with seven addressing modes for source operand and four addressing modes for destination operand.

The CPU is integrated with 16 registers that provide reduced instruction execution time. The register-to-register operation execution time is one cycle of the CPU clock. Four of the registers, R0 to R3, are dedicated as program counter, stack pointer, status register, and constant generator, respectively. The remaining registers are general-purpose registers (see Figure 6-1).

Peripherals are connected to the CPU using data, address, and control buses, and can be handled with all instructions.

The instruction set consists of the original 51 instructions with three formats and seven address modes and additional instructions for the expanded address range. Each instruction can operate on word and byte data.

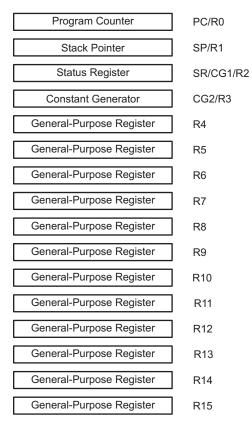


Figure 6-1. Integrated CPU Registers



#### 6.2 Instruction Set

The instruction set consists of 51 instructions with three formats and seven address modes. Each instruction can operate on word and byte data. Table 6-1 lists examples of the three types of instruction formats. Table 6-2 lists the address modes.

**Table 6-1. Instruction Word Formats** 

INSTRUCTION FORMAT	EXAMPLE	OPERATION
Dual operands, source-destination	ADD R4,R5	R4 + R5 → R5
Single operands, destination only	CALL R8	$PC \rightarrow (TOS), R8 \rightarrow PC$
Relative jump, unconditional or conditional	JNE	Jump-on-equal bit = 0

## **Table 6-2. Address Mode Descriptions**

ADDRESS MODE	S <sup>(1)</sup>	D	SYNTAX	EXAMPLE	OPERATION
Register	✓	✓	MOV Rs,Rd	MOV R10,R11	R10 → R11
Indexed	✓	✓	MOV X(Rn),Y(Rm)	MOV 2(R5),6(R6)	$M(2+R5) \rightarrow M(6+R6)$
Symbolic (PC relative)	✓	✓	MOV EDE,TONI		$M(EDE) \rightarrow M(TONI)$
Absolute	✓	<b>√</b>	MOV &MEM,&TCDAT		$M(MEM) \rightarrow M(TCDAT)$
Indirect	✓		MOV @Rn,Y(Rm)	MOV @R10,Tab(R6)	$M(R10) \rightarrow M(Tab+R6)$
Indirect autoincrement	1		MOV @Rn+,Rm	MOV @R10+,R11	M(R10) → R11 R10 + 2 → R10
Immediate	✓		MOV #X,TONI	MOV #45,TONI	#45 → M(TONI)

<sup>(1)</sup> S = source, D = destination



## 6.3 Operating Modes

These microcontrollers have one active mode and five software-selectable low-power modes of operation. An interrupt event can wake the device from any of the low-power modes, service the request, and restore back to the low-power mode on return from the interrupt program.

Software can configure the following operating modes:

- Active mode (AM)
  - All clocks are active
- Low-power mode 0 (LPM0)
  - CPU is disabled
  - ACLK and SMCLK remain active, MCLK is disabled
- Low-power mode 1 (LPM1)
  - CPU is disabled
  - ACLK and SMCLK remain active, MCLK is disabled
  - DC generator of the DCO is disabled if DCO not used in active mode
- Low-power mode 2 (LPM2)
  - CPU is disabled
  - MCLK and SMCLK are disabled
  - DC generator of the DCO remains enabled
  - ACLK remains active
- Low-power mode 3 (LPM3)
  - CPU is disabled
  - MCLK and SMCLK are disabled
  - DC generator of the DCO is disabled
  - ACLK remains active
- Low-power mode 4 (LPM4)
  - CPU is disabled
  - ACLK is disabled
  - MCLK and SMCLK are disabled
  - DC generator of the DCO is disabled
  - Crystal oscillator is stopped



#### 6.4 **Interrupt Vector Addresses**

The interrupt vectors and the power-up starting address are in the address range 0FFFFh to 0FFC0h (see Table 6-3). The vector contains the 16-bit address of the appropriate interrupt handler instruction sequence.

If the reset vector (at address 0FFFEh) contains 0FFFFh (for example, if the flash is not programmed), the CPU goes into LPM4 immediately after power-up.

Table 6-3. Interrupt Sources, Flags, and Vectors

INTERRUPT SOURCE	INTERRUPT FLAG	SYSTEM INTERRUPT	WORD ADDRESS	PRIORITY
Power up External reset Watchdog Timer+ Flash key violation PC out of range <sup>(1)</sup>	PORIFG RSTIFG WDTIFG KEYV <sup>(2)</sup>	Reset	0FFFEh	31, highest
NMI Oscillator fault Flash memory access violation	NMIIFG OFIFG ACCVIFG <sup>(2)</sup>	(non)-maskable <sup>(3)</sup> (non)-maskable (non)-maskable	0FFFCh	30
Timer1_A3	TACCR0 CCIFG <sup>(4)</sup>	maskable	0FFFAh	29
Timer1_A3	TACCR2 TACCR1 CCIFG, TAIFG <sup>(2)(4)</sup>	maskable	0FFF8h	28
			0FFF6h	27
Watchdog Timer+	WDTIFG	maskable	0FFF4h	26
Timer0_A3	TACCR0 CCIFG <sup>(4)</sup>	maskable	0FFF2h	25
Timer0_A3	TACCR2 TACCR1 CCIFG, TAIFG	maskable	0FFF0h	24
USCI_A0, USCI_B0 receive USCI_B0 I <sup>2</sup> C status	UCA0RXIFG, UCB0RXIFG <sup>(2)(5)</sup>	maskable	0FFEEh	23
USCI_A0, USCI_B0 transmit USCI_B0 I <sup>2</sup> C receive or transmit	UCA0TXIFG, UCB0TXIFG <sup>(2)(6)</sup>	maskable	0FFECh	22
ADC10 (MSP430G2x33 only)	ADC10IFG <sup>(4)</sup>	maskable	0FFEAh	21
			0FFE8h	20
I/O Port P2 (up to eight flags)	P2IFG.0 to P2IFG.7 <sup>(2)(4)</sup>	maskable	0FFE6h	19
I/O Port P1 (up to eight flags)	P1IFG.0 to P1IFG.7 <sup>(2)(4)</sup>	maskable	0FFE4h	18
			0FFE2h	17
			0FFE0h	16
See (7)			0FFDEh	15
See <sup>(8)</sup>			0FFDEh to 0FFC0h	14 to 0, lowest

<sup>(1)</sup> A reset is generated if the CPU tries to fetch instructions from within the module register memory address range (0h to 01FFh) or from within unused address ranges.

Multiple source flags

<sup>(</sup>non)-maskable: the individual interrupt-enable bit can disable an interrupt event, but the general interrupt enable cannot.

Interrupt flags are in the module.

In SPI mode: UCB0RXIFG. In  $I^2$ C mode: UCALIFG, UCNACKIFG, ICSTTIFG, UCSTPIFG. In UART or SPI mode: UCB0TXIFG. In  $I^2$ C mode: UCB0RXIFG, UCB0TXIFG.

This location is used as bootloader security key (BSLSKEY). A 0xAA55 at this location disables the BSL completely. A zero (0h) disables the erasure of the flash if an invalid password is supplied.

The interrupt vectors at addresses 0FFDEh to 0FFC0h are not used in this device and can be used for regular program code if necessary.



## 6.5 Special Function Registers (SFRs)

Most interrupt and module enable bits are collected into the lowest address space. Special function register bits not allocated to a functional purpose are not physically present in the device. Simple software access is provided with this arrangement.

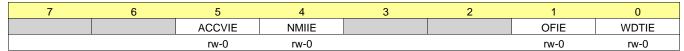
#### Legend

rw Bit can be read and written.

rw-0, rw-1 Bit can be read and written. It is reset or set by PUC. rw-(0), rw-(1) Bit can be read and written. It is reset or set by POR.

SFR bit is not present in device.

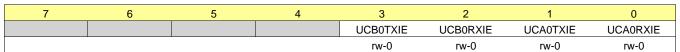
#### Figure 6-2. Interrupt Enable Register 1 (Address = 00h)



## Table 6-4. Interrupt Enable Register 1 Description

Bit	Field	Туре	Reset	Description
5	ACCVIE	RW	0h	Flash access violation interrupt enable
4	NMIIE	RW	0h	(Non)maskable interrupt enable
1	OFIE	RW	0h	Oscillator fault interrupt enable
0	WDTIE	RW	0h	Watchdog Timer interrupt enable. Inactive if watchdog mode is selected. Active if Watchdog Timer is configured in interval timer mode.

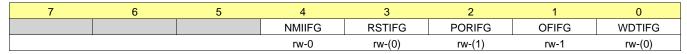
#### Figure 6-3. Interrupt Enable Register 2 (Address = 01h)



#### Table 6-5. Interrupt Enable Register 2 Description

Bit	Field	Туре	Reset	Description
3	UCB0TXIE	RW	0h	USCI_B0 transmit interrupt enable
2	UCB0RXIE	RW	0h	USCI_B0 receive interrupt enable
1	UCA0TXIE	RW	0h	USCI_A0 transmit interrupt enable
0	UCA0RXIE	RW	0h	USCI_A0 receive interrupt enable

#### Figure 6-4. Interrupt Flag Register 1 (Address = 02h)



#### Table 6-6. Interrupt Flag Register 1 Description

	· · · · · · · · · · · · · · · · · · · ·						
Bit	Field	Туре	Reset	Description			
4	NMIIFG	RW	0h	Set by the RST/NMI pin			
3	RSTIFG	RW	0h	External reset interrupt flag. Set on a reset condition at $\overline{\text{RST}}/\text{NMI}$ pin in reset mode. Reset on $V_{\text{CC}}$ power-up.			
2	PORIFG	RW	1h	Power-On Reset interrupt flag. Set on V <sub>CC</sub> power-up.			
1	OFIFG	RW	1h	Flag set on oscillator fault.			
0	WDTIFG	RW	0h	Set on watchdog timer overflow (in watchdog mode) or security key violation. Reset on V <sub>CC</sub> power-on or a reset condition at the RST/NMI pin in reset mode.			



#### Figure 6-5. Interrupt Flag Register 2 (Address = 03h)

7	6	5	4	3	2	1	0
				UCB0TXIFG	UCB0RXIFG	UCA0TXIFG	UCA0RXIFG
				rw-1	rw-0	rw-1	rw-0

## Table 6-7. Interrupt Flag Register 2 Description

Bit	Field	Туре	Reset	Description
3	UCB0TXIFG	RW	0h	USCI_B0 transmit interrupt flag
2	UCB0RXIFG	RW	1h	USCI_B0 receive interrupt flag
1	UCA0TXIFG	RW	1h	USCI_A0 transmit interrupt flag
0	UCA0RXIFG	RW	0h	USCI_A0 receive interrupt flag

## 6.6 Memory Organization

Table 6-8 summarizes the memory map.

**Table 6-8. Memory Organization** 

		MSP430G2233 MSP430G2203	MSP430G2333 MSP430G2303	MSP430G2433 MSP430G2403	MSP430G2533
Memory	Size	2KB	4KB	8KB	16KB
Main: interrupt vector	Flash	FFFFh to FFC0h	FFFFh to FFC0h	FFFFh to FFC0h	FFFFh to FFC0h
Main: code memory	Flash	FFFFh to F800h	FFFFh to F000h	FFFFh to E000h	FFFFh to C000h
Information memory	Size	256 byte	256 byte	256 byte	256 byte
	Flash	010FFh to 01000h	010FFh to 01000h	010FFh to 01000h	010FFh to 01000h
RAM	Size	256 byte	256 byte	512 byte	512 byte
		02FFh to 0200h	02FFh to 0200h	03FFh to 0200h	03FFh to 0200h
Peripherals	16-bit	01FFh to 0100h	01FFh to 0100h	01FFh to 0100h	01FFh to 0100h
	8-bit	0FFh to 010h	0FFh to 010h	0FFh to 010h	0FFh to 010h
	8-bit SFR	0Fh to 00h	0Fh to 00h	0Fh to 00h	0Fh to 00h

#### 6.7 Bootloader (BSL)

The MSP430 BSL enables users to program the flash memory or RAM using a UART serial interface. Access to the MSP430 memory through the BSL is protected by user-defined password. For complete description of the features of the BSL and its implementation, see the MSP430 Programming With the Bootloader User's Guide (SLAU319). Table 6-9 lists the BSL function pins.

**Table 6-9. BSL Function Pins** 

BSL FUNCTION	20-PIN PW PACKAGE 20-PIN N PACKAGE	28-PIN PW PACKAGE	32-PIN RHB PACKAGE
Data transmit	3 - P1.1	3 - P1.1	1 - P1.1
Data receive	7 - P1.5	7 - P1.5	5 - P1.5



#### 6.8 Flash Memory

The flash memory can be programmed through the Spy-Bi-Wire/JTAG port or in-system by the CPU. The CPU can perform single-byte and single-word writes to the flash memory. Features of the flash memory include:

- Flash memory has n segments of main memory and four segments of information memory (A to D) of 64 bytes each. Each segment in main memory is 512 bytes in size.
- Segments 0 to n may be erased in one step, or each segment may be individually erased.
- Segments A to D can be erased individually or as a group with segments 0 to n. Segments A to D are also called *information memory*.
- Segment A contains calibration data. After reset segment A is protected against programming and
  erasing. It can be unlocked but care should be taken not to erase this segment if the device-specific
  calibration data is required.

#### 6.9 Peripherals

Peripherals are connected to the CPU through data, address, and control buses. The peripherals can be managed using all instructions. For complete module descriptions, see the *MSP430x2xx Family User's Guide* (SLAU144).

## 6.9.1 Oscillator and System Clock

The clock system is supported by the basic clock module that includes support for a 32768-Hz watch crystal oscillator, an internal very-low-power low-frequency oscillator and an internal digitally controlled oscillator (DCO). The basic clock module is designed to meet the requirements of both low system cost and low power consumption. The internal DCO provides a fast turnon clock source and stabilizes in less than 1 µs. The basic clock module provides the following clock signals:

- Auxiliary clock (ACLK), sourced either from a 32768-Hz watch crystal or the internal LF oscillator.
- Main clock (MCLK), the system clock used by the CPU.
- Sub-Main clock (SMCLK), the subsystem clock used by the peripheral modules.

The DCO settings to calibrate the DCO output frequency are stored in the information memory segment A.



#### 6.9.2 Calibration Data Stored in Information Memory Segment A

Calibration data is stored for both the DCO and for ADC10 organized in a tag-length-value structure (see Table 6-10 and Table 6-11).

Table 6-10. Tags Used by the ADC Calibration Tags

NAME	ADDRESS	VALUE	DESCRIPTION
TAG_DCO_30	0x10F6	0x01	DCO frequency calibration at V <sub>CC</sub> = 3 V and T <sub>A</sub> = 30°C
TAG_ADC10_1	0x10DA	0x10	ADC10_1 calibration tag
TAG_EMPTY	_	0xFE	Identifier for empty memory areas

Table 6-11. Labels Used by the ADC Calibration Tags

LABEL	ADDRESS OFFSET	SIZE	CONDITION AT CALIBRATION
CAL_ADC_25T85	0x0010	word	INCHx = 1010b, REF2_5 = 1, T <sub>A</sub> = 85°C
CAL_ADC_25T30	0x000E	word	INCHx = 1010b, REF2_5 = 1, T <sub>A</sub> = 30°C
CAL_ADC_25VREF_FACTOR	0x000C	word	REF2_5 = 1, $T_A = 30$ °C, $I_{VREF+} = 1$ mA
CAL_ADC_15T85	0x000A	word	INCHx = 1010b, REF2_5 = 0, T <sub>A</sub> = 85°C
CAL_ADC_15T30	0x0008	word	INCHx = 1010b, REF2_5 = 0, T <sub>A</sub> = 30°C
CAL_ADC_15VREF_FACTOR	0x0006	word	REF2_5 = 0, $T_A = 30$ °C, $I_{VREF+} = 0.5$ mA
CAL_ADC_OFFSET	0x0004	word	External VREF = 1.5 V, f <sub>ADC10CLK</sub> = 5 MHz
CAL_ADC_GAIN_FACTOR	0x0002	word	External VREF = 1.5 V, f <sub>ADC10CLK</sub> = 5 MHz
CAL_BC1_1MHZ	0x0009	byte	_
CAL_DCO_1MHZ	0x0008	byte	_
CAL_BC1_8MHZ	0x0007	byte	-
CAL_DCO_8MHZ	0x0006	byte	-
CAL_BC1_12MHZ	0x0005	byte	-
CAL_DCO_12MHZ	0x0004	byte	-
CAL_BC1_16MHZ	0x0003	byte	-
CAL_DCO_16MHZ	0x0002	byte	_

#### 6.9.3 Brownout

The brownout circuit is implemented to provide the proper internal reset signal to the device during power on and power off.

## 6.9.4 Digital I/O

Up to three 8-bit I/O ports are implemented:

- All individual I/O bits are independently programmable.
- Any combination of input, output, and interrupt condition (port P1 and port P2 only) is possible.
- Edge-selectable interrupt input capability for all bits of port P1 and port P2 (if available).
- · Read/write access to port-control registers is supported by all instructions.
- Each I/O has an individually programmable pullup or pulldown resistor.
- Each I/O has an individually programmable pin oscillator enable bit to enable low-cost capacitive touch detection.

## 6.9.5 WDT+ Watchdog Timer

The primary function of the watchdog timer (WDT+) module is to perform a controlled system restart after a software problem occurs. If the selected time interval expires, a system reset is generated. If the watchdog function is not needed in an application, the module can be disabled or configured as an interval timer and can generate interrupts at selected time intervals.

# 6.9.6 Timer\_A3 (TA0, TA1)

Timer0\_A3 and Timer1\_A3 are 16-bit timers/counters with three capture/compare registers. Timer\_A3 can support multiple capture/compares, PWM outputs, and interval timing (see Table 6-12 and Table 6-13). Timer\_A3 also has extensive interrupt capabilities. Interrupts may be generated from the counter on overflow conditions and from each of the capture/compare registers.

Table 6-12. Timer0\_A3 Signal Connections

INP	UT PIN NUMI	BER	DEVICE	MODULE	MODULE	MODULE	OUTF	PUT PIN NUM	IBER	
PW20, N20	PW28	RHB32	INPUT SIGNAL	INPUT NAME	BLOCK	OUTPUT SIGNAL	PW20, N20	PW28	RHB32	
P1.0-2	P1.0-2	P1.0-31	TACLK	TACLK						
			ACLK	ACLK	- Timer NA	NIA				
			SMCLK	SMCLK		NA				
PinOsc	PinOsc	PinOsc	TACLK	INCLK						
P1.1-3	P1.1-3	P1.1-1	TA0.0	CCI0A	- CCR0			P1.1-3	P1.1-3	P1.1-1
			ACLK	CCI0B		CCR0 TA0	P1.5-7	P1.5-7	P1.5-5	
			V <sub>SS</sub>	GND			_	P3.4-15	P3.4-14	
			V <sub>CC</sub>	V <sub>CC</sub>						
P1.2-4	P1.2-4	P1.2-2	TA0.1	CCI1A			P1.2-4	P1.2-4	P1.2-2	
			CAOUT	CCI1B	CCD4	T 4 4	P1.6-14	P1.6-22	P1.6-21	
			V <sub>SS</sub>	GND	CCR1	TA1	P2.6-19	P2.6-27	P2.6-26	
			V <sub>CC</sub>	V <sub>CC</sub>			_	P3.5-19	P3.5-18	
-	P3.0-9	P3.0-7	TA0.2	CCI2A			_	P3.0-9	P3.0-7	
PinOsc	PinOsc	PinOsc	TA0.2	CCI2B	CCDa	TA 2	_	P3.6-20	P3.6-19	
			V <sub>SS</sub>	GND	CCR2	TA2				
			V <sub>CC</sub>	V <sub>CC</sub>						



#### Table 6-13. Timer1 A3 Signal Connections

INP	UT PIN NUME	BER	DEVICE	MODULE	MODULE	MODULE	OUTI	PUT PIN NUN	BER			
PW20, N20	PW28	RHB32	INPUT SIGNAL	INPUT NAME	BLOCK	OUTPUT SIGNAL	PW20, N20	PW28	RHB32			
-	P3.7-21	P3.7-20	TACLK	TACLK								
			ACLK	ACLK	Timer NA	Times						
			SMCLK	SMCLK								
-	P3.7-21	P3.7-20	TACLK	INCLK								
P2.0-8	P2.0-10	P2.0-9	TA1.0	CCI0A	- CCR0			P2.0-8	P2.0-10	P2.0-9		
P2.3-11	P2.3-16	P2.3-12	TA1.0	CCI0B		CCR0 TA0	P2.3-11	P2.3-16	P2.3-15			
			V <sub>SS</sub>	GND				P3.1-8	P3.1-6			
			V <sub>CC</sub>	V <sub>CC</sub>								
P2.1-9	P2.1-11	P2.1-10	TA1.1	CCI1A			P2.1-9	P2.1-11	P2.1-10			
P2.2-10	P2.2-12	P2.2-11	TA1.1	CCI1B	CCD4	0004	P2.2-10	P2.2-12	P2.2-11			
			$V_{SS}$	GND	CCR1	TA1		P3.2-13	P3.2-12			
			V <sub>CC</sub>	V <sub>CC</sub>								
P2.4-12	P2.4-17	P2.4-16	TA1.2	CCI2A			P2.4-12	P2.4-17	P2.4-16			
P2.5-13	P2.5-18	P2.5-17	TA1.2	CCI2B	CCD2	TA0	P2.5-13	P2.5-18	P2.5-17			
			V <sub>SS</sub>	GND	- CCR2	CCR2 TA2	CCR2	CCR2 TA2	1 AZ		P3.3-14	P3.3-13
			V <sub>CC</sub>	V <sub>CC</sub>								

## 6.9.7 Universal Serial Communications Interface (USCI)

The USCI module is used for serial data communication. The USCI module supports synchronous communication protocols such as SPI (3-pin or 4-pin) and I<sup>2</sup>C, and asynchronous communication protocols such as UART, enhanced UART with automatic baud rate detection (LIN), and IrDA. Not all packages support the USCI functionality.

USCI\_A0 provides support for SPI (3-pin or 4-pin), UART, enhanced UART, and IrDA.

USCI B0 provides support for SPI (3-pin or 4-pin) and I<sup>2</sup>C.

#### 6.9.8 ADC10 (MSP430G2x33 Only)

The ADC10 module supports fast 10-bit analog-to-digital conversions. The module implements a 10-bit SAR core, sample select control, reference generator, and data transfer controller (DTC) for automatic conversion result handling, allowing ADC samples to be converted and stored without any CPU intervention.



## 6.9.9 Peripheral File Map

Table 6-14 lists the registers that support word access. Table 6-15 that support byte access.

**Table 6-14. Peripherals With Word Access** 

MODULE	REGISTER DESCRIPTION	ACRONYM	OFFSET
	ADC data transfer start address	ADC10SA	1BCh
ADC40 (MCD420C0+22)	ADC memory	ADC10MEM	1B4h
ADC10 (MSP430G2x33 only)	ADC control register 1	ADC10CTL1	1B2h
	ADC control register 0	ADC10CTL0	1B0h
	Capture/compare register	TA1CCR2	0196h
	Capture/compare register	TA1CCR1	0194h
	Capture/compare register	TA1CCR0	0192h
	Timer_A register	TA1R	0190h
Timer1_A3	Capture/compare control	TA1CCTL2	0186h
	Capture/compare control	TA1CCTL1	0184h
	Capture/compare control	TA1CCTL0	0182h
	Timer_A control	TA1CTL	0180h
	Timer_A interrupt vector	TA1IV	011Eh
	Capture/compare register	TA0CCR2	0176h
	Capture/compare register	TA0CCR1	0174h
	Capture/compare register	TA0CCR0	0172h
	Timer_A register	TA0R	0170h
Timer0_A3	Capture/compare control	TA0CCTL2	0166h
	Capture/compare control	TA0CCTL1	0164h
	Capture/compare control	TA0CCTL0	0162h
	Timer_A control	TA0CTL	0160h
	Timer_A interrupt vector	TAOIV	012Eh
	Flash control 3	FCTL3	012Ch
Flash Memory	Flash control 2	FCTL2	012Ah
	Flash control 1	FCTL1	0128h
Watchdog Timer+	Watchdog timer control	WDTCTL	0120h

## Table 6-15. Peripherals With Byte Access

MODULE	REGISTER DESCRIPTION	ACRONYM	OFFSET
	USCI_B0 transmit buffer	UCB0TXBUF	06Fh
	USCI_B0 receive buffer	UCB0RXBUF	06Eh
	USCI_B0 status	UCB0STAT	06Dh
	USCI B0 I <sup>2</sup> C Interrupt enable	UCB0CIE	06Ch
LICCI DO	USCI_B0 bit rate control 1	UCB0BR1	06Bh
USCI_B0	USCI_B0 bit rate control 0	UCB0BR0	06Ah
	USCI_B0 control 1	UCB0CTL1	069h
	USCI_B0 control 0	UCB0CTL0	068h
	USCI_B0 I <sup>2</sup> C slave address	UCB0SA	011Ah
	USCI_B0 I <sup>2</sup> C own address	UCB0OA	0118h



## Table 6-15. Peripherals With Byte Access (continued)

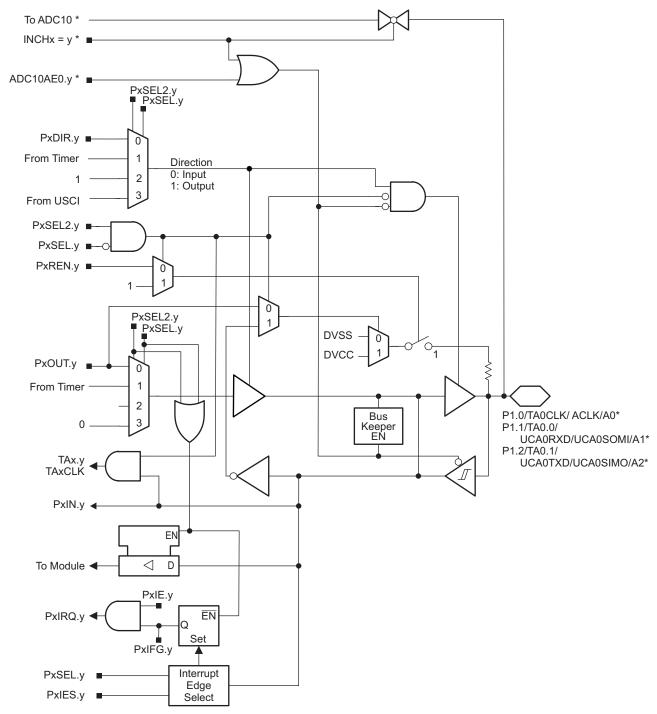
MODULE	REGISTER DESCRIPTION	ACRONYM	OFFSET
	USCI_A0 transmit buffer	UCA0TXBUF	067h
	USCI_A0 receive buffer	UCA0RXBUF	066h
	USCI_A0 status	UCA0STAT	065h
	USCI_A0 modulation control	UCA0MCTL	064h
	USCI_A0 baud rate control 1	UCA0BR1	063h
USCI_A0	USCI_A0 baud rate control 0	UCA0BR0	062h
	USCI_A0 control 1	UCA0CTL1	061h
	USCI_A0 control 0	UCA0CTL0	060h
	USCI_A0 IrDA receive control	UCA0IRRCTL	05Fh
	USCI_A0 IrDA transmit control	UCA0IRTCTL	05Eh
	USCI_A0 auto baud rate control	UCA0ABCTL	05Dh
	ADC analog enable 0	ADC10AE0	04Ah
	ADC analog enable 1	ADC10AE1	04Bh
ADC10 (MSP430G2x33 only)	ADC data transfer control register 1	ADC10DTC1	049h
	ADC data transfer control register 0	ADC10DTC0	048h
	Basic clock system control 3	BCSCTL3	053h
	Basic clock system control 2	BCSCTL2	058h
Basic Clock System+	Basic clock system control 1	BCSCTL1	057h
	DCO clock frequency control	DCOCTL	056h
	Port P3 selection 2. pin	P3SEL2	043h
	Port P3 resistor enable	P3REN	010h
Port P3	Port P3 selection	P3SEL	01Bh
(28-pin PW and 32-pin RHB only)	Port P3 direction	P3DIR	01Ah
	Port P3 output	P3OUT	019h
	Port P3 input	P3IN	018h
	Port P2 selection 2	P2SEL2	042h
	Port P2 resistor enable	P2REN	02Fh
	Port P2 selection	P2SEL	02Eh
	Port P2 interrupt enable	P2IE	02Dh
Port P2	Port P2 interrupt edge select	P2IES	02Ch
	Port P2 interrupt flag	P2IFG	02Bh
	Port P2 direction	P2DIR	02Ah
	Port P2 output	P2OUT	029h
	Port P2 input	P2IN	028h
	Port P1 selection 2	P1SEL2	041h
	Port P1 resistor enable	P1REN	027h
	Port P1 selection	P1SEL	026h
	Port P1 interrupt enable	P1IE	025h
Port P1	Port P1 interrupt edge select	P1IES	024h
	Port P1 interrupt flag	P1IFG	023h
	Port P1 direction	P1DIR	022h
	Port P1 output	P1OUT	021h
	Port P1 input	P1IN	020h
	SFR interrupt flag 2	IFG2	003h
On a sint Formation	SFR interrupt flag 1	IFG1	002h
Special Function	SFR interrupt enable 2	IE2	001h
	SFR interrupt enable 1	IE1	000h



# 6.10 I/O Port Diagrams

## 6.10.1 Port P1 Pin Diagram: P1.0 to P1.2, Input/Output With Schmitt Trigger

Figure 6-6 shows the port diagram. Table 6-16 summarizes the selection of the pin functions.



<sup>\*</sup> Note: MSP430G2x33 devices only. MSP430G2x03 devices have no ADC10.

Figure 6-6. Port P1 (P1.0 to P1.2) Diagram



## Table 6-16. Port P1 (P1.0 to P1.2) Pin Functions

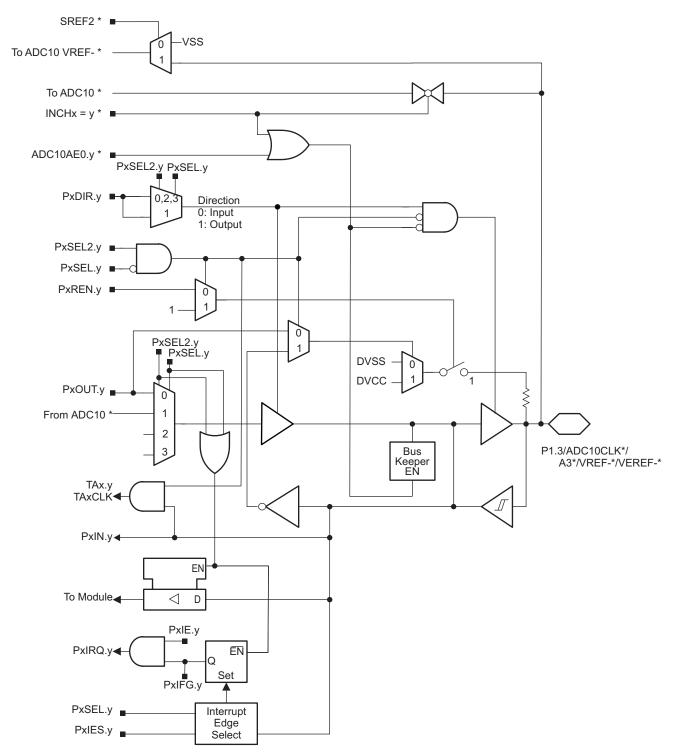
PIN NAME				CONTROL BITS OR SIGNALS <sup>(1)</sup>				
(P1.x)	х	FUNCTION	P1DIR.x	P1SEL.x	P1SEL2.x	ADC10AE.x (INCH.y = 1) <sup>(2)</sup>		
P1.0/		P1.x (I/O)	I: 0; O: 1	0	0	0		
TA0CLK/		TA0.TACLK	0	1	0	0		
ACLK/	0	ACLK	1	1	0	0		
A0 <sup>(2)</sup> /		A0	Х	Х	Х	1 (y = 0)		
Pin Osc		Capacitive sensing	Х	0	1	0		
P1.1/		P1.x (I/O)	I: 0; O: 1	0	0	0		
TA0.0/		TA0.0	1	1	0	0		
		TA0.CCI0A	0	1	0	0		
UCA0RXD/	1	UCA0RXD	from USCI	1	1	0		
UCA0SOMI/		UCA0SOMI	from USCI	1	1	0		
A1 <sup>(2)</sup> /		A1	X	Х	Х	1 (y = 1)		
Pin Osc		Capacitive sensing	X	0	1	0		
P1.2/		P1.x (I/O)	I: 0; O: 1	0	0	0		
TA0.1/		TA0.1	1	1	0	0		
		TA0.CCI1A	0	1	0	0		
UCA0TXD/	2	UCA0TXD	from USCI	1	1	0		
UCA0SIMO/		UCA0SIMO	from USCI	1	1	0		
A2 <sup>(2)</sup> /		A2	Х	Х	Х	1 (y = 2)		
Pin Osc		Capacitive sensing	Х	0	1	0		

<sup>(1)</sup> X = don't care

<sup>(2)</sup> MSP430G2x33 devices only

## 6.10.2 Port P1 Pin Diagram: P1.3, Input/Output With Schmitt Trigger

Figure 6-7 shows the port diagram. Table 6-17 summarizes the selection of the pin functions.



<sup>\*</sup> Note: MSP430G2x33 devices only. MSP430G2x03 devices have no ADC10.

Figure 6-7. Port P1 (P1.3) Diagram



## Table 6-17. Port P1 (P1.3) Pin Functions

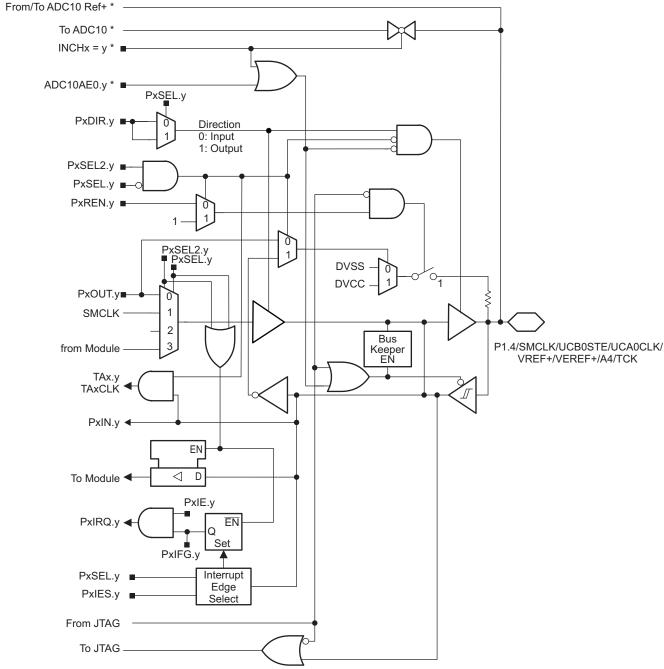
PIN NAME		FUNCTION	CONTROL BITS OR SIGNALS <sup>(1)</sup>				
(P1.x)	X		P1DIR.x	P1SEL.x	P1SEL2.x	ADC10AE.x (INCH.y = 1) <sup>(2)</sup>	
P1.3/		P1.x (I/O)	I: 0; O: 1	0	0	0	
ADC10CLK <sup>(2)</sup> /		ADC10CLK	1	1	0	0	
A3 <sup>(2)</sup> /	2	A3	X	X	X	1 (y = 3)	
VREF- <sup>(2)</sup> /	3	VREF-	Х	X	X	1	
VEREF-(2)/		VEREF-	Х	X	X	1	
Pin Osc		Capacitive sensing	Х	0	1	0	

<sup>(1)</sup> X = don't care

<sup>(2)</sup> MSP430G2x33 devices only

## 6.10.3 Port P1 Pin Diagram: P1.4, Input/Output With Schmitt Trigger

Figure 6-8 shows the port diagram. Table 6-18 summarizes the selection of the pin functions.



<sup>\*</sup> Note: MSP430G2x33 devices only. MSP430G2x03 devices have no ADC10.

Figure 6-8. Port P1 (P1.4) Diagram



## Table 6-18. Port P1 (P1.4) Pin Functions

			CONTROL BITS OR SIGNALS <sup>(1)</sup>					
PIN NAME (P1.x)	x	FUNCTION	P1DIR.x	P1SEL.x	P1SEL2.x	ADC10AE.x (INCH.y = 1) <sup>(2)</sup>	JTAG Mode	
P1.4/		P1.x (I/O)	I: 0; O: 1	0	0	0	0	
SMCLK/		SMCLK	1	1	0	0	0	
UCB0STE/		UCB0STE <sup>(3)(4)</sup>	from USCI	1	1	0	0	
UCA0CLK/		UCA0CLK <sup>(3)(4)</sup>	from USCI	1	1	0	0	
VREF+(2)/	4	VREF+	Х	Х	Х	1	0	
VEREF+(2)/		VEREF+	Х	Х	Х	1	0	
A4 <sup>(2)</sup> /		A4	Х	Х	Х	1 (y = 4)	0	
TCK/		TCK	Х	Х	Х	0	1	
Pin Osc		Capacitive sensing	Х	0	1	0	0	

<sup>(1)</sup> X = don't care

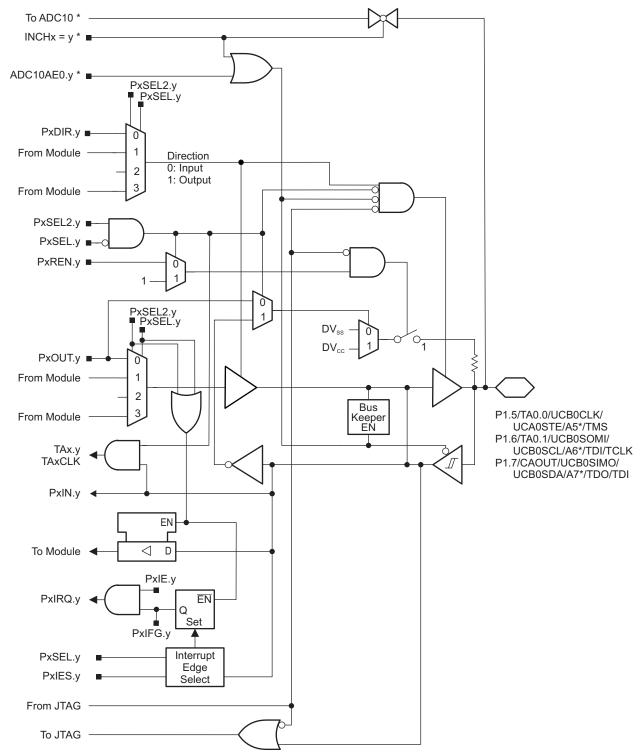
<sup>(2)</sup> MSP430G2x33 devices only

<sup>3)</sup> The pin direction is controlled by the USCI module.

<sup>(4)</sup> UCAOCLK function takes precedence over UCBOSTE function. If the pin is required as UCAOCLK input or output, USCI\_B0 is forced to 3-wire SPI mode if 4-wire SPI mode is selected.

## 6.10.4 Port P1 Pin Diagram: P1.5 to P1.7, Input/Output With Schmitt Trigger

Figure 6-9 shows the port diagram. Table 6-19 summarizes the selection of the pin functions.



<sup>\*</sup> Note: MSP430G2x33 devices only. MSP430G2x03 devices have no ADC10.

Figure 6-9. Port P1 (P1.5 to P1.7) Diagram



## Table 6-19. Port P1 (P1.5 to P1.7) Pin Functions

				CONTRO	OL BITS OR SIG	NALS <sup>(1)</sup>	
PIN NAME (P1.x)	х	FUNCTION	P1DIR.x	P1SEL.x	P1SEL2.x	ADC10AE.x (INCH.y = 1) <sup>(2)</sup>	JTAG Mode
P1.5/		P1.x (I/O)	I: 0; O: 1	0	0	0	0
TA0.0/		TA0.0	1	1	0	0	0
UCB0CLK/		UCB0CLK <sup>(3)(4)</sup>	from USCI	1	1	0	0
UCA0STE/	5	UCA0STE <sup>(3)(4)</sup>	from USCI	1	1	0	0
A5 <sup>(2)</sup> /		A5	Х	Х	Х	1 (y = 5)	0
TMS		TMS	Х	Х	Х	0	1
Pin Osc		Capacitive sensing	Х	0	1	0	0
P1.6/		P1.x (I/O)	I: 0; O: 1	0	0	0	0
TA0.1/		TA0.1	1	1	0	0	0
UCB0SOMI/		UCB0SOMI	from USCI	1	1	0	0
UCB0SCL/	6	UCB0SCL	from USCI	1	1	0	0
A6 <sup>(2)</sup> /		A6	Х	Х	Х	1 (y = 6)	0
TDI/TCLK/		TDI/TCLK	Х	Х	Х	0	1
Pin Osc		Capacitive sensing	Х	0	1	0	0
P1.7/		P1.x (I/O)	I: 0; O: 1	0	0	0	0
UCB0SIMO/		UCB0SIMO	from USCI	1	1	0	0
UCB0SDA/	7	UCB0SDA	from USCI	1	1	0	0
A7 <sup>(2)</sup> /		A7	Х	Х	Х	1 (y = 7)	0
TDO/TDI/		TDO/TDI	Х	Х	Х	0	1
Pin Osc		Capacitive sensing	Х	0	1	0	0

X = don't care

MSP430G2x33 devices only

The pin direction is controlled by the USCI module.

UCB0CLK function takes precedence over UCA0STE function. If the pin is required as UCB0CLK input or output, USCI\_A0 is forced to 3-wire SPI mode if 4-wire SPI mode is selected.

# 6.10.5 Port P2 Pin Diagram: P2.0 to P2.5, Input/Output With Schmitt Trigger

Figure 6-10 shows the port diagram. Table 6-20 summarizes the selection of the pin functions.

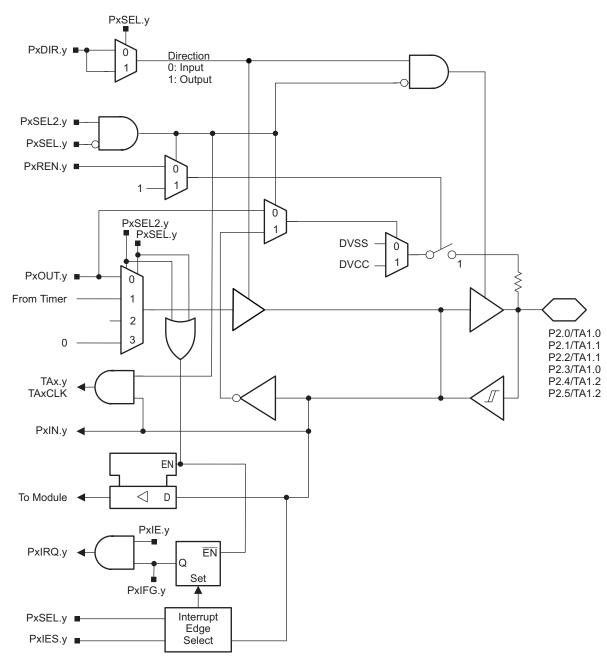


Figure 6-10. Port P2 (P2.0 to P2.5) Diagram



## Table 6-20. Port P2 (P2.0 to P2.5) Pin Functions

PIN NAME		FUNCTION	CONTRO	CONTROL BITS OR SIGNALS <sup>(1)</sup>			
(P2.x)	X	FUNCTION	P2DIR.x	P2SEL.x	P2SEL2.x		
P2.0/		P2.x (I/O)	I: 0; O: 1	0	0		
TA1.0/		Timer1_A3.CCI0A	0	1	0		
	0	Timer1_A3.TA0	1	1	0		
Pin Osc		Capacitive sensing	Х	0	1		
P2.1/		P2.x (I/O)	I: 0; O: 1	0	0		
TA1.1/		Timer1_A3.CCI1A	0	1	0		
	1	Timer1_A3.TA1	1	1	0		
Pin Osc		Capacitive sensing	Х	0	1		
P2.2/		P2.x (I/O)	I: 0; O: 1	0	0		
TA1.1/	2	Timer1_A3.CCI1B	0	1	0		
	2	Timer1_A3.TA1	1	1	0		
Pin Osc		Capacitive sensing	Х	0	1		
P2.3/		P2.x (I/O)	I: 0; O: 1	0	0		
TA1.0/		Timer1_A3.CCI0B	0	1	0		
	3	Timer1_A3.TA0	1	1	0		
Pin Osc		Capacitive sensing	X	0	1		
P2.4/		P2.x (I/O)	I: 0; O: 1	0	0		
TA1.2/		Timer1_A3.CCI2A	0	1	0		
	4	Timer1_A3.TA2	1	1	0		
Pin Osc		Capacitive sensing	X	0	1		
P2.5/		P2.x (I/O)	I: 0; O: 1	0	0		
TA1.2/	_	Timer1_A3.CCI2B	0	1	0		
	5	Timer1_A3.TA2	1	1	0		
Pin Osc		Capacitive sensing	X	0	1		

<sup>(1)</sup> X = don't care

## 6.10.6 Port P2 Pin Diagram: P2.6, Input/Output With Schmitt Trigger

Figure 6-11 shows the port diagram. Table 6-21 summarizes the selection of the pin functions.

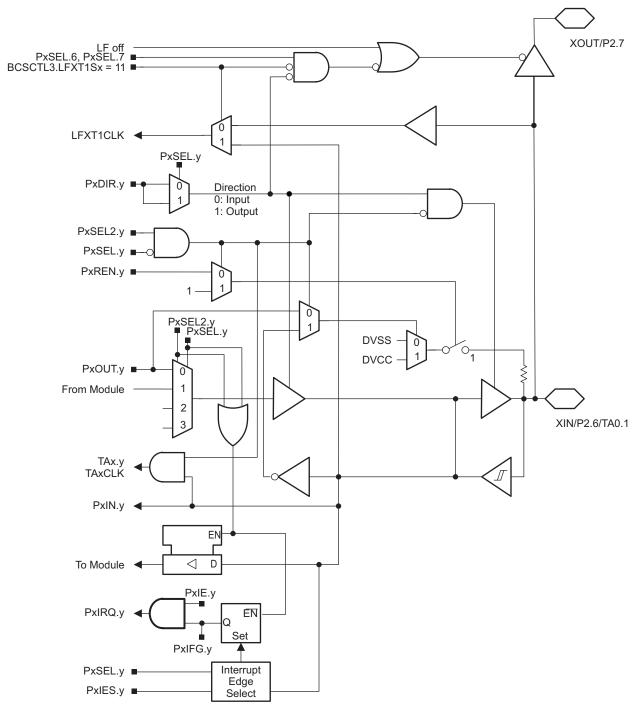


Figure 6-11. Port P2 (P2.6) Diagram



# Table 6-21. Port P2 (P2.6) Pin Functions

PIN NAME			CONTROL BITS OR SIGNALS <sup>(1)</sup>			
(P2.x)	х	FUNCTION	P2DIR.x	P2SEL.6 P2SEL.7	P2SEL2.6 P2SEL2.7	
XIN	6	XIN	0	1 1	0	
P2.6		P2.x (I/O)	I: 0; O: 1	0 X	0	
TA0.1		ь	Timer0_A3.TA1	1	1 0	0
Pin Osc		Capacitive sensing	Х	0 X	1 X	

<sup>(1)</sup> X = don't care



## 6.10.7 Port P2 Pin Diagram: P2.7, Input/Output With Schmitt Trigger

Figure 6-12 shows the port diagram. Table 6-22 summarizes the selection of the pin functions.

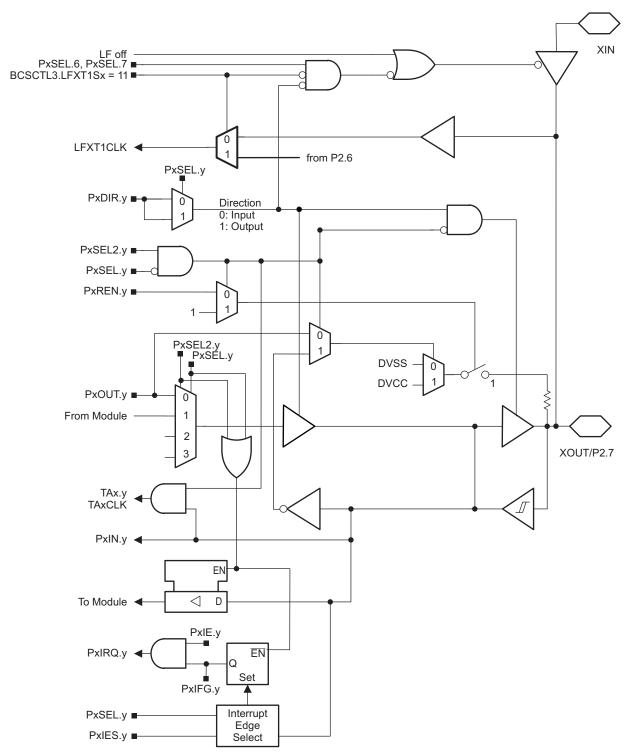


Figure 6-12. Port P2 (P2.7) Diagram



# Table 6-22. Port P2 (P2.7) Pin Functions

PIN NAME			CONTROL BITS OR SIGNALS <sup>(1)</sup>			
(P2.x)	х	FUNCTION	P2DIR.x	P2SEL.6 P2SEL.7	P2SEL2.6 P2SEL2.7	
XOUT/		XOUT	1	1 1	0	
P2.7/	7	P2.x (I/O)	I: 0; O: 1	0 X	0	
Pin Osc		Capacitive sensing	Х	0 X	1 X	

<sup>(1)</sup> X = don't care



# 6.10.8 Port P3 Pin Diagram: P3.0 to P3.7, Input/Output With Schmitt Trigger (RHB and PW28 Package Only)

Figure 6-13 shows the port diagram. Table 6-23 summarizes the selection of the pin functions.

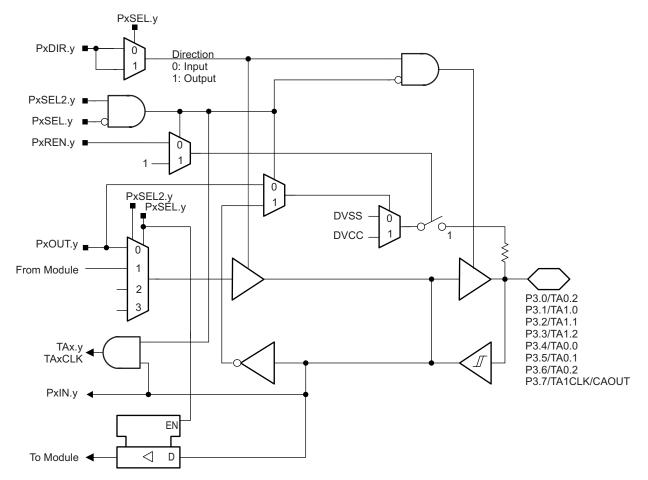


Figure 6-13. Port P3 (P3.0 to P3.7) Diagram (RHB and PW28 Package Only)



## Table 6-23. Port P3 (P3.0 to P3.7) Pin Functions (RHB and PW28 Package Only)

PIN NAME	x	FUNCTION	CONTRO	CONTROL BITS OR SIGNALS <sup>(1)</sup>			
(P3.x)			P3DIR.x	P3SEL.x	P3SEL2.x		
P3.0/		P3.x (I/O)	I: 0; O: 1	0	0		
TA0.2/	0	Timer0_A3.CCI2A	0	1	0		
	U	Timer0_A3.TA2	1	1	0		
Pin Osc		Capacitive sensing	X	0	1		
P3.1/		P3.x (I/O)	I: 0; O: 1	0	0		
TA1.0/	1	Timer1_A3.TA0	1	1	0		
Pin Osc		Capacitive sensing	X	0	1		
P3.2/		P3.x (I/O)	I: 0; O: 1	0	0		
TA1.1/	2	Timer1_A3.TA1	1	1	0		
Pin Osc		Capacitive sensing	X	0	1		
P3.3/		P3.x (I/O)	I: 0; O: 1	0	0		
TA1.2/	3	Timer1_A3.TA2	1	1	0		
Pin Osc		Capacitive sensing	X	0	1		
P3.4/		P3.x (I/O)	I: 0; O: 1	0	0		
TA0.0/	4	Timer0_A3.TA0	1	1	0		
Pin Osc		Capacitive sensing	Х	0	1		
P3.5/		P3.x (I/O)	I: 0; O: 1	0	0		
TA0.1/	5	Timer0_A3.TA1	1	1	0		
Pin Osc		Capacitive sensing	Х	0	1		
P3.6/		P3.x (I/O)	I: 0; O: 1	0	0		
TA0.2/	6	Timer0_A3.TA2	1	1	0		
Pin Osc		Capacitive sensing	Х	0	1		
P3.7/		P3.x (I/O)	I: 0; O: 1	0	0		
TA1CLK/	7	Timer1_A3.TACLK	0	1	0		
Pin Osc		Capacitive sensing	Х	0	1		

<sup>(1)</sup> X = don't care



# 7 Device and Documentation Support

#### 7.1 Getting Started and Next Steps

For more information on the MSP430<sup>™</sup> family of devices and the tools and libraries that are available to help with your development, visit the Getting Started page.

#### 7.2 Device Nomenclature

To designate the stages in the product development cycle, TI assigns prefixes to the part numbers of all MSP430 MCU devices and support tools. Each MSP430 MCU commercial family member has one of three prefixes: MSP, PMS, or XMS (for example, MSP430F5438A). TI recommends two of three possible prefix designators for its support tools: MSP and MSPX. These prefixes represent evolutionary stages of product development from engineering prototypes (with XMS for devices and MSPX for tools) through fully qualified production devices and tools (with MSP for devices and MSP for tools).

Device development evolutionary flow:

**XMS** – Experimental device that is not necessarily representative of the electrical specifications for the final device

**PMS** – Final silicon die that conforms to the electrical specifications for the device but has not completed quality and reliability verification

MSP - Fully qualified production device

Support tool development evolutionary flow:

MSPX – Development-support product that has not yet completed Tl's internal qualification testing.

MSP - Fully-qualified development-support product

XMS and PMS devices and MSPX development-support tools are shipped against the following disclaimer:

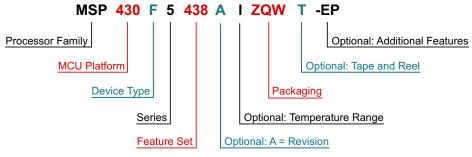
"Developmental product is intended for internal evaluation purposes."

MSP devices and MSP development-support tools have been characterized fully, and the quality and reliability of the device have been demonstrated fully. Tl's standard warranty applies.

Predictions show that prototype devices (XMS and PMS) have a greater failure rate than the standard production devices. TI recommends that these devices not be used in any production system because their expected end-use failure rate still is undefined. Only qualified production devices are to be used.

TI device nomenclature also includes a suffix with the device family name. This suffix indicates the package type (for example, PZP) and temperature range (for example, T). Figure 7-1 provides a legend for reading the complete device name for any family member.





Processor Family  MCU Platform	CC = Embedded RF Radio MSP = Mixed-Signal Processor XMS = Experimental Silicon PMS = Prototype Device  430 = MSP430 low-power microcontroller platform			
Device Type	Memory Type C = ROM F = Flash FR = FRAM G = Flash or FRAM (Value Line) L = No Nonvolatile Memory	Specialized Application AFE = Analog Front End BT = Preprogrammed with Bluetooth BQ = Contactless Power CG = ROM Medical FE = Flash Energy Meter FG = Flash Medical FW = Flash Electronic Flow Meter		
Series	1 Series = Up to 8 MHz 2 Series = Up to 16 MHz 3 Series = Legacy 4 Series = Up to 16 MHz with LCD	5 Series = Up to 25 MHz 6 Series = Up to 25 MHz with LCD 0 = Low-Voltage Series		
Feature Set	Various Levels of Integration Within a Series			
Optional: A = Revision	N/A			
Optional: Temperature Range	S = 0°C to 50°C C = 0°C to 70°C I = -40°C to 85°C T = -40°C to 105°C			
Packaging	http://www.ti.com/packaging			
Optional: Tape and Reel	T = Small Reel R = Large Reel No Markings = Tube or Tray			
Optional: Additional Features	-EP = Enhanced Product (-40°C to 105°C) -HT = Extreme Temperature Parts (-55°C to 150°C) -Q1 = Automotive Q100 Qualified			

Figure 7-1. Device Nomenclature



#### 7.3 Tools and Software

All MSP microcontrollers are supported by a wide variety of software and hardware development tools. Tools are available from TI and various third parties. See them all at MSP Tools.

Table 7-1 lists the debug features of these devices. See the Code Composer Studio for MSP430 User's Guide (SLAU157) for details on the available features.

**Table 7-1. Hardware Features** 

MSP430 ARCHITECTURE	4-WIRE JTAG	2-WIRE JTAG	BREAK- POINTS (N)	RANGE BREAK- POINTS	CLOCK CONTROL	STATE SEQUENCER	TRACE BUFFER	LPMx.5 DEBUGGING SUPPORT
MSP430	Yes	Yes	2	No	Yes	No	No	No

#### **Design Kits and Evaluation Modules**

- 28-Pin Target Development Board and MSP-FET USB Programmer Bundle for MSP430F2x and MSP430G2x MCUs The MSP-FET430U28A kit includes all of the hardware and software required to quickly begin application development on the MSP430 MCU. This kit includes a ZIF socket target board (MSP-TS430PW28A) that accepts some MSP430 devices in 20- or 28-pin TSSOP packages (TI Package Code: PW). It is also bundled with a USB flash emulation tool (MSP-FET) that interfaces the target board to a PC, allowing developers to program and debug their MSP430 devices through in-system emulation through the JTAG interface or the pin-saving Spy Bi-Wire (2-wire JTAG) protocol.
- MSP430 LaunchPad™ Value Line Development Kit The MSP-EXP430G2 LaunchPad Development Kit is an easy-to-use microcontroller development board for the low-power and low-cost MSP430G2x MCUs. It has on-board emulation for programming and debugging and features a 14- or 20-pin DIP socket, on-board buttons and LEDs and BoosterPack Plug-in Module pinouts that support a wide range of modules for added functionality such as wireless, displays, and more.
- MSP430 Capacitive Touch BoosterPack™ Plug-in Module The Capacitive Touch BoosterPack (430BOOST-SENSE1) is a plug-in module for MCU LaunchPad Development Kits. This BoosterPack also includes a preprogrammed MSP430G2452IN20 Value Line device for the MSP-EXP430G2 LaunchPad. Developers can use this BoosterPack as a solution for adding capacitive touch differentiation in many applications such as consumer electronics, point of sales machines, and other devices with a physical button.

#### **Software**

- MSP430G2x53, MSP430G2x33, MSP430G2x13, MSP430G2x03 Code Examples C Code examples are available for every MSP device that configures each of the integrated peripherals for various application needs.
- MSPWare™ Software MSPWare software is a collection of code examples, data sheets, and other design resources for all MSP devices delivered in a convenient package. In addition to providing a complete collection of existing MSP design resources, MSPWare software also includes a high-level API called MSP Driver Library. This library makes it easy to program MSP hardware. MSPWare software is available as a component of CCS or as a stand-alone package.
- MSP Driver Library Driver Library's abstracted API keeps you above the bits and bytes of the MSP430 hardware by providing easy-to-use function calls. Thorough documentation is delivered through a helpful API Guide, which includes details on each function call and the recognized parameters. Developers can use Driver Library functions to write complete projects with minimal overhead.



- Capacitive Touch Software Library Free C libraries for enabling capacitive touch capabilities on MSP430 MCUs and MSP432 MCUs. The MSP430 MCU version of the library features several capacitive touch implementations including the RO and RC method.
- MSP EnergyTrace™ Technology EnergyTrace technology for MSP430 microcontrollers is an energy-based code analysis tool that measures and displays the application's energy profile and helps to optimize it for ultra-low-power consumption.
- ULP (Ultra-Low Power) Advisor ULP Advisor™ software is a tool for guiding developers to write more efficient code to fully utilize the unique ultra-low power features of MSP and MSP432 microcontrollers. Aimed at both experienced and new microcontroller developers, ULP Advisor checks your code against a thorough ULP checklist to squeeze every last nano amp out of your application. At build time, ULP Advisor will provide notifications and remarks to highlight areas of your code that can be further optimized for lower power.
- IEC60730 Software Package The IEC60730 MSP430 software package was developed to be useful in assisting customers in complying with IEC 60730-1:2010 (Automatic Electrical Controls for Household and Similar Use Part 1: General Requirements) for up to Class B products, which includes home appliances, arc detectors, power converters, power tools, e-bikes, and many others. The IEC60730 MSP430 software package can be embedded in customer applications running on MSP430s to help simplify the customer's certification efforts of functional safety-compliant consumer devices to IEC 60730-1:2010 Class B.
- Fixed-Point Math Library for MSP The MSP IQmath and Qmath Libraries are a collection of highly optimized and high-precision mathematical functions for C programmers to seamlessly port a floating-point algorithm into fixed-point code on MSP430 and MSP432 devices. These routines are typically used in computationally intensive real-time applications where optimal execution speed, high accuracy, and ultra-low energy are critical. By using the IQmath and Qmath libraries, it is possible to achieve execution speeds considerably faster and energy consumption considerably lower than equivalent code written using floating-point math.

#### **Development Tools**

- Code Composer Studio™ Integrated Development Environment for MSP Microcontrollers

  Code

  Composer Studio is an integrated development environment (IDE) that supports all MSP microcontroller devices. Code Composer Studio comprises a suite of embedded software utilities used to develop and debug embedded applications. It includes an optimizing C/C++ compiler, source code editor, project build environment, debugger, profiler, and many other features. The intuitive IDE provides a single user interface taking you through each step of the application development flow. Familiar utilities and interfaces allow users to get started faster than ever before. Code Composer Studio combines the advantages of the Eclipse software framework with advanced embedded debug capabilities from TI resulting in a compelling feature-rich development environment for embedded developers. When using CCS with an MSP MCU, a unique and powerful set of plugins and embedded software utilities are made available to fully leverage the MSP microcontroller.
- Grace Graphical Peripheral Configuration Tool Enable and configure ADCs, DACs, timers, clocks, serial communication interfaces, and more, by interacting with buttons, drop-down menus, and text fields. Navigate through the MSP430 MCUs highly integrated peripheral set with ease.
- MSP Flasher Command Line Programmer MSP Flasher is an open-source shell-based interface for programming MSP microcontrollers through a FET programmer or eZ430 using JTAG or Spy-Bi-Wire (SBW) communication. MSP Flasher can download binary files (.txt or .hex) files directly to the MSP microcontroller without an IDE.



- MSP MCU Programmer and Debugger The MSP-FET is a powerful emulation development tool often called a debug probe which allows users to quickly begin application development on MSP low-power microcontrollers (MCU). Creating MCU software usually requires downloading the resulting binary program to the MSP device for validation and debugging. The MSP-FET provides a debug communication pathway between a host computer and the target MSP. Furthermore, the MSP-FET also provides a Backchannel UART connection between the computer's USB interface and the MSP UART. This affords the MSP programmer a convenient method for communicating serially between the MSP and a terminal running on the computer. It also supports loading programs (often called firmware) to the MSP target using the BSL (bootloader) through the UART and I<sup>2</sup>C communication protocols.
- MSP-GANG Production Programmer The MSP Gang Programmer is an MSP430 or MSP432 device programmer that can program up to eight identical MSP430 or MSP432 Flash or FRAM devices at the same time. The MSP Gang Programmer connects to a host PC using a standard RS-232 or USB connection and provides flexible programming options that allow the user to fully customize the process. The MSP Gang Programmer is provided with an expansion board, called the Gang Splitter, that implements the interconnections between the MSP Gang Programmer and multiple target devices. Eight cables are provided that connect the expansion board to eight target devices (through JTAG or Spy-Bi-Wire connectors). The programming can be done with a PC or as a stand-alone device. A PC-side graphical user interface is also available and is DLL-based.

#### 7.4 Documentation Support

The following documents describe the MSP430G2x33 and MSP430G2x03 devices. Copies of these documents are available on the Internet at www.ti.com.

#### **Receiving Notification of Document Updates**

To receive notification of documentation updates—including silicon errata—go to the product folder for your device on ti.com (for example, MSP430G2533). In the upper right corner, click the "Alert me" button. This registers you to receive a weekly digest of product information that has changed (if any). For change details, check the revision history of any revised document.

#### **Errata**

- MSP430G2533 Device Erratasheet Describes the known exceptions to the functional specifications for the MSP430G2533 device.
- MSP430G2433 Device Erratasheet Describes the known exceptions to the functional specifications for the MSP430G2433 device.
- MSP430G2333 Device Erratasheet Describes the known exceptions to the functional specifications for the MSP430G2333 device.
- MSP430G2233 Device Erratasheet Describes the known exceptions to the functional specifications for the MSP430G2233 device.
- MSP430G2403 Device Erratasheet Describes the known exceptions to the functional specifications for the MSP430G2403 device.
- MSP430G2303 Device Erratasheet Describes the known exceptions to the functional specifications for the MSP430G2303 device.
- MSP430G2203 Device Erratasheet Describes the known exceptions to the functional specifications for the MSP430G2203 device.

www.ti.com

#### **User's Guides**

- MSP430x2xx Family User's Guide Detailed information on the modules and peripherals available in this device family.
- Code Composer Studio v6.1 for MSP430 User's Guide This manual describes the use of TI Code Composer Studio IDE v6.1 (CCS v6.1) with the MSP430 ultra-low-power microcontrollers. This document applies only for the Windows version of the Code Composer Studio IDE. The Linux version is similar and, therefore, is not described separately.
- IAR Embedded Workbench Version 3+ for MSP430 User's Guide This manual describes the use of IAR Embedded Workbench (EW430) with the MSP430 ultra-low-power microcontrollers.
- MSP430 Programming With the Bootloader (BSL) The MSP430 bootloader (BSL, formerly known as the bootstrap loader) allows users to communicate with embedded memory in the MSP430 microcontroller during the prototyping phase, final production, and in service. Both the programmable memory (flash memory) and the data memory (RAM) can be modified as required. Do not confuse the bootloader with the bootstrap loader programs found in some digital signal processors (DSPs) that automatically load program code (and data) from external memory to the internal memory of the DSP.
- MSP430 Programming Via the JTAG Interface This document describes the functions that are required to erase, program, and verify the memory module of the MSP430 flash-based and FRAM-based microcontroller families using the JTAG communication port. In addition, it describes how to program the JTAG access security fuse that is available on all MSP430 devices. This document describes device access using both the standard 4-wire JTAG interface and the 2-wire JTAG interface, which is also referred to as Spy-Bi-Wire (SBW).
- MSP430 Hardware Tools User's Guide This manual describes the hardware of the TI MSP-FET430 Flash Emulation Tool (FET). The FET is the program development tool for the MSP430 ultra-low-power microcontroller. Both available interface types, the parallel port interface and the USB interface, are described.

#### **Application Reports**

- MSP430 32-kHz Crystal Oscillators Selection of the right crystal, correct load circuit, and proper board layout are important for a stable crystal oscillator. This application report summarizes crystal oscillator function and explains the parameters to select the correct crystal for MSP430 ultra-low-power operation. In addition, hints and examples for correct board layout are given. The document also contains detailed information on the possible oscillator tests to ensure stable oscillator operation in mass production.
- MSP430 System-Level ESD Considerations System-Level ESD has become increasingly demanding with silicon technology scaling towards lower voltages and the need for designing cost-effective and ultra-low-power components. This application report addresses three different ESD topics to help board designers and OEMs understand and design robust system-level designs: (1) Component-level ESD testing and system-level ESD testing, their differences and why component-level ESD rating does not ensure system-level robustness. (2) General design guidelines for system-level ESD protection at different levels including enclosures, cables, PCB layout, and on-board ESD protection devices. (3) Introduction to System Efficient ESD Design (SEED), a co-design methodology of on-board and on-chip ESD protection to achieve system-level ESD robustness, with example simulations and test results. A few real-world system-level ESD protection design examples and their results are also discussed.



- General Oversampling of MSP ADCs for Higher Resolution Multiple MSP ultra-low-power microcontrollers offer analog-to-digital converters (ADCs) to convert physical quantities into digital numbers, a function that is widely used across numerous applications. There are times, however, when a customer design demands a higher resolution than the ADC of the selected MSP can offer. This application report, which is based on the previously-published Oversampling the ADC12 for Higher Resolution (SLAA323), therefore describes how an oversampling method can be incorporated to increase ADC resolution past the currently available number of bits.
- Capacitive Touch Hardware Design Guide Capacitive touch detection is sometimes considered more art than science. This often results in multiple design iterations before the optimum performance is achieved. There are, however, good design practices for circuit layout and principles of materials that need to be understood to keep the number of iterations to a minimum. This design guide describes a process for creating and designing capacitive touch solutions, starting with the schematic, working through the mechanicals, and finally designing the electrodes for the application.
- Capacitive Touch Sensing, MSP430 Slider and Wheel Tuning Guide This application report provides guidelines on how to tune capacitive touch sliders and wheels running on the MSP430™ microcontrollers. It identifies the hardware and software parameters as well as explains the steps used in tuning sliders and wheels. The slider and wheel tuning is based on the APIs defined in the Capacitive Touch Sense Library (CAPSENSELIBRARY).
- Capacitive Touch Sensing, MSP430 Button Gate Time Optimization and Tuning Guide MSP430™ microcontroller based capacitive touch buttons can offer increased performance when properly optimized and tuned for their specific application. Performance benefits that result from button optimization can include, but are not limited to, decreased power consumption, improved response time, and the ability to grow a design to include more buttons. This application report provides the reader with a starting point for button design at the system and software level.

#### 7.5 Related Links

Table 7-2 lists quick access links. Categories include technical documents, support and community resources, tools and software, and quick access to sample or buy.

Table 7-2. Related Links

PARTS	PRODUCT FOLDER	SAMPLE & BUY	TECHNICAL DOCUMENTS	TOOLS & SOFTWARE	SUPPORT & COMMUNITY
MSP430G2533	Click here	Click here	Click here	Click here	Click here
MSP430G2433	Click here	Click here	Click here	Click here	Click here
MSP430G2333	Click here	Click here	Click here	Click here	Click here
MSP430G2233	Click here	Click here	Click here	Click here	Click here
MSP430G2403	Click here	Click here	Click here	Click here	Click here
MSP430G2303	Click here	Click here	Click here	Click here	Click here
MSP430G2203	Click here	Click here	Click here	Click here	Click here



#### 7.6 Community Resources

The following links connect to TI community resources. Linked contents are provided "AS IS" by the respective contributors. They do not constitute TI specifications and do not necessarily reflect TI's views; see TI's Terms of Use.

#### TI E2E™ Community

TI's Engineer-to-Engineer (E2E) Community. Created to foster collaboration among engineers. At e2e.ti.com, you can ask questions, share knowledge, explore ideas, and help solve problems with fellow engineers.

#### TI Embedded Processors Wiki

Texas Instruments Embedded Processors Wiki. Established to help developers get started with embedded processors from Texas Instruments and to foster innovation and growth of general knowledge about the hardware and software surrounding these devices.

#### 7.7 Trademarks

MSP430, LaunchPad, BoosterPack, MSPWare, EnergyTrace, ULP Advisor, Code Composer Studio, E2E are trademarks of Texas Instruments.

All other trademarks are the property of their respective owners.

#### 7.8 Electrostatic Discharge Caution



These devices have limited built-in ESD protection. The leads should be shorted together or the device placed in conductive foam during storage or handling to prevent electrostatic damage to the MOS gates.

#### 7.9 Glossary

TI Glossary This glossary lists and explains terms, acronyms, and definitions.

www.ti.com

## 8 Mechanical, Packaging, and Orderable Information

The following pages include mechanical, packaging, and orderable information. This information is the most current data available for the designated devices. This data is subject to change without notice and revision of this document. For browser-based versions of this data sheet, refer to the left-hand navigation.





13-Jan-2019

## **PACKAGING INFORMATION**

Orderable Device	Status	Package Type		Pins	_	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	<b>Device Marking</b>	Samples
	(1)		Drawing		Qty	(2)	(6)	(3)		(4/5)	
MSP430G2203IN20	ACTIVE	PDIP	N	20	20	Pb-Free (RoHS)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	M430G2203	Samples
MSP430G2203IPW20	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2203	Samples
MSP430G2203IPW20R	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2203	Samples
MSP430G2203IPW28	ACTIVE	TSSOP	PW	28	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2203	Samples
MSP430G2203IPW28R	ACTIVE	TSSOP	PW	28	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2203	Samples
MSP430G2203IRHB32R	ACTIVE	VQFN	RHB	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	MSP430 G2203	Samples
MSP430G2233IN20	ACTIVE	PDIP	N	20	20	Pb-Free (RoHS)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	M430G2233	Samples
MSP430G2233IPW20	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2233	Samples
MSP430G2233IPW20R	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2233	Samples
MSP430G2233IPW28	ACTIVE	TSSOP	PW	28	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2233	Samples
MSP430G2233IPW28R	ACTIVE	TSSOP	PW	28	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2233	Samples
MSP430G2233IRHB32R	ACTIVE	VQFN	RHB	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	MSP430 G2233	Samples
MSP430G2233IRHB32T	ACTIVE	VQFN	RHB	32	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	MSP430 G2233	Samples
MSP430G2303IN20	OBSOLETE	PDIP	N	20		TBD	Call TI	Call TI	-40 to 85	M430G2303	
MSP430G2303IPW20	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2303	Samples
MSP430G2303IPW20R	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2303	Samples
MSP430G2303IPW28	ACTIVE	TSSOP	PW	28	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2303	Samples





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Orderable Device	Status	Package Type	Package Drawing	Pins	Package Qty	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking (4/5)	Samples
MSP430G2303IPW28R	ACTIVE	TSSOP	PW	28	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2303	Samples
MSP430G2303IRHB32R	ACTIVE	VQFN	RHB	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	MSP430 G2303	Samples
MSP430G2303IRHB32T	ACTIVE	VQFN	RHB	32	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	MSP430 G2303	Samples
MSP430G2333IN20	ACTIVE	PDIP	N	20	20	Pb-Free (RoHS)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	M430G2333	Samples
MSP430G2333IPW20	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2333	Samples
MSP430G2333IPW20R	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2333	Samples
MSP430G2333IPW28	ACTIVE	TSSOP	PW	28	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2333	Samples
MSP430G2333IPW28R	ACTIVE	TSSOP	PW	28	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2333	Samples
MSP430G2333IRHB32R	ACTIVE	VQFN	RHB	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	MSP430 G2333	Samples
MSP430G2333IRHB32T	ACTIVE	VQFN	RHB	32	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	MSP430 G2333	Samples
MSP430G2403IN20	ACTIVE	PDIP	N	20	20	Pb-Free (RoHS)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	M430G2403	Samples
MSP430G2403IPW20	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2403	Samples
MSP430G2403IPW20R	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2403	Samples
MSP430G2403IPW28	ACTIVE	TSSOP	PW	28	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2403	Samples
MSP430G2403IPW28R	ACTIVE	TSSOP	PW	28	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2403	Samples
MSP430G2403IRHB32R	ACTIVE	VQFN	RHB	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	MSP430 G2403	Samples
MSP430G2403IRHB32T	ACTIVE	VQFN	RHB	32	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	MSP430 G2403	Samples
MSP430G2433IN20	ACTIVE	PDIP	N	20	20	Pb-Free (RoHS)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	M430G2433	Samples





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Orderable Device	Status	Package Type	Package	Pins	Package	Eco Plan	Lead/Ball Finish	MSL Peak Temp	Op Temp (°C)	Device Marking	Samples
	(1)		Drawing		Qty	(2)	(6)	(3)		(4/5)	
MSP430G2433IPW20	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2433	Samples
MSP430G2433IPW20R	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2433	Samples
MSP430G2433IPW28	ACTIVE	TSSOP	PW	28	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2433	Samples
MSP430G2433IPW28R	ACTIVE	TSSOP	PW	28	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2433	Samples
MSP430G2433IRHB32R	ACTIVE	VQFN	RHB	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	MSP430 G2433	Samples
MSP430G2433IRHB32T	ACTIVE	VQFN	RHB	32	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	MSP430 G2433	Samples
MSP430G2533IN20	ACTIVE	PDIP	N	20	20	Pb-Free (RoHS)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	M430G2533	Samples
MSP430G2533IPW20	ACTIVE	TSSOP	PW	20	70	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2533	Sample
MSP430G2533IPW20R	ACTIVE	TSSOP	PW	20	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2533	Samples
MSP430G2533IPW28	ACTIVE	TSSOP	PW	28	50	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2533	Samples
MSP430G2533IPW28R	ACTIVE	TSSOP	PW	28	2000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-1-260C-UNLIM	-40 to 85	430G2533	Samples
MSP430G2533IRHB32R	ACTIVE	VQFN	RHB	32	3000	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	MSP430 G2533	Sample
MSP430G2533IRHB32T	ACTIVE	VQFN	RHB	32	250	Green (RoHS & no Sb/Br)	CU NIPDAU	Level-2-260C-1 YEAR	-40 to 85	MSP430 G2533	Samples

<sup>(1)</sup> The marketing status values are defined as follows:

**ACTIVE:** Product device recommended for new designs.

LIFEBUY: TI has announced that the device will be discontinued, and a lifetime-buy period is in effect.

NRND: Not recommended for new designs. Device is in production to support existing customers, but TI does not recommend using this part in a new design.

**PREVIEW**: Device has been announced but is not in production. Samples may or may not be available.

**OBSOLETE:** TI has discontinued the production of the device.

<sup>(2)</sup> RoHS: TI defines "RoHS" to mean semiconductor products that are compliant with the current EU RoHS requirements for all 10 RoHS substances, including the requirement that RoHS substance do not exceed 0.1% by weight in homogeneous materials. Where designed to be soldered at high temperatures, "RoHS" products are suitable for use in specified lead-free processes. TI may reference these types of products as "Pb-Free".



## PACKAGE OPTION ADDENDUM

13-Jan-2019

RoHS Exempt: TI defines "RoHS Exempt" to mean products that contain lead but are compliant with EU RoHS pursuant to a specific EU RoHS exemption.

Green: TI defines "Green" to mean the content of Chlorine (CI) and Bromine (Br) based flame retardants meet JS709B low halogen requirements of <=1000ppm threshold. Antimony trioxide based flame retardants must also meet the <=1000ppm threshold requirement.

- (3) MSL, Peak Temp. The Moisture Sensitivity Level rating according to the JEDEC industry standard classifications, and peak solder temperature.
- (4) There may be additional marking, which relates to the logo, the lot trace code information, or the environmental category on the device.
- (5) Multiple Device Markings will be inside parentheses. Only one Device Marking contained in parentheses and separated by a "~" will appear on a device. If a line is indented then it is a continuation of the previous line and the two combined represent the entire Device Marking for that device.
- (6) Lead/Ball Finish Orderable Devices may have multiple material finish options. Finish options are separated by a vertical ruled line. Lead/Ball Finish values may wrap to two lines if the finish value exceeds the maximum column width.

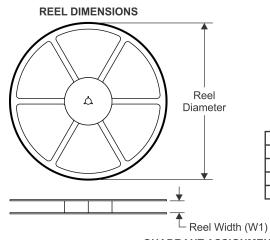
**Important Information and Disclaimer:** The information provided on this page represents TI's knowledge and belief as of the date that it is provided. TI bases its knowledge and belief on information provided by third parties, and makes no representation or warranty as to the accuracy of such information. Efforts are underway to better integrate information from third parties. TI has taken and continues to take reasonable steps to provide representative and accurate information but may not have conducted destructive testing or chemical analysis on incoming materials and chemicals. TI and TI suppliers consider certain information to be proprietary, and thus CAS numbers and other limited information may not be available for release.

In no event shall TI's liability arising out of such information exceed the total purchase price of the TI part(s) at issue in this document sold by TI to Customer on an annual basis.

PACKAGE MATERIALS INFORMATION

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## TAPE AND REEL INFORMATION



# TAPE DIMENSIONS KO P1 BO W Cavity AO

	Dimension designed to accommodate the component width
B0	Dimension designed to accommodate the component length
K0	Dimension designed to accommodate the component thickness
W	Overall width of the carrier tape
P1	Pitch between successive cavity centers

## QUADRANT ASSIGNMENTS FOR PIN 1 ORIENTATION IN TAPE



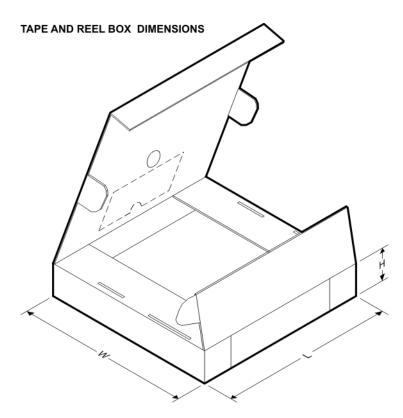
\*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
MSP430G2203IPW28R	TSSOP	PW	28	2000	330.0	16.4	6.9	10.2	1.8	12.0	16.0	Q1
MSP430G2203IPW28R	TSSOP	PW	28	2000	330.0	16.4	6.9	10.2	1.8	12.0	16.0	Q1
MSP430G2203IRHB32R	VQFN	RHB	32	3000	330.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2
MSP430G2233IPW20R	TSSOP	PW	20	2000	330.0	16.4	6.95	7.1	1.6	8.0	16.0	Q1
MSP430G2233IPW28R	TSSOP	PW	28	2000	330.0	16.4	6.9	10.2	1.8	12.0	16.0	Q1
MSP430G2233IPW28R	TSSOP	PW	28	2000	330.0	16.4	6.9	10.2	1.8	12.0	16.0	Q1
MSP430G2233IRHB32R	VQFN	RHB	32	3000	330.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2
MSP430G2233IRHB32T	VQFN	RHB	32	250	180.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2
MSP430G2303IPW28R	TSSOP	PW	28	2000	330.0	16.4	6.9	10.2	1.8	12.0	16.0	Q1
MSP430G2303IPW28R	TSSOP	PW	28	2000	330.0	16.4	6.9	10.2	1.8	12.0	16.0	Q1
MSP430G2303IRHB32R	VQFN	RHB	32	3000	330.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2
MSP430G2303IRHB32T	VQFN	RHB	32	250	180.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2
MSP430G2333IPW28R	TSSOP	PW	28	2000	330.0	16.4	6.9	10.2	1.8	12.0	16.0	Q1
MSP430G2333IRHB32R	VQFN	RHB	32	3000	330.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2
MSP430G2333IRHB32T	VQFN	RHB	32	250	180.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2
MSP430G2403IPW28R	TSSOP	PW	28	2000	330.0	16.4	6.9	10.2	1.8	12.0	16.0	Q1
MSP430G2403IRHB32R	VQFN	RHB	32	3000	330.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2
MSP430G2403IRHB32T	VQFN	RHB	32	250	180.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2

# **PACKAGE MATERIALS INFORMATION**

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Device	Package Type	Package Drawing		SPQ	Reel Diameter (mm)	Reel Width W1 (mm)	A0 (mm)	B0 (mm)	K0 (mm)	P1 (mm)	W (mm)	Pin1 Quadrant
MSP430G2433IPW28R	TSSOP	PW	28	2000	330.0	16.4	6.9	10.2	1.8	12.0	16.0	Q1
MSP430G2433IRHB32R	VQFN	RHB	32	3000	330.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2
MSP430G2433IRHB32T	VQFN	RHB	32	250	180.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2
MSP430G2533IPW28R	TSSOP	PW	28	2000	330.0	16.4	6.9	10.2	1.8	12.0	16.0	Q1
MSP430G2533IPW28R	TSSOP	PW	28	2000	330.0	16.4	6.9	10.2	1.8	12.0	16.0	Q1
MSP430G2533IRHB32R	VQFN	RHB	32	3000	330.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2
MSP430G2533IRHB32T	VQFN	RHB	32	250	180.0	12.4	5.3	5.3	1.1	8.0	12.0	Q2



## \*All dimensions are nominal

Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
MSP430G2203IPW28R	TSSOP	PW	28	2000	367.0	367.0	38.0
MSP430G2203IPW28R	TSSOP	PW	28	2000	350.0	350.0	43.0
MSP430G2203IRHB32R	VQFN	RHB	32	3000	367.0	367.0	35.0
MSP430G2233IPW20R	TSSOP	PW	20	2000	367.0	367.0	38.0
MSP430G2233IPW28R	TSSOP	PW	28	2000	367.0	367.0	38.0
MSP430G2233IPW28R	TSSOP	PW	28	2000	350.0	350.0	43.0
MSP430G2233IRHB32R	VQFN	RHB	32	3000	367.0	367.0	35.0
MSP430G2233IRHB32T	VQFN	RHB	32	250	210.0	185.0	35.0
MSP430G2303IPW28R	TSSOP	PW	28	2000	350.0	350.0	43.0
MSP430G2303IPW28R	TSSOP	PW	28	2000	367.0	367.0	38.0



# **PACKAGE MATERIALS INFORMATION**

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Device	Package Type	Package Drawing	Pins	SPQ	Length (mm)	Width (mm)	Height (mm)
MSP430G2303IRHB32R	VQFN	RHB	32	3000	367.0	367.0	35.0
MSP430G2303IRHB32T	VQFN	RHB	32	250	210.0	185.0	35.0
MSP430G2333IPW28R	TSSOP	PW	28	2000	350.0	350.0	43.0
MSP430G2333IRHB32R	VQFN	RHB	32	3000	367.0	367.0	35.0
MSP430G2333IRHB32T	VQFN	RHB	32	250	210.0	185.0	35.0
MSP430G2403IPW28R	TSSOP	PW	28	2000	350.0	350.0	43.0
MSP430G2403IRHB32R	VQFN	RHB	32	3000	367.0	367.0	35.0
MSP430G2403IRHB32T	VQFN	RHB	32	250	210.0	185.0	35.0
MSP430G2433IPW28R	TSSOP	PW	28	2000	350.0	350.0	43.0
MSP430G2433IRHB32R	VQFN	RHB	32	3000	367.0	367.0	35.0
MSP430G2433IRHB32T	VQFN	RHB	32	250	210.0	185.0	35.0
MSP430G2533IPW28R	TSSOP	PW	28	2000	350.0	350.0	43.0
MSP430G2533IPW28R	TSSOP	PW	28	2000	367.0	367.0	38.0
MSP430G2533IRHB32R	VQFN	RHB	32	3000	367.0	367.0	35.0
MSP430G2533IRHB32T	VQFN	RHB	32	250	210.0	185.0	35.0

PW (R-PDSO-G20)

## PLASTIC SMALL OUTLINE



- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M—1994.
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.
- E. Falls within JEDEC MO-153



# PW (R-PDSO-G20)

# PLASTIC SMALL OUTLINE



- All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
  C. Publication IPC-7351 is recommended for alternate design.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



# N (R-PDIP-T\*\*)

## PLASTIC DUAL-IN-LINE PACKAGE

16 PINS SHOWN

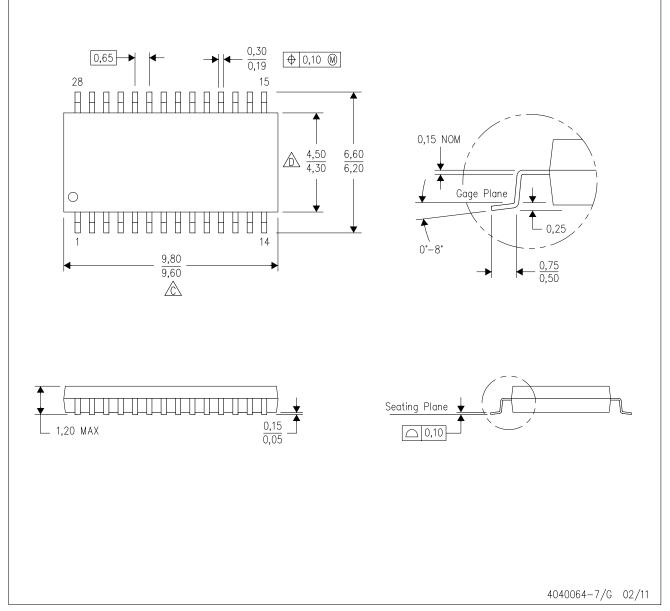


- A. All linear dimensions are in inches (millimeters).
- B. This drawing is subject to change without notice.
- Falls within JEDEC MS-001, except 18 and 20 pin minimum body length (Dim A).
- The 20 pin end lead shoulder width is a vendor option, either half or full width.



PW (R-PDSO-G28)

## PLASTIC SMALL OUTLINE

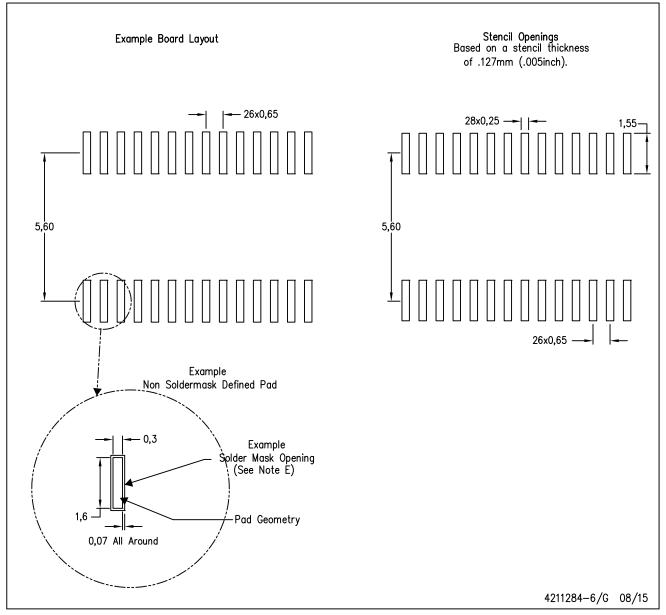


- A. All linear dimensions are in millimeters. Dimensioning and tolerancing per ASME Y14.5M—1994.
- B. This drawing is subject to change without notice.
- Body length does not include mold flash, protrusions, or gate burrs. Mold flash, protrusions, or gate burrs shall not exceed 0,15 each side.
- Body width does not include interlead flash. Interlead flash shall not exceed 0,25 each side.
- E. Falls within JEDEC MO-153



# PW (R-PDSO-G28)

# PLASTIC SMALL OUTLINE



- All linear dimensions are in millimeters.
- B. This drawing is subject to change without notice.
  C. Publication IPC-7351 is recommended for alternate design.
- D. Laser cutting apertures with trapezoidal walls and also rounding corners will offer better paste release. Customers should contact their board assembly site for stencil design recommendations. Refer to IPC-7525 for other stencil recommendations.
- E. Customers should contact their board fabrication site for solder mask tolerances between and around signal pads.



5 x 5, 0.5 mm pitch

PLASTIC QUAD FLATPACK - NO LEAD



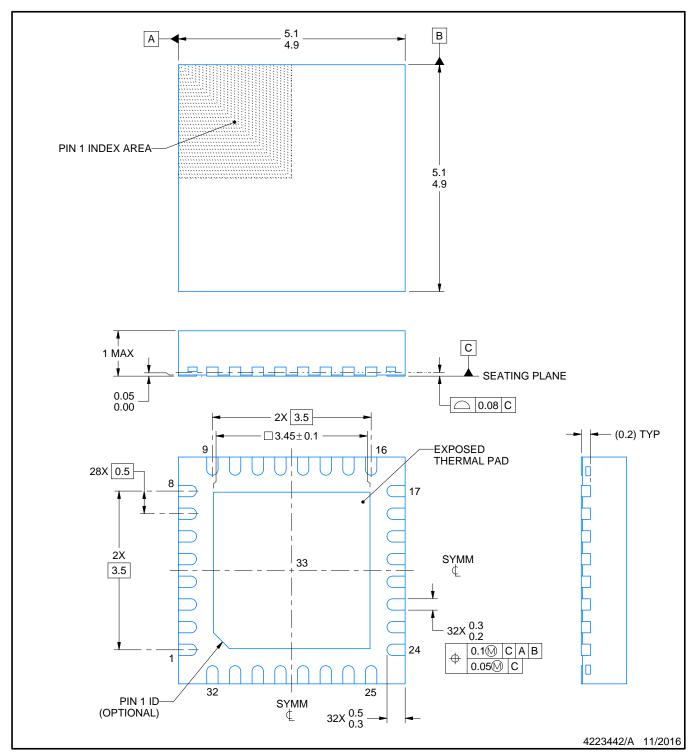
Images above are just a representation of the package family, actual package may vary. Refer to the product data sheet for package details.

4224745/A





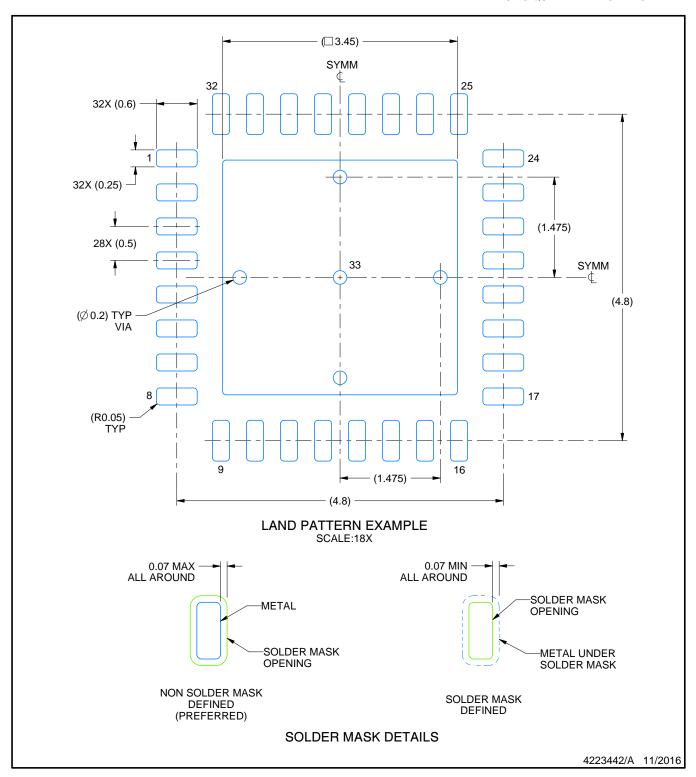
PLASTIC QUAD FLATPACK - NO LEAD



- 1. All linear dimensions are in millimeters. Any dimensions in parenthesis are for reference only. Dimensioning and tolerancing per ASME Y14.5M.
  2. This drawing is subject to change without notice.
- 3. The package thermal pad must be soldered to the printed circuit board for thermal and mechanical performance.



PLASTIC QUAD FLATPACK - NO LEAD

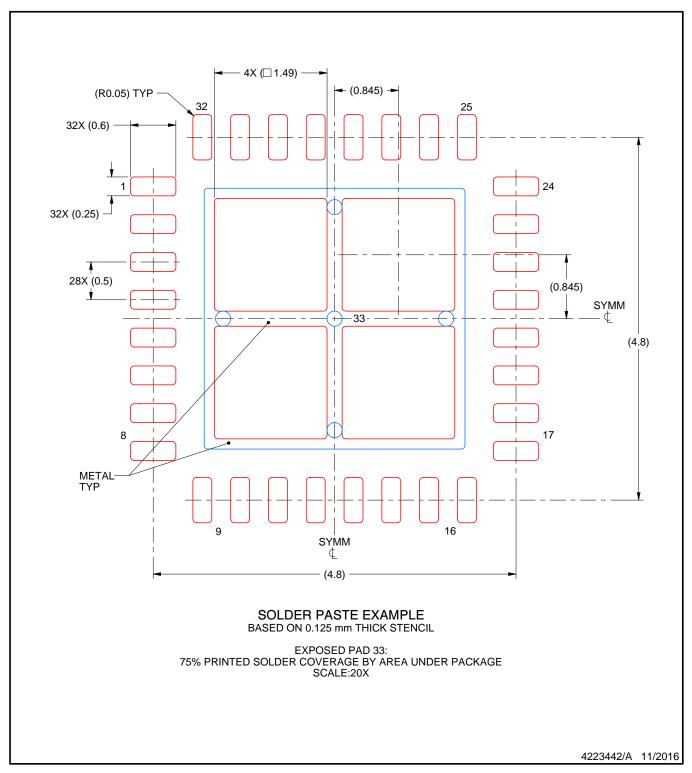


NOTES: (continued)

- 4. This package is designed to be soldered to a thermal pad on the board. For more information, see Texas Instruments literature number SLUA271 (www.ti.com/lit/slua271).
- Vias are optional depending on application, refer to device data sheet. If any vias are implemented, refer to their locations shown on this view. It is recommended that vias under paste be filled, plugged or tented.



PLASTIC QUAD FLATPACK - NO LEAD



NOTES: (continued)

6. Laser cutting apertures with trapezoidal walls and rounded corners may offer better paste release. IPC-7525 may have alternate design recommendations.



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